

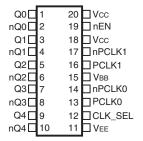
DATA SHEET

General Description

The ICS853S014I is a low skew, high performance 1-to-5, 2.5V/3.3V Differential-to-LVPECL/ECL Fanout Buffer. The ICS853S014I has two selectable clock inputs.

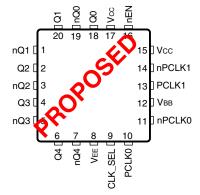
Guaranteed output and part-to-part skew characteristics make the ICS853S014I ideal for those applications demanding well defined performance and repeatability.

Pin Assignments



ICS853S014I

20-Lead TSSOP 6.5mm x 4.4mm x 0.925mm package body G Package Top View



ICS853S014I

20-Lead VFQFN 4mm x 4mm x 0.925mm package body K Package Top View

Features

- Five differential LVPECL/ECL outputs
- Two selectable differential LVPECL clock inputs
- PCLKx, nPCLKx pairs can accept the following differential input levels: LVPECL, LVDS, CML, SSTL
- Maximum output frequency: 2GHz
- Output skew: 55ps (maximum)
- Part-to-part skew: 100ps (maximum)
- Propagation delay: 500ps (maximum)
- Additive phase jitter, RMS: 0.10ps (maximum)
- LVPECL mode operating voltage supply range:
 V_{CC} = 2.375V to 3.8V, V_{EE} = 0V
- ECL mode operating voltage supply range: $V_{CC} = 0V$, $V_{EE} = -3.8V$ to -2.375V
- -40°C to 85°C ambient operating temperature
- · Available in lead-free (RoHS 6) packages

Block Diagram

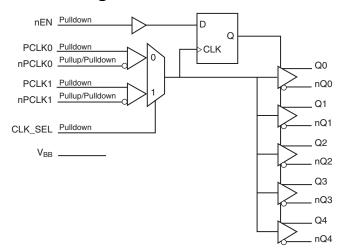


Table 1. Pin Descriptions

Number	Name	1	уре	Description
1, 2	Q0, nQ0	Output		Differential output pair. LVPECL/ECL interface levels.
3, 4	Q1, nQ1	Output		Differential output pair. LVPECL/ECL interface levels.
5, 6	Q2, nQ2	Output		Differential output pair. LVPECL/ECL interface levels.
7, 8	Q3, nQ3	Output		Differential output pair. LVPECL/ECL interface levels.
9, 10	Q4, nQ4	Output		Differential output pair. LVPECL/ECL interface levels.
11	V _{EE}	Power		Negative supply pin.
12	CLK_SEL	Input	Pulldown	Clock select input. When HIGH, selects PCLK1, nPCLK1 inputs. When LOW, selects PCLK0, nPCLK0 inputs. Single-ended LVPECL interface levels.
13	PCLK0	Input	Pulldown	Non-inverting differential LVPECL clock input.
14	nPCLK0	Input	Pullup/ Pulldown	Inverting differential LVPECL clock input. V _{CC} /2 default when left floating.
15	V_{BB}	Output		Bias voltage.
16	PCLK1	Input	Pulldown	Non-inverting differential LVPECL clock input.
17	nPCLK1	Input	Pullup/ Pulldown	Inverting differential LVPECL clock input. V _{CC} /2 default when left floating.
18, 20	V _{CC}	Power		Positive supply pins.
19	nEN	Input	Pulldown	Synchronizing clock enable. When LOW, clock outputs follow clock input. When HIGH, Qx outputs are forced low, nQx outputs are forced high. Single-ended LVPECL interface levels.

NOTE: Pulldown refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

Table 2. Pin Characteristics

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
R _{PULLDOWN}	Input Pulldown Resistor			37		kΩ
R _{VCC/2}	Pullup/Pulldown Resistors			37		kΩ

Function Tables

Table 3A. Control Input Function Table

	Inputs	Outputs			
nEN	CLK_SEL	Selected Source	Q0:Q4	nQ0:nQ4	
1	0	PCLK0, nPCLK0	Disabled; Low	Disabled; High	
1	1	PCLK1, nPCLK1	Disabled; Low	Disabled; High	
0	0	PCLK0, nPCLK0	Enabled	Enabled	
0	1	PCLK1, nPCLK1	Enabled	Enabled	

After nEN switches, the clock outputs are disabled or enabled following a falling input clock edge as shown in *Figure 1*. In the active mode, the state of the outputs are a function of the PCLK0, nPCLK0 and PCLK1, nPCLK1 inputs as described in Table 3B.

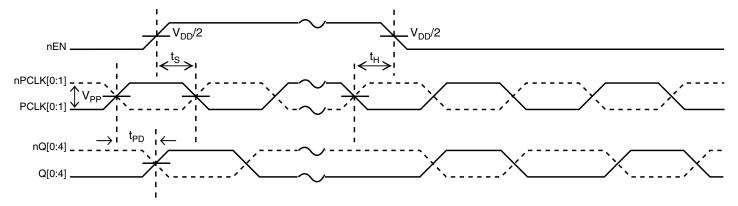


Figure 1. nEN Timing Diagram

Table 3B. Clock Input Function Table

Inj	Ou	tputs			
PCLK0 or PCLK1	nPCLK0 or nPCLK1	Q0:Q4	nQ0:nQ4	Input to Output Mode	Polarity
0	1	LOW	HIGH	Differential to Differential	Non-Inverting
1	0	HIGH	LOW	Differential to Differential	Non-Inverting
0	Biased; NOTE 1	LOW	HIGH	Single-Ended to Differential	Non-Inverting
1	Biased; NOTE 1	HIGH	LOW	Single-Ended to Differential	Non-Inverting
Biased; NOTE 1	0	HIGH	LOW	Single-Ended to Differential	Inverting
Biased; NOTE 1	1	LOW	HIGH	Single-Ended to Differential	Inverting

NOTE 1: Please refer to the Application Information section. Wiring the Differential Input to Accept Single-ended Levels.

Absolute Maximum Ratings

NOTE: Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

Item	Rating
Supply Voltage, V _{CC}	4.6V (LVPECL mode, V _{EE} = 0V)
Negative Supply Voltage, V _{EE}	-4.6V (ECL mode, V _{CC} = 0V)
Inputs, V _I (LVPECL mode)	-0.5V to V _{CC} + 0.5V
Inputs, V _I (ECL mode)	0.5V to V _{EE} – 0.5V
Outputs, I _O Continuous Current Surge Current	50mA 100mA
V _{BB} Sink//Source, I _{BB}	± 0.5mA
Operating Temperature Range, T _A	-40°C to +85°C
Package Thermal Impedance, θ _{JA} 20 Lead TSSOP 20 Lead VFQFN PROPOSED	92.1°C/W (0 mps) 62.2°C/W (0 mps)
Storage Temperature, T _{STG}	-65°C to 150°C

DC Electrical Characteristics

Table 4A. Power Supply DC Characteristics, V_{CC} = 2.375V to 3.8V; V_{EE} = 0V, T_A = -40°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V _{CC}	Positive Supply Voltage		2.375	3.3	3.8	V
I _{EE}	Power Supply Current				68	mA

Table 4B. DC Characteristics, $V_{CC} = 3.3V$; $V_{EE} = 0V$, $T_A = -40^{\circ}C$ to $85^{\circ}C$

				-40°C			25°C			85°C		
Symbol	Parameter		Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Units
V _{OH}	Output High Vo	ltage; NOTE 1	2.175	2.275	2.50	2.225	2.295	2.495	2.22	2.295	2.485	V
V _{OL}	Output Low Vol	tage; NOTE 1	1.405	1.545	1.68	1.425	1.52	1.615	1.44	1.535	1.63	V
V _{IH}	Input High Voltag	ge (CLK_SEL, nEN)	2.075		2.36	2.075		2.36	2.075		2.36	V
V _{IL}	Input Low Voltag	e (CLK_SEL, nEN)	1.43		1.765	1.43		1.765	1.43		1.765	V
V _{BB}	Output Voltage NOTE 2	Reference;	1.72		1.98	1.72		1.98	1.72		1.98	V
V _{CMR}	Input High Volta Range; NOTE	age Common Mode 3	1.2		3.3	1.2		3.3	1.2		3.3	V
V _{PP}	Peak-to-Peak II NOTE 4	nput Voltage;	150	800	1200	150	800	1200	150	800	1200	mV
I _{IH}	Input High Current	PCLK0, PCLK1 nPCLK0, nPCLK1			150			150			150	μΑ
	Input	PCLK0, PCLK1	-10			-10			-10			μΑ
IIL	Low Current	nPCLK0, nPCLK1	-150			-150			-150			μΑ

NOTE: Input and output parameters vary 1:1 with V_{CC} . V_{EE} can vary +0.925V to -0.5V.

NOTE 1: Outputs terminated with 50 $\!\Omega$ to V $_{CC}$ – 2V.

NOTE 2: Single-ended input operation is limited. $V_{CC} \ge 3V$ in LVPECL mode.

NOTE 3: Common mode voltage is defined as VIH for the differential inputs.

NOTE 4: The V_{CMB} and V_{PP} levels should be such that input low voltage never goes below V_{EE}.

Table 4C. LVPECL DC Characteristics, $V_{CC} = 2.5V$; $V_{EE} = 0V$, $T_A = -40^{\circ}C$ to $85^{\circ}C$

			-40°C		25°C			85°C				
Symbol	Parameter		Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Units
V _{OH}	Output High Vo	Output High Voltage; NOTE 1		1.475	1.70	1.425	1.495	1.69	1.42	1.495	1.685	V
V _{OL}	Output Low Vol	Output Low Voltage; NOTE 1		0.745	0.88	0.625	0.72	0.86	0.64	0.735	0.85	V
V _{IH}	Input High Voltag	ge (CLK_SEL, nEN)	1.275		1.56	1.275		1.56	1.275		1.56	V
V _{IL}	Input Low Voltag	ge (CLK_SEL, nEN)	0.63		0.965	0.63		0.965	0.63		0.965	V
V _{CMR}	Input High Volta Range; NOTE 2	age Common Mode 2	1.2		2.5	1.2		2.5	1.2		2.5	V
V _{PP}	Peak-to-Peak II NOTE 3	nput Voltage;	150	800	1200	150	800	1200	150	800	1200	mV
I _{IH}	Input High Current	PCLK0, PCLK1 nPCLK0, nPCLK1			150			150			150	μА
	Input	PCLK0, PCLK1	-10			-10			-10			μΑ
IIL	Low Current	nPCLK0, nPCLK1	-150			-150			-150			μΑ

NOTE: Input and output parameters vary 1:1 with V_{CC} . V_{EE} can vary +0.925V to -0.5V.

NOTE 1: Outputs terminated with 50Ω to V_{CC} – 2V.

NOTE 2: Common mode voltage is defined as V_{IH} for the differential inputs.

NOTE 3: The V_{CMR} and V_{PP} levels should be such that input low voltage never goes below V_{EE}.

Table 4D. ECL DC Characteristics, V_{CC} = 0V; V_{EE} = -3.8V to -2.375V, T_A = -40°C to 85°C

				-40°C			25°C			85°C		
Symbol	Parameter		Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Units
V _{OH}	Output High V	oltage; NOTE 1	-1.125	-1.025	-0.80	-1.075	-1.005	-0.805	-1.08	-1.005	-0.815	V
V _{OL}	Output Low Vo	oltage; NOTE 1	-1.895	-1.755	-1.62	-1.875	-1.78	-1.685	-1.86	-1.765	-1.67	V
V _{IH}	Input High Voltage (CLK_SEL, nEN)		-1.225		-0.94	-1.225		-0.94	-1.225		-0.94	V
V _{IL}	Input Low Volta	ge (CLK_SEL, nEN)	-1.87		-1.535	-1.87		-1.535	-1.87		-1.535	V
V_{BB}	Output Voltage NOTE 2	e Reference;	-1.58		-1.32	-1.58		-1.32	-1.58		-1.32	V
V_{CMR}	Input High Vol Mode Range;	tage Common NOTE 3	V _{EE} +1.2		0	V _{EE} +1.2		0	V _{EE} +1.2		0	V
V_{PP}	Peak-to-Peak NOTE 4	Input Voltage;	150	800	1200	150	800	1200	150	800	1200	mV
I _{IH}	Input High Current	PCLK0, PCLK1 nPCLK0, nPCLK1			150			150			150	μΑ
	la a d	PCLK0, PCLK1	-10			-10			-10			μΑ
I _{IL}	Input Low Current	nPCLK0, nPCLK1	-150			-150			-150			μΑ

NOTE: Input and output parameters vary 1:1 with V_{CC} . V_{EE} can vary +0.925V to -0.5V. NOTE 1: Outputs terminated with 50Ω to $V_{CC}-2V$. NOTE 2: Single-ended input operation is limited. $V_{CC} \ge 3V$ in LVPECL mode. NOTE 3: Common mode voltage is defined as V_{IH} for the differential inputs.

NOTE 4: The V_{CMR} and V_{PP} levels should be such that input low voltage never goes below V_{EE} .

AC Electrical Characteristics

Table 5. AC Characteristics, $V_{CC} = -3.8V$ to -2.375V or , $V_{CC} = 2.375V$ to 3.8V; $V_{EE} = 0V$, $T_A = -40^{\circ}C$ to $85^{\circ}C$

	Parameter			-40°C		25°C		85°C				
Symbol			Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Units
f _{MAX}	Output Frequence	ру			2			2			2	GHz
t _{PD}	Propagation Dela	ay; NOTE 1	250		425	300		450	350		500	ps
<i>t</i> jit	Buffer Additive P RMS; refer to Ad Jitter Section (<i>f</i> = 12kHz - 20MHz)	lditive Phase = 156.25MHz,		0.06	0.10		0.07	0.10		0.08	0.10	ps
tsk(o)	Output Skew; NO	OTE 2, 4			55			55			55	ps
tsk(pp)	Part-to-Part Skev	w; NOTE 3, 4			100			100			100	ps
t _R / t _F	Output Rise/Fall Time	20% to 80%	70		220	80		220	90		220	ps
t _S	Clock Enable Se	tup Time	100	50		100	50		100	50		ps
t _H	Clock Enable Ho	ld Time	200	140		200	140		200	140		ps

NOTE: All parameters are measured at $f \le 1$ GHz, unless otherwise noted.

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE 1: Measured from the differential input crossing point to the differential output crossing point.

NOTE 2: Defined as skew between outputs at the same supply voltage and with equal load conditions.

Measured at the output differential cross points.

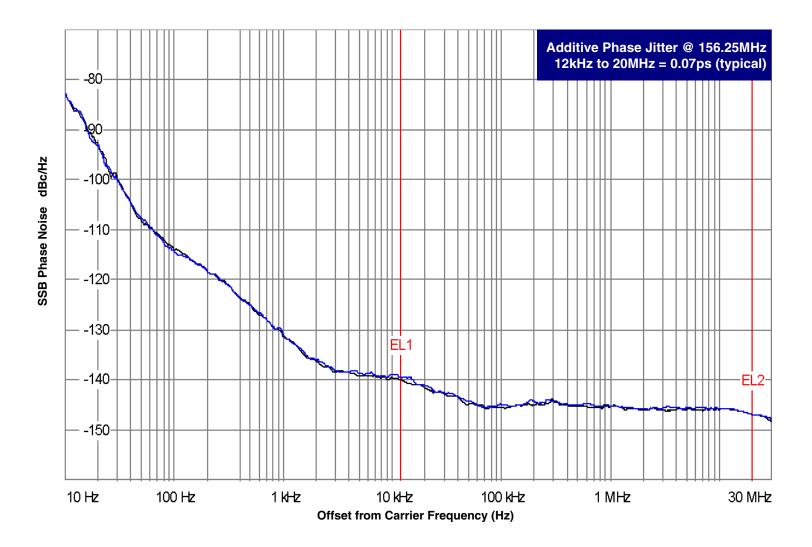
NOTE 3: Defined as skew between outputs on different devices operating at the same supply voltage, same temperature, same frequency and with equal load conditions. Using the same type of inputs on each device, the outputs are measured at the differential cross points.

NOTE 4: This parameter is defined in accordance with JEDEC Standard 65.

Additive Phase Jitter

The spectral purity in a band at a specific offset from the fundamental compared to the power of the fundamental is called the *dBc Phase Noise*. This value is normally expressed using a Phase noise plot and is most often the specified plot in many applications. Phase noise is defined as the ratio of the noise power present in a 1Hz band at a specified offset from the fundamental frequency to the power value of the fundamental. This ratio is expressed in decibels (dBm) or a ratio

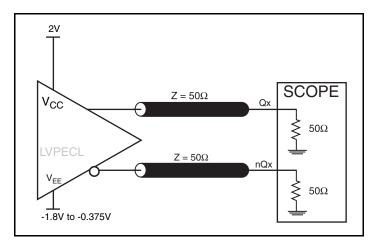
of the power in the 1Hz band to the power in the fundamental. When the required offset is specified, the phase noise is called a *dBc* value, which simply means dBm at a specified offset from the fundamental. By investigating jitter in the frequency domain, we get a better understanding of its effects on the desired application over the entire time record of the signal. It is mathematically possible to calculate an expected bit error rate given a phase noise plot.



As with most timing specifications, phase noise measurements has issues relating to the limitations of the equipment. Often the noise floor of the equipment is higher than the noise floor of the device. This is illustrated above. The device meets the noise floor of what is shown, but can actually be lower. The phase noise is dependent on the input source and measurement equipment.

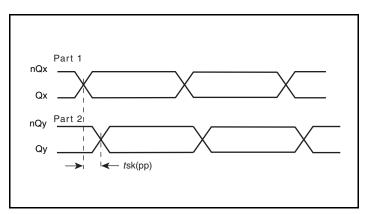
The source generator "IFR2042 10kHz – 5.4GHz Low Noise Signal Generator used as external input to an Agilent 8133A 3GHz Pulse Generator".

Parameter Measurement Information

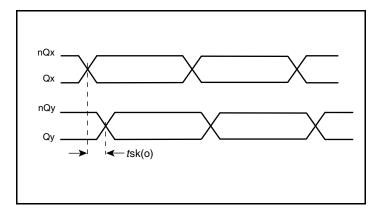


PCLKx — Cross Points — V_{CMR} — V_{CMR}

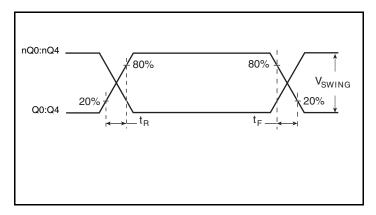
LVPECL Output Load AC Test Circuit



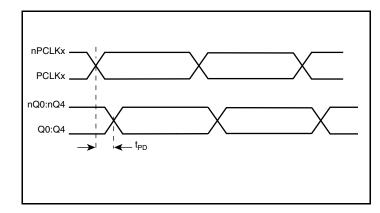
Differential Input Level



Part-to-Part Skew



Output Skew



Output Rise/Fall Time

Propagation Delay

Applications Information

Wiring the Differential Input to Accept Single-Ended Levels

Figure 2 shows how a differential input can be wired to accept single ended levels. The reference voltage $V_{REF} = V_{CC}/2$ is generated by the bias resistors R1 and R2. The bypass capacitor (C1) is used to help filter noise on the DC bias. This bias circuit should be located as close to the input pin as possible. The ratio of R1 and R2 might need to be adjusted to position the V_{REF} in the center of the input voltage swing. For example, if the input clock swing is 2.5V and $V_{CC} = 3.3V$, R1 and R2 value should be adjusted to set V_{REF} at 1.25V. The values below are for when both the single ended swing and V_{CC} are at the same voltage. This configuration requires that the sum of the output impedance of the driver (Ro) and the series resistance (Rs) equals the transmission line impedance. In addition, matched termination at the input will attenuate the signal in half. This can be done in one of two ways. First, R3 and R4 in parallel should equal the transmission

line impedance. For most 50Ω applications, R3 and R4 can be 100Ω . The values of the resistors can be increased to reduce the loading for slower and weaker LVCMOS driver. When using single-ended signaling, the noise rejection benefits of differential signaling are reduced. Even though the differential input can handle full rail LVCMOS signaling, it is recommended that the amplitude be reduced. The datasheet specifies a lower differential amplitude, however this only applies to differential signals. For single-ended applications, the swing can be larger, however $V_{\rm IL}$ cannot be less than -0.3V and $V_{\rm IH}$ cannot be more than $V_{\rm CC}$ + 0.3V. Though some of the recommended components might not be used, the pads should be placed in the layout. They can be utilized for debugging purposes. The datasheet specifications are characterized and guaranteed by using a differential signal.

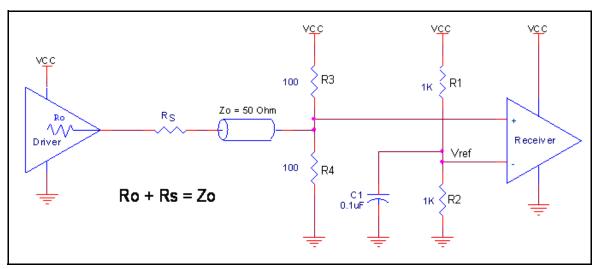


Figure 2. Recommended Schematic for Wiring a Differential Input to Accept Single-ended Levels

LVPECL Clock Input Interface

The PCLK/nPCLK accepts LVPECL, LVDS, CML, SSTL and other differential signals. Both V_{SWING} and V_{OH} must meet the V_{PP} and V_{CMR} input requirements. Figures 3A to 3F show interface examples for the PCLK/nPCLK input driven by the most common driver types.

The input interfaces suggested here are examples only. If the driver is from another vendor, use their termination recommendation. Please consult with the vendor of the driver component to confirm the driver termination requirements.

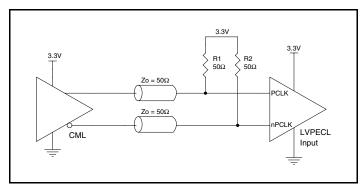


Figure 3A. PCLK/nPCLK Input Driven by an Open Collector CML Driver

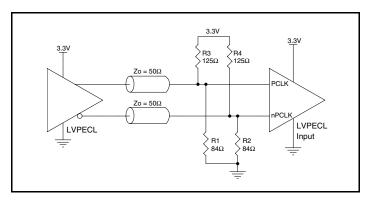


Figure 3C. PCLK/nPCLK Input Driven by a 3.3V LVPECL Driver

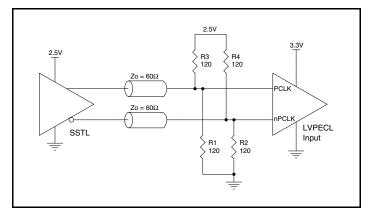


Figure 3E. PCLK/nPCLK Input Driven by an SSTL Driver

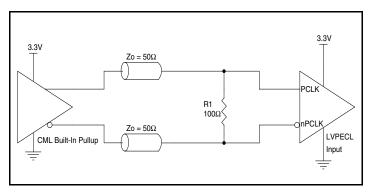


Figure 3B. PCLK/nPCLK Input Driven by a Built-In Pullup CML Driver

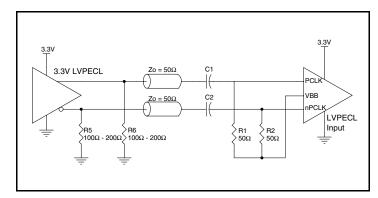


Figure 3D. PCLK/nPCLK Input Driven by a 3.3V LVPECL Driver with AC Couple

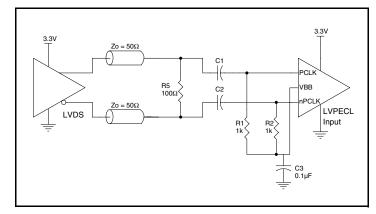


Figure 3F. PCLK/nPCLK Input Driven by a 3.3V LVDS Driver

Recommendations for Unused Output Pins

Inputs:

PCLK/nPCLK Inputs

For applications not requiring the use of a differential input, both the PCLK and nPCLK pins can be left floating. Though not required, but for additional protection, a 1k Ω resistor can be tied from PCLK to ground.

LVCMOS Control Pins

All control pins have internal pulldowns; additional resistance is not required but can be added for additional protection. A $1k\Omega$ resistor can be used.

Outputs:

LVPECL Outputs

All unused LVPECL outputs can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

Termination for 3.3V LVPECL Outputs

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

The differential outputs are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to drive 50Ω

transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. *Figures 4A and 4B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

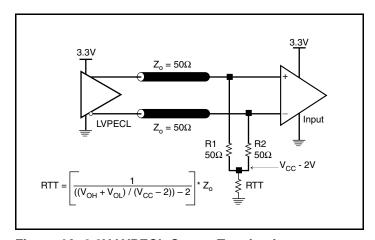


Figure 4A. 3.3V LVPECL Output Termination

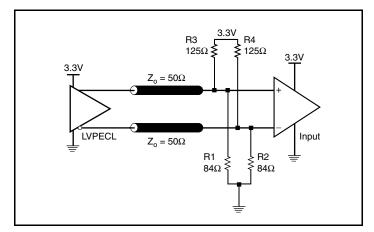


Figure 4B. 3.3V LVPECL Output Termination

Termination for 2.5V LVPECL Outputs

Figure 5A and Figure 5B show examples of termination for 2.5V LVPECL driver. These terminations are equivalent to terminating 50Ω to V_{CC} – 2V. For V_{CC} = 2.5V, the V_{CC} – 2V is very close to ground

level. The R3 in Figure 5B can be eliminated and the termination is shown in *Figure 5C*.

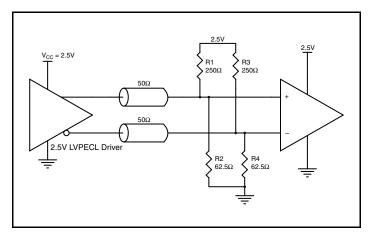


Figure 5A. 2.5V LVPECL Driver Termination Example

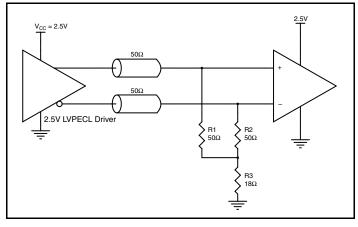


Figure 5B. 2.5V LVPECL Driver Termination Example

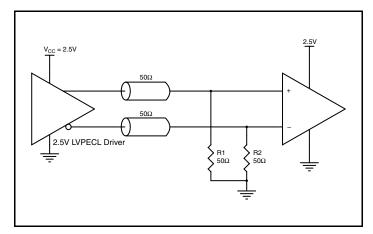


Figure 5C. 2.5V LVPECL Driver Termination Example

Schematic Example

This application note provides general design guide using ICS853S014I LVPECL buffer. *Figure 6* shows a schematic example of the ICS853S014I LVPECL clock buffer. In this example, the input

is driven by an LVPECL driver. CLK_SEL is set at logic high to select PCLK1, nPCLK1 input.

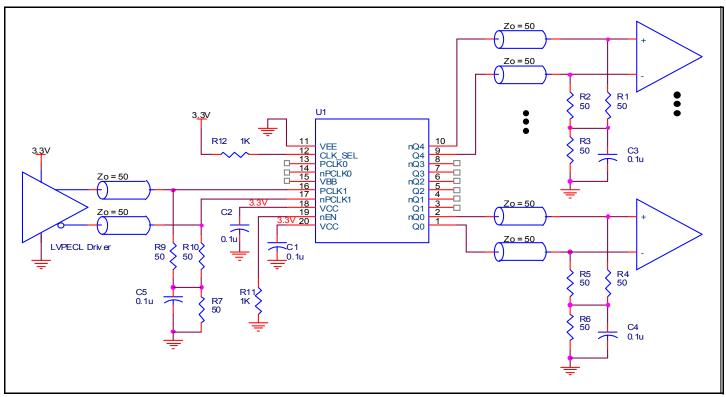


Figure 6. ICS853S014I Example LVPECL Clock Output Buffer Schematic

VFQFN EPAD Thermal Release Path

In order to maximize both the removal of heat from the package and the electrical performance, a land pattern must be incorporated on the Printed Circuit Board (PCB) within the footprint of the package corresponding to the exposed metal pad or exposed heat slug on the package, as shown in *Figure 7*. The solderable area on the PCB, as defined by the solder mask, should be at least the same size/shape as the exposed pad/slug area on the package to maximize the thermal/electrical performance. Sufficient clearance should be designed on the PCB between the outer edges of the land pattern and the inner edges of pad pattern for the leads to avoid any shorts.

While the land pattern on the PCB provides a means of heat transfer and electrical grounding from the package to the board through a solder joint, thermal vias are necessary to effectively conduct from the surface of the PCB to the ground plane(s). The land pattern must be connected to ground through these vias. The vias act as "heat pipes". The number of vias (i.e. "heat pipes") are application specific.

and dependent upon the package power dissipation as well as electrical conductivity requirements. Thus, thermal and electrical analysis and/or testing are recommended to determine the minimum number needed. Maximum thermal and electrical performance is achieved when an array of vias is incorporated in the land pattern. It is recommended to use as many vias connected to ground as possible. It is also recommended that the via diameter should be 12 to 13mils (0.30 to 0.33mm) with 1oz copper via barrel plating. This is desirable to avoid any solder wicking inside the via during the soldering process which may result in voids in solder between the exposed pad/slug and the thermal land. Precautions should be taken to eliminate any solder voids between the exposed heat slug and the land pattern. Note: These recommendations are to be used as a guideline only. For further information, please refer to the Application Note on the Surface Mount Assembly of Amkor's Thermally/ Electrically Enhance Leadframe Base Package, Amkor Technology.

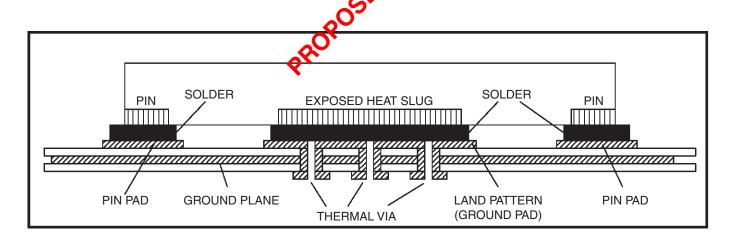


Figure 7. P.C. Assembly for Exposed Pad Thermal Release Path – Side View (drawing not to scale)

Power Considerations

This section provides information on power dissipation and junction temperature for the ICS853S014I. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the ICS853S014I is the sum of the core power plus the power dissipated in the load(s).

The following is the power dissipation for $V_{CC} = 3.8V$, which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)_{MAX} = V_{CC MAX} * I_{EE MAX} = 3.8V * 68mA = 258.4mW
- Power (outputs)_{MAX} = 30.94mW/Loaded Output pair
 If all outputs are loaded, the total power is 5 * 30.94mW = 154.7mW

Total Power_MAX (3.8V, with all outputs switching) = 258.4mW + 154.7mW = 413.1mW

2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, Tj, to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for Tj is as follows: Tj = θ_{JA} * Pd_total + T_A

Tj = Junction Temperature

 θ_{JA} = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in section 1 above)

T_A = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance θ_{JA} must be used. Assuming no air flow and a multi-layer board, the appropriate value is 92.1°C/W per Table 6A below.

Therefore, Tj for an ambient temperature of 85°C with all outputs switching is:

 $85^{\circ}\text{C} + 0.413\text{W} * 92.1^{\circ}\text{C/W} = 123^{\circ}\text{C}$. This is below the limit of 125°C .

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

Table 6A. Thermal Resistance θ_{JA} for 20 Lead TSSOP, Forced Convection

	θ_{JA} by Velocity							
Meters per Second	0	1	2.5					
Multi-Layer PCB, JEDEC Standard Test Boards	92.1°C/W	86.5°C/W	83.0°C/W					

Table 6B. Thermal Resistance θ_{JA} for 20 Lead VFQFN, Forced onvection

θ _{JA} by Velocity							
Meters per Second	000	1	2.5				
Multi-Layer PCB, JEDEC Standard Test Boards	62.2°C/W	54.4°C/W	48.8°C/W				

3. Calculations and Equations.

The purpose of this section is to calculate the power dissipation for the LVPECL output pair.

LVPECL output driver circuit and termination are shown in Figure 8.

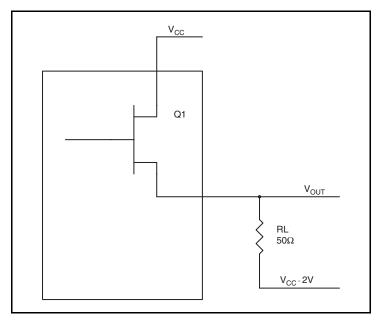


Figure 8. LVPECL Driver Circuit and Termination

To calculate worst case power dissipation into the load, use the following equations which assume a 50Ω load, and a termination voltage of V_{CC} – 2V.

- For logic high, V_{OUT} = V_{OH_MAX} = V_{CC_MAX} 0.935V
 (V_{CC_MAX} V_{OH_MAX}) = 0.935V
- For logic low, $V_{OUT} = V_{OL_MAX} = V_{CC_MAX} 1.67V$ $(V_{CC_MAX} - V_{OL_MAX}) = 1.67V$

Pd_H is power dissipation when the output drives high.

Pd_L is the power dissipation when the output drives low.

$$Pd_H = [(V_{OH_MAX} - (V_{CC_MAX} - 2V))/R_L] * (V_{CC_MAX} - V_{OH_MAX}) = [(2V - (V_{CC_MAX} - V_{OH_MAX}))/R_L] * (V_{CC_MAX} - V_{OH_MAX}) = [(2V - 0.935V)/50\Omega] * 0.935V = \textbf{19.92mW}$$

$$Pd_L = [(V_{OL_MAX} - (V_{CC_MAX} - 2V))/R_L] * (V_{CC_MAX} - V_{OL_MAX}) = [(2V - (V_{CC_MAX} - V_{OL_MAX}))/R_L] * (V_{CC_MAX} - V_{OL_MAX}) = [(2V - 1.67V)/50\Omega] * 1.67V = 11.02mW$$

Total Power Dissipation per output pair = Pd_H + Pd_L = 30.94mW

Reliability Information

Table 7A. θ_{JA} vs. Air Flow Table for a 20 Lead TSSOP

θ_{JA} by Velocity			
Meters per Second	0	1	2.5
Multi-Layer PCB, JEDEC Standard Test Boards	92.1°C/W	86.5°C/W	83.0°C/W

Table 7B. θ_{JA} vs. Air Flow Table for a 20 VFQFN TSSOP

θ _{JA} by Velocity				
Meters per Second	0	1	2.5	
Multi-Layer PCB, JEDEC Standard Test Boards	62.2°C/W	54.4°C/W	48.8°C/W	
	20			

Transistor Count

The transistor count for ICS853S014I is: 407

Pin compatible with ICS853014

Package Outline and Package Dimensions

Package Outline - G Suffix for 20 Lead TSSOP

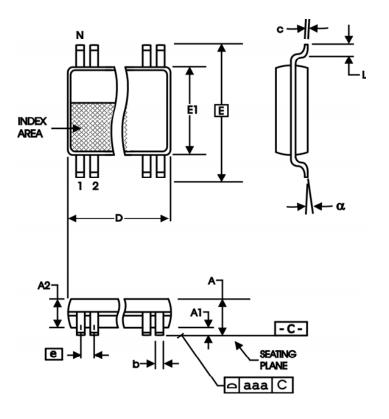


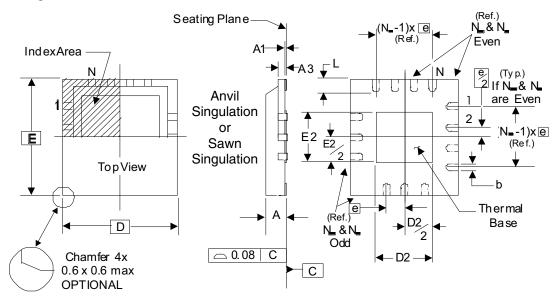
Table 8A. Package Dimensions

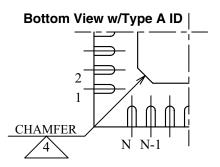
All Dimensions in Millimeters			
Symbol	Minimum	Maximum	
N	2	0	
Α		1.20	
A 1	0.05	0.15	
A2	0.80	1.05	
b	0.19	0.30	
С	0.09	0.20	
D	6.40	6.60	
E	6.40 Basic		
E1	4.30	4.50	
е	0.65 Basic		
L	0.45	0.75	
α	0°	8°	
aaa		0.10	

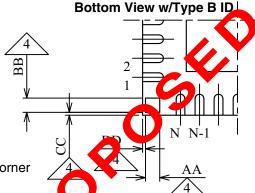
Reference Document: JEDEC Publication 95, MO-153

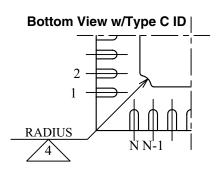
Package Outline and Package Dimensions

Package Outline - K Suffix for 20 Lead VFQFN









There are 3 methods of indicating pin 1 corner at the back of the VFQFN package are:

- 1. Type A: Chamfer on the paddle (near pin 1)
- 2. Type B: Dummy pad between pin 1 and Na
- 3. Type C: Mouse bite on the paddle (nearly)

Table 8B. Package Dimensions

	· · · J ·			
	JEDEC Variation: VGGD-1/-5 All Dimensions in Millimeters			
Symbol	Minimum	Nominal	Maximum	
N		20		
Α	0.80		1.00	
A 1	0		0.05	
А3		0.25 Ref.		
b	0.18		0.30	
N _D & N _E		5		
D&E		4.00 Basic		
D2 & E2	1.95		2.25	
е		0.50 Basic		
L	0.35		0.75	

Reference Document: JEDEC Publication 95, MO-220

NOTE: The following package mechanical drawing is a generic drawing that applies to any pin count VFQFN package. This drawing is not intended to convey the actual pin count or pin layout of this device. The pin count and pinout are shown on the front page. The package dimensions are in Table 8B.

Ordering Information

Table 9. Ordering Information

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
853S014AGILF	ICS53S014AIL	"Lead-Free" 20 Lead TSSOP	Tube	-40°C to 85°C
853S014AGILFT	ICS53S014AIL	"Lead-Free" 20 Lead TSSOP	2500 Tape & Reel	-40°C to 85°C
853S014AKILF		"Lead-Free" 20 Lead VFQFN	Tray	-40°C to 85°C
853S014AKILFT	OPOSED -	"Lead-Free" 20 Lead VFQFN	2500 Tape & Reel	-40°C to 85°C

NOTE: Parts that are ordered with an "LF" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

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Revision History Sheet

Rev	Table	Page	Description of Change	Date
В	T5	7	AC Characteristics Table - added Additive Phase Jitter max. spec.	9/1/10
С	T6B T7B T8B T9	1 4 15 16 18 19 20	Added 20 Lead VFQFN proposed pin assignment. Absolute Maximum Ratings - added 32 Lead VFQN Package Thermal Impedance. Added VFQFN EPad Thermal Release section. Added proposed 20 Lead VFQFN Thermal Resistance table. Added proposed 20 Lead VFQFN theta ja table. Added proposed 20 Lead VFQFN Package Outline and Dimensions. Ordering Information Table added proposed 20 Lead VFQFN ordering information.	10/29/10

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