



General Description

The MAX4810/MAX4811/MAX4812 integrated circuits generate high-voltage, high-frequency, unipolar or bipolar pulses from low-voltage logic inputs. These dual pulsers feature independent logic inputs, independent high-voltage pulser outputs with active clamps and independent high-voltage supply inputs.

The MAX4810/MAX4811/MAX4812 feature a 9Ω output impedance for the high-voltage outputs, and a 27Ω impedance for the active clamp. The high-voltage outputs are guaranteed to provide 1.3A output current.

All devices use three logic inputs per channel to control the positive and negative pulses and active clamp. Also included are two independent enable inputs. Disabling EN ensures the output MOSFETs are not accidentally turned on during fast power-supply ramping. This allows for faster ramp times and smaller delays between pulsing modes. A low-power shutdown mode reduces power consumption to less than 1µA. All digital inputs are CMOS compatible.

The MAX4810 includes clamp output overvoltage protection, while the MAX4811 features both pulser output and clamp output overvoltage protection. The MAX4812 does not provide overvoltage protection. See the Ordering Information/Selector Guide.

The MAX4810/MAX4811/MAX4812 are available in a 56-pin (7mm x 7mm), TQFN exposed-pad package and are specified over the 0°C to +70°C commercial temperature range.

Applications

Ultrasound Medical Flaw Detection **Imaging** Piezoelectric Drivers Cleaning Equipment Test Instruments

Ordering Information/ **Selector Guide**

PART	PROTECTED OUTPUTS	OUTPUT CURRENT (A)	PIN- PACKAGE
MAX4810CTN+	OCP_, OCN_	1.3	56 TQFN-EP**
MAX4811CTN+	OCP_, OCN_, OP_, ON_	1.3	56 TQFN-EP**
MAX4812CTN+*	None	1.3	56 TQFN-EP**

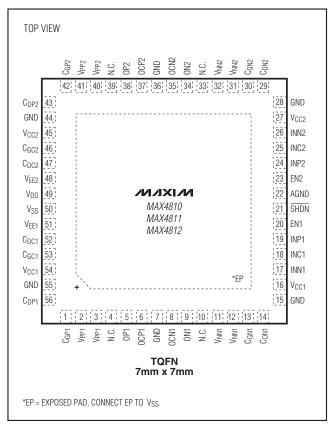
Note: All devices are specified over the 0°C to +70°C operating temperature range.

- +Denotes a lead-free/RoHS-compliant package.
- *Future product—contact factory for availability.
- **EP = Exposed pad.

Features

- ♦ Highly Integrated, High-Voltage, High-Frequency Unipolar/Bipolar Pulser
- ♦ 9Ω Output Impedance and 1.3A (min) Output Current
- ♦ 27Ω Active Clamp
- **♦ Pulser and Clamp Overvoltage Protection** (MAX4810/MAX4811)
- ♦ 0 to +220V Unipolar or ±110V Bipolar Outputs
- ♦ Matched Rise/Fall Times and Matched **Propagation Delays**
- **♦ CMOS-Compatible Logic Inputs**
- ♦ 56-Pin, 7mm x 7mm, TQFN Package

Pin Configuration



Warning: The MAX4810/MAX4811/MAX4812 are designed to operate with high voltages. Exercise caution.

ABSOLUTE MAXIMUM RATINGS

(Voltages referenced to GND.)
V _{DD} Logic Supply Voltage0.3V to +6V
V _{CC} Output Driver Positive Supply Voltage0.3V to +15V
VEE_Output Driver Negative Supply Voltage15V to +0.3V
VPP_High Positive Supply Voltage0.3V to +230V
V _{NN} _ High Negative Supply Voltage230V to +0.3V
Vss Voltage(Vpp 250V) to V _{NN}
V _{PP1} - V _{NN1} , V _{PP2} - V _{NN2} Supply Voltage0.6V to +250V
INP_, INN_, INC_, EN_, SHDN Logic Input0.3V to V _{DD} + 0.3V
Op_, Ocp_, Oln_, On(-0.3V + Vnn_) to (-0.3V to Vpp_
C _{GN} _ Voltage(-0.3V + V _{NN} _) to (+15V + V _{NN} _
C _{GP} _ Voltage(+0.3V + V _{PP} _) to (-15V + V _{PP} _
C _{GC} _Voltage15V to +15V

CDC CDP , CDN Voltage	0.3V to V _C C
Peak Current per Output Cha	annel3.0Ā
Continuous Power Dissipatio	$n (T_A = +70^{\circ}C) (Note 1)$
56-Pin TQFN (derate 40m)	N/°C above +70°C)3200mW
Thermal Resistance (Note 2)	
θJA	25°C/W
	0.8°C/W
Operating Temperature Rang	ge0°C to +70°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering	g, 10s)+300°C

- **Note 1:** This specification is based on the thermal characteristic of the package, the maximum junction temperature, and the setup described by JEDEC 51. The maximum power dissipation for the MAX4810/MAX4811/MAX4812 might be limited by the thermal protection included in the device.
- **Note 2:** Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to www.maxim-ic.com/thermal-tutorial.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(V_{DD} = +2.7V \text{ to } +6V, V_{CC} = +4.75V \text{ to } +12.6V, V_{EE} = -12.6V \text{ to } -4.75V, V_{NN} = -200V \text{ to } 0, V_{PP} = 0 \text{ to } (V_{NN} + 200V), V_{SS} \le \text{the lower of } V_{NN1} \text{ or } V_{NN2}, T_A = T_J = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } T_A = +25^{\circ}C.) \text{ (Note 3) (See Figures 8, 9, and 10.)}$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS			
POWER SUPPLY (VDD, VCC_, VEE_, VPP_, VNN_)									
Logic Supply Voltage	V_{DD}		+2.7	+3	+6	٧			
Positive Drive Supply Voltage	Vcc_		+4.75	+12	+12.6	V			
Negative Drive Supply Voltage	V _{EE} _		-12.6	-12	-4.75	٧			
High-Side Supply Voltage	V _{PP} _		0		V _{NN} _ + 220	٧			
Low-Side Supply Voltage	V _{NN} _		-200		0	V			
V _{PP} V _{NN} _ Supply Voltage			0		+220	V			
SUPPLY CURRENT (Single Cha	annel)								
		$V_{INN}/V_{INP} = 0$, $V_{\overline{SHDN}} = 0$			1				
V _{DD} Supply Current	I _{DD}	$V_{EN} = V_{DD}$, $V_{\overline{SHDN}} = V_{DD}$, $V_{INC} = 0$ or V_{DD} , $V_{INN} = V_{\overline{INP}}$, $f = 5MHz$		100	200	μΑ			
		V _{SHDN} = 0, CH1 and CH2			1	^			
		V _{EN_} = V _{DD} , V _{SHDN} = V _{DD} , CH1 and CH2		130	200	μΑ			
V _{CC} _ Supply Current	Icc_	$\begin{split} &V_{EN_} = V_{DD}, V_{\overline{SHDN}} = V_{DD}, V_{INC_} = 0 \text{ or } V_{DD}, \\ &V_{INN_} = V_{\overline{INP_}}, f = 5MHz, V_{CC_} = 5V, V_{DD} = 3V, \\ &\text{only one channel switching} \end{split}$		15		mA			
		$V_{EN_} = V_{DD}, V_{\overline{SHDN}} = V_{DD}, V_{INC_} = 0 \text{ or } V_{DD}, V_{INN_} = V_{\overline{INP_}}, f = 5MHz, V_{CC_} = 12V, V_{DD} = 3V, only one channel switching$		36		IIIA			

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DD} = +2.7 \text{V to } +6 \text{V}, V_{CC} = +4.75 \text{V to } +12.6 \text{V}, V_{EE} = -12.6 \text{V to } -4.75 \text{V}, V_{NN} = -200 \text{V to } 0, V_{PP} = 0 \text{ to } (V_{NN} + 200 \text{V}), V_{SS} \leq \text{the lower of } V_{NN1} \text{ or } V_{NN2}, T_A = T_J = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } T_A = +25 ^{\circ}\text{C.}) \text{ (Note 3) (See Figures 8, 9, and 10.)}$

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
		VSHDN = 0, CH1 and CH2			25		
		$V_{EN} = V_{DD}$, $V_{\overline{SHDN}} = V_{DD}$, CH1 and CH2			1	ļ	
V _{EE} _ Supply Current	I _{EE} _	$V_{EN} = V_{DD}$, $V_{\overline{SHDN}} = V_{DD}$, $V_{INC} = 0$ or V_{DD} , $V_{INN} = V_{\overline{INP}}$, $f = 5MHz$, $V_{EE} = -5V$, only one channel switching			200	μΑ	
		$V_{EN_} = V_{DD}$, $V_{\overline{SHDN}} = V_{DD}$, $V_{INC_} = 0$ or V_{DD} , $V_{INN_} = V_{\overline{INP_}}$, $f = 5MHz$, $V_{EE_} = -12V$, only one channel switching			200		
		VSHDN = 0, CH1 and CH2			1		
		V _{EN} _ = V _{DD} , V _{SHDN} = V _{DD} , CH1 and CH2		90	160	μΑ	
V _{PP_} Supply Current	I _{PP} _	VEN_ = VDD, VSHDN = VDD, VINC_ = 0 or VDD, VINN_ = VINP_, f = 5MHz, VPP_ = +5V, VNN_ = -5V, no load, only one channel switching		9			
		VEN_ = VDD, VSHDN = VDD, VINC_ = 0 or VDD, VPP_ = +80V, VNN_ = -80V, pulse repetition frequency = 10kHz, f = 10MHz, 4 periods, no load, only one channel switching		0.6		mA	
		VSHDN = 0, CH1 and CH2			1		
		V _{EN_} = V _{DD} , V SHDN = V _{DD} , CH1 and CH2		40	80	μA	
V _{NN} Supply Current	I _{NN} _	VEN_ = VDD, VSHDN = VDD, VINC_ = 0 or VDD, VINN_ = VINP_, f = 5MHz, V _{NN} _ = -5V, V _{PP} _ = +5V, no load, only one channel switching					
		VEN_ = VDD, VSHDN = VDD, VINC_ = 0 or VDD, VPP_ = +80V, VNN_ = -80V, pulse repetition frequency = 10kHz, f = 10MHz, 4 periods, no load, only one channel switching		0.6		mA	
LOGIC INPUTS (EN_, SHDN, IN	NN_, INP_, II	NC_)					
Low-Level Input Voltage	VIL				0.25 x V _{DD}	V	
High-Level Input Voltage	VIH		0.75 x V _{DD}			V	
Logic-Input Capacitance	CIN			5		рF	
Logic-Input Leakage	I _{IN}	$V_{IN} = 0$ or V_{DD}			±1	μΑ	
OUTPUT (OUT_)							
OUT_ Output-Voltage Range		No load at OUT_	V _{NN} _		V _{PP} _		
	V _{OUT} _	Unprotected outputs (see the <i>Ordering Information/Selector Guide</i>), 100mA load	V _{NN} _ + 1.5		V _{PP} 1.5	V	
		Protected outputs (see the <i>Ordering Information/Selector Guide</i>), 100mA load	V _{NN} _ + 2.5		V _{PP} 2.5		
Low-Side Small-Signal Output		I _{OP} _ = -100mA, V _{CC} _ = +12V ±5%, DC-coupled		9	17		
Impedance	Rols	I _{OP} _ = -100mA, V _{CC} _ = +5V ±5%, DC-coupled		9.5	18	Ω	

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DD} = +2.7 \text{V to } +6 \text{V}, V_{CC} = +4.75 \text{V to } +12.6 \text{V}, V_{EE} = -12.6 \text{V to } -4.75 \text{V}, V_{NN} = -200 \text{V to } 0, V_{PP} = 0 \text{ to } (V_{NN} + 200 \text{V}), V_{SS} \leq \text{the lower of } V_{NN1} \text{ or } V_{NN2}, T_A = T_J = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } T_A = +25 ^{\circ}\text{C.}) \text{ (Note 3) (See Figures 8, 9, and 10.)}$

PARAMETER	SYMBOL	YMBOL CONDITIONS			TYP	MAX	UNITS
High-Side Small-Signal Output		I _{OP} _ = -100mA, V _{CC} _ = +12V ±5%, DC-coupled			10.5	17	
Impedance	Rohs	I_{OP} = -100mA, V_{CC} = +5V ±5%, DC-coupled			12	18	Ω
Low-Side Output Current	loL	V _{CC} = +12V ±5%, V _{OUT} - V _{NN} =		1.3			А
High-Side Output Current	I _{OH}	V _{CC} = +12V ±5%, V _{OUT} - V _{PP} =	: 100V	1.3			А
		OP_, ON_, OCP_ and OCN_	MAX4810		45		
Off-Output Capacitance	C _{O(OFF)}	connected together, VPP_ = +100V, V _{NN} _ = -100V	MAX4811		75		рF
Off-Output Leakage Current	I _{LK}	V _{NN} _ = -100V, V _{PP} _ = 100V, EN_ = OUT = -100V to +100V	: 0,	-1		+1	μΑ
Low-Side Signal-Clamp Output	-	I _{OCN} = -100mA, DC-coupled, V _{CC} V _{EE} = -V _{CC}	_ = +12V ±5%,		22	50	
Impedance	Rcls	I _{OCN_} = -100mA, DC-coupled, V _{CC} V _{EE_} = -V _{CC_}	C_ = +5V ±5%,		24	65	Ω
High-Side Signal-Clamp Output		I _{OCP} _ = -100mA, DC-coupled, V _{CC} _ VEE_ = -V _{CC} _	_ = +12V ±5%,		28	50	
Impedance	R _{CHS}	IOCP_ = -100mA, DC-coupled, VCC VEE_ = -VCC_		38	65	Ω	
Low-Side Gate Short	_	V _{CC} _ = +12V ±5%, V _{EE} _ = -V _{CC} _, EN_ = 0			100	Ω	
Impedance	R _{LSH}	V _{CC} _ = +12V ±5%, V _{EE} _ = -V _{CC} _, EN_ = V _{DD}	5	7.5	10	kΩ	
High-Side Gate Short		V _{CC} _ = +12V ±5%, V _{EE} _ = -V _{CC} _, EN_ = 0	I _{CGN} = 10mA,			100	Ω
Impedance	R _{HSH}	V _{CC} _ = +12V ±5%, V _{EE} _ = -V _{CC} _, EN_ = V _{DD}	ICGN = 10mA,	5	7.5	10	kΩ
THERMAL SHUTDOWN							
Thermal Shutdown	T _{SHDN}	Junction temperature rising			150		°C
Thermal-Shutdown Hysteresis					20		°C
DYNAMIC CHARACTERISTICS	$(R_L = 100\Omega$	$C_{\rm L}$ = 100pF, unless otherwise no	oted)				
Logic Input to Output Rise Propagation Delay	tpLH	V _{CC} _ = +12V, V _{PP} _ = +5V, V _{NN} _ =	-5V, Figure 4		15		ns
Logic Input to Output Fall Propagation Delay	t _{PHL}	V _{CC} _ = +12V, V _{PP} _ = +5V, V _{NN} _ = -5V, Figure 4			15		ns
Logic Input to Output Rise Propagation Delay	tpoh	V _{CC} _ = +12V, V _{PP} _ = +5V, V _{NN} _ = -5V, Figure 4			15		ns
Logic Input to Output Fall Propagation Delay	t _{POL}	V _{CC} _ = +12V, V _{PP} _ = +5V, V _{NN} _ = -5V, Figure 4			15		ns
Logic Input to Output-Rise Propagation Delay Clamp	t _{PLO}	V _{CC} _ = +12V, V _{PP} _ = +5V, V _{NN} _ =	-5V, Figure 4		15		ns

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ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DD} = +2.7 \text{V to } +6 \text{V}, V_{CC} = +4.75 \text{V to } +12.6 \text{V}, V_{EE} = -12.6 \text{V to } -4.75 \text{V}, V_{NN} = -200 \text{V to } 0, V_{PP} = 0 \text{ to } (V_{NN} + 200 \text{V}), V_{SS} \le \text{the lower of } V_{NN1} \text{ or } V_{NN2}, T_A = T_J = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted.}$ Typical values are at $T_A = +25 ^{\circ}\text{C}$.) (Note 3) (See Figures 8, 9, and 10.)

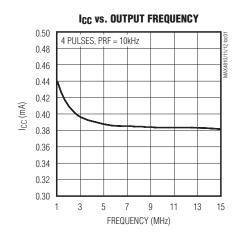
PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Logic Input to Output-Fall Propagation Delay Clamp	tpHO	V _{CC} _ = +12V, V _{PP} _ = +5V, V _{NN} _ = -5V, Figure 4		15		ns
OUT_ Rise Time (GND to V _{PP} _)	tROP	VPP_ = +100V, V _{NN} _ = -100V, V _{CC} _ = +12V ± 5%, V _{EE} _ = -V _{CC} _, Figure 4		9	20	ns
OUT_ Rise Time (V _{NN} _ to GND)	t _{RN0}	VPP_ = +100V, V _{NN} _ = -100V, V _{CC} _ = +12V ± 5%, V _{EE} _ = -V _{CC} _, Figure 4		17	35	ns
OUT_ Rise Time (V _{NN} _ to V _{PP} _)	t _{RNP}	V _{PP} _ = +100V, V _{NN} _ = -100V, V _{CC} _ = +12V ± 5%, V _{EE} _ = -V _{CC} _, Figure 4		10.5	35	ns
OUT_ Fall Time (GND to V _{NN} _)	tFON	V _{PP} = +100V, V _{NN} = -100V, V _{CC} = +12V ± 5%, V _{EE} = -V _{CC} , Figure 4		9	20	ns
OUT_ Fall Time (VPP_ to GND)	t _{FP0}	V _{PP} = +100V, V _{NN} = -100V, V _{CC} = +12V ± 5%, V _{EE} = -V _{CC} , Figure 4		17	35	ns
OUT_ Fall Time (V _{PP} _ to V _{NN} _)	t _{FPN}	V _{PP} _ = +100V, V _{NN} _ = -100V, V _{CC} _ = +12V ± 5%, V _{EE} _ = -V _{CC} _, Figure 4		10.5	35	ns
OUT Enable Time from EN	t _{EN}	V _{CC} = +12V ± 5%, V _{EE} = -V _{CC}			100	ns
(Figure 5)		V _{CC} _ = +5V ± 5%, V _{EE} _ = -V _{CC} _			150	
OUT Disable Time from EN (Figure 5)	t _{DI}	V _{CC} = +12V ± 5%, V _{EE} = -V _{CC}			100	ns
		V _{CC} = +5V ± 5%, V _{EE} = -V _{CC} V _{CC} = +12V ± 5%, V _{EE} = -V _{CC}			150 150	
Clamp Enable Time from INC (Figure 6)	ten-cl	VCC_ = +5V ± 5%, VEE_ = -VCC_			180	ns
Clamp Disable Time from INC		Vcc_ = +12V ± 5%, VEE_ = -Vcc_			150	
(Figure 6)	tDI-CL	Vcc = +5V ± 5%, VEE = -Vcc			150	ns
Short Enable Time from EN		VPP_ = 12V, V _{NN} _ = 0, V _{CC} _ = +12V ± 5%, VEE_ = -V _{CC} _			1000	
(Figure 7)	tEN_SH	VPP_ = 5V, V _{NN} _ = 0, V _{CC} _ = +5V ± 5%, VEE_ = -V _{CC} _			1000	ns
Short Disable Time from EN		VPP_ = 12V, V _{NN} _ = 0, V _{CC} _ = +12V ± 5%, VEE_ = -V _{CC} _			250	
(Figure 7)	t _{DI_SH}	VPP_ = 5V, V _{NN} _ = 0, V _{CC} _ = +5V ± 5%, VEE_ = -V _{CC} _	250		ns	
INP_to INN_ Overlap Tolerance				131		ns
Crosstalk		VPP_ = VCC_ = +5V, V _{NN} _ = V _{EE} _ = -5V, f = 5MHz		69		dB
2nd Harmonic Distortion	2HD	V _{PP} _ = V _{NN} _ = 100V, f _{OUT} = 5MHz, V _{CC} _ = 12V		-48		dB
RMS Output Jitter	tJ	V _{CC} _ = 12V		9		ps

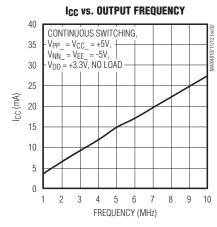
Note 3: Specifications are guaranteed for the stated global conditions, unless otherwise noted and are 100% production tested at $T_A = +25^{\circ}C$ and $T_A = +70^{\circ}C$. Specifications at $T_A = 0^{\circ}C$ are guaranteed by design.

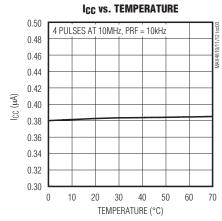
Note 4: 100% production tested at $T_A = +25$ °C. Specifications over temperature are guaranteed by design.

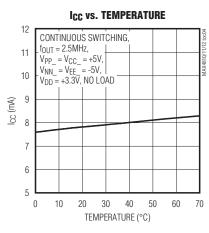
Typical Operating Characteristics

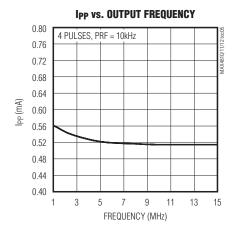
 $(V_{DD} = +3.3V, V_{CC_} = +12V, V_{EE_} = -12V, V_{SS} = -100V, V_{PP_} = +100V, V_{NN_} = -100V, f_{OUT} = 5MHz, T_{A} = +25^{\circ}C, unless otherwise noted.)$

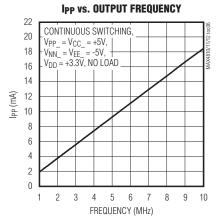


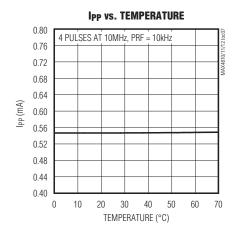


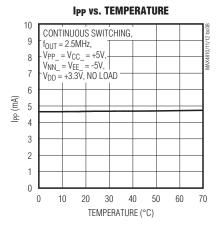


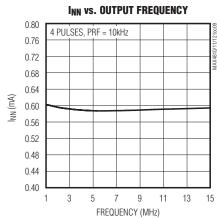






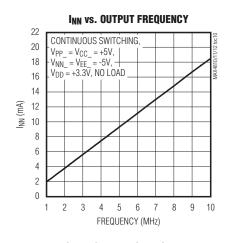


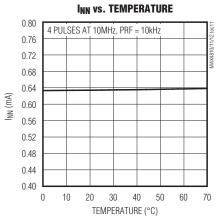


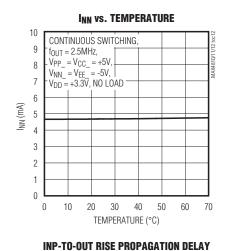


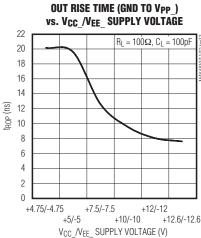
Typical Operating Characteristics (continued)

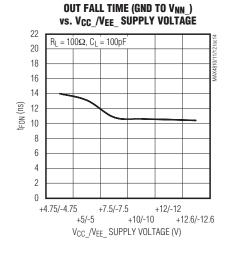
 $(V_{DD} = +3.3V, V_{CC} = +12V, V_{EE} = -12V, V_{SS} = -100V, V_{PP} = +100V, V_{NN} = -100V, f_{OUT} = 5MHz, T_A = +25^{\circ}C$, unless otherwise noted.)

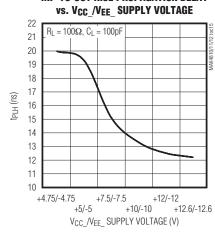


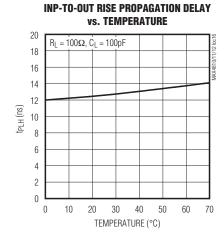


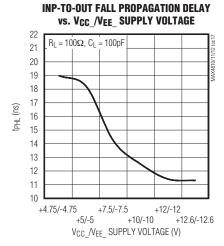


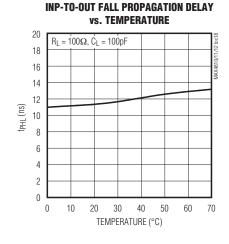












Pin Description

PIN	NAME	FUNCTION
1	C _{GP1}	Channel 1 High-Side Gate Input. Connect a 1nF to 10nF capacitor between C _{DP1} and C _{GP1} as close as possible to the device.
2,3	VPP1	Channel 1 High-Side Positive Supply Voltage Input. Bypass V _{PP1} to GND with a 0.1µF as close as possible to the device. See the <i>Power Supplies and Bypassing</i> section. Depending on the output, additional bypassing may be required.
4, 10, 33, 39	N.C.	No Connection. Not connected internally.
5	OP1	Channel 1 High-Side Drain Output
6	OCP1	Channel 1 High-Side Clamp Output
7, 15, 28, 36, 44, 55	GND	Ground
8	OCN1	Channel 1 Low-Side Clamp Output
9	ON1	Channel 1 Low-Side Drain Output
11, 12	V _{NN1}	Channel 1 High-Side Negative Supply Voltage Input. Bypass V _{NN1} to GND with a 0.1µF as close as possible to the device. See the <i>Power Supplies and Bypassing</i> section. Depending on the output, additional bypassing may be required.
13	C _{GN1}	Channel 1 Low-Side Gate Input. Connect a 1nF to 10nF capacitor between C _{DN1} and C _{GN1} as close as possible to the device.
14	C _{DN1}	Channel 1 Low-Side Driver Output. Connect a 1nF to 10nF capacitor between C _{DN1} and C _{GN1} as close as possible to the device.
16, 54	V _{CC1}	Channel 1 Gate-Drive Supply Voltage Input. Bypass V _{CC1} to GND with a 0.1µF as close as possible to the device. See the <i>Power Supplies and Bypassing</i> section. Depending on the output, additional bypassing may be required.
17	INN1	Channel 1 Low-Side Logic Input (Table 1)
18	INC1	Channel 1 Clamp Logic Input. Clamps OCP1 and OCN1 are turned on when INC1 is high and when INP1 and INN1 are low (see Table 1).
19	INP1	Channel 1 High-Side Logic Input (Table 1)
20	EN1	Channel 1 Enable Logic Input. Drive EN1 high to enable OP1 and ON1. Pull EN1 low to turn on the gate-source short circuit (see Table 1).
21	SHDN	Shutdown Logic Input (Table 1)
22	AGND	Analog Ground. Must be connected to common GND.
23	EN2	Channel 2 Enable Logic Input. Drive EN2 high to enable OP2 and ON2. Pull EN2 low to turn on the gate-source short circuit. See Table 1.
24	INP2	Channel 2 High-Side Logic Input (Table 1)
25	INC2	Channel 2 Clamp Logic Input. Clamps OCP2 and OCN2 are turned on when INC2 is high and when INP2 and INN2 are low. See Table 1.
26	INN2	Channel 2 Low-Side Logic Input (Table 1)
27, 45	V _{CC2}	Channel 2 Gate-Drive Supply Voltage Input. Bypass V _{CC2} to GND with a 0.1µF as close as possible to the device. See the <i>Power Supplies and Bypassing</i> section. Depending on the output, additional bypassing may be required.
29	C _{DN2}	Channel 2 Low-Side Driver Output. Connect a 1nF to 10nF capacitor between C _{DN2} and C _{GN2} as close as possible to the device.
30	C _{GN2}	Channel 2 Low-Side Gate Input. Connect a 1nF to 10nF capacitor between C _{DN2} and C _{GN2} as close as possible to the device.

Pin Description (continued)

PIN	NAME	FUNCTION
31, 32	V _{NN2}	Channel 2 High-Side Negative Supply Voltage Input. Bypass V _{NN2} to GND with a 0.1µF as close as possible to the device. See the <i>Power Supplies and Bypassing</i> section. Depending on the output, additional bypassing may be required.
34	ON2	Channel 2 Low-Side Drain Output
35	OCN2	Channel 2 Low-Side Clamp Output
37	OCP2	Channel 2 High-Side Clamp Output
38	OP2	Channel 2 High-Side Drain Output
40, 41	V _{PP2}	Channel 2 High-Side Supply Voltage Input. Bypass V _{PP2} to GND with a 0.1µF as close as possible to the device. See the <i>Power Supplies and Bypassing</i> section. Depending on the output, additional bypassing may be required.
42	C _{GP2}	Channel 2 High-Side Gate Input. Connect a 1nF to 10nF capacitor between C _{DP2} and C _{GP2} as close as possible to the device.
43	C _{DP2}	Channel 2 High-Side Driver Output. Connect a 1nF to 10nF capacitor between C _{DP2} and C _{GP2} as close as possible to the device.
46	C _{GC2}	Channel 2 High-Side Clamp Gate Input. Connect a 1nF to 10nF capacitor between C _{DC2} and C _{GC2} as close as possible to the device.
47	C _{DC2}	Channel 2 High-Side Clamp Driver Output. Connect a 1nF to 10nF capacitor between CDC2 and CGC2 as close as possible to the device.
48	V _{EE2}	Channel 2 Negative Supply Input. $ V_{EE2} \le V_{CC2}$. Gate Drive Supply Voltage for the OCP clamp. Bypass V_{EE2} to GND with a $0.1\mu F$ as close as possible to the device. See the <i>Power Supplies and Bypassing</i> section. Depending on the output, additional bypassing may be required.
49	V _{DD}	Logic Supply Voltage Input. Bypass V _{DD} to GND with a 0.1µF as close as possible to the device. See the <i>Power Supplies and Bypassing</i> section. Depending on the output, additional bypassing may be required.
50	V _{SS}	Substrate Voltage. Connect V_{SS} to a voltage equal to or more negative than the more negative of V_{NN1} or V_{NN2} .
51	V _{EE1}	Channel 1 Negative Supply Input. $ V_{EE1} \le V_{CC1}$. Gate Drive Supply Voltage for the OCP clamp. Bypass V_{EE1} to GND with a $0.1\mu F$ as close as possible to the device. See the <i>Power Supplies and Bypassing</i> section. Depending on the output, additional bypassing may be required.
52	C _{DC1}	Channel 1 High-Side Clamp Driver Output. Connect a 1nF to 10nF capacitor between C _{DC1} and C _{GC1} as close as possible to the device.
53	C _{GC1}	Channel 1 High-Side Clamp Gate Input. Connect a 1nF to 10nF capacitor between C _{DC1} and C _{GC1} as close as possible to the device.
56	C _{DP1}	Channel 1 High-Side Driver Output. Connect a 1nF to 10nF capacitor between C _{DP1} and C _{GP1} as close as possible to the device.
_	EP	Exposed Pad. EP must be connected to VSS. Do not use EP as the only VSS connection for the device.

Detailed Description

The MAX4810/MAX4811/MAX4812 are dual high-voltage, high-speed pulsers that can be independently configured for either unipolar or bipolar pulse outputs. These devices have independent logic inputs for full pulse control and independent active clamps. The clamp input, INC_, can be set high to activate the

clamp automatically when the device is not pulsing to the positive or negative high-voltage supplies.

Logic Inputs (INP_, INN_, INC_, EN_, SHDN)
The MAX4810/MAX4811/MAX4812 have a total of nine logic input signals. SHDN controls power-up and power-down of the device. There are two sets of INP_, INN_, INC_, and EN_ signals: one for each channel. INP_

Table 1. Truth Table

	II	NPUTS				OUTPUTS		
SDHN	EN_	INP_	INN_	INC_	OP_	ON_	OCP_, OCN_	STATE
0	Х	Х	Х	0	High impedance	High impedance	High impedance	Powered down, INP_/INN_ disabled, gate-source short disabled
0	Х	Х	Х	1	High impedance	High impedance	GND	Powered down, INP_/INN_ disabled, gate-source short disabled
1	0	Х	Х	0	High impedance	High impedance	High impedance	Powered up, INP_/INN_ disabled, gate-source short enabled
1	0	Х	Х	1	High impedance	High impedance	GND	Powered up, INP_/INN_ disabled, gate-source short enabled
1	1	0	0	0	High impedance	High impedance	High impedance	Powered up, all inputs enabled, gate-source short disabled
1	1	0	0	1	High impedance	High impedance	GND	Powered up, all inputs enabled, gate-source short disabled
1	1	0	1	Х	High impedance	V _{NN} _	High impedance	Powered up, all inputs enabled, gate-source short disabled
1	1	1	0	Х	V _{PP} _	High impedance	High impedance	Powered up, all inputs enabled, gate-source short disabled
1	1	1	1	Х	V _{PP} _	V _{NN} _	High impedance	Not allowed (3ns maximum overlap)

X = Don't care.

controls the on and off states of the high side FET, INN_controls the on and off states of the low side FET, INC_controls the active clamp and EN_controls the gate to source short. These signals give complete control of the output stage of each driver (see Table 1 for all logic combinations).

The MAX4810/MAX4811/MAX4812 logic inputs are CMOS logic compatible and the logic level are referenced to V_{DD} for maximum flexibility. The low 5pF (typ) input capacitance of the logic inputs reduces loading and increases switching speed.

High-Voltage Output Protection (MAX4811 Only)

The high-voltage outputs of the MAX4811 feature an integrated overvoltage protection circuit that allows the user to implement multilevel pulsing by connecting the outputs of multiple pulser channels in parallel. Internal diodes in series with the ON_ and OP_ outputs prevent the body diode of the high-side and low-side FETs from switching on when a voltage greater than V_{NN_} or V_{PP_} is present on the output. See Figure 2.

Active Clamps

The MAX4810/MAX4811/MAX4812 feature an active clamp circuit to improve pulse quality and reduce 2nd harmonic output. The clamp circuit consists of an Nchannel (DC-coupled) and a P-channel (AC and DC delay coupled) high-voltage FETs that are switched on or off by the logic clamp input (INC_). The MAX4810/ MAX4811 feature protected clamp devices, allowing the clamp circuit to be used in bipolar pulsing circuits (see Figures 1 and 2). A diode in series with the OCN_ output prevents the body diode of the low-side FET from turning on when a voltage lower than GND is present. Another diode in series with the OCP_ output prevents the body diode of the high-side FET from turning on when a voltage higher than ground is present. The MAX4812 does not have diode protection on the clamp outputs. Thus, the device is suitable for use in circuits where only unipolar pulsing is required.

The user can connect the active clamp input (INC_) to a logic-high voltage and drive only the INP_ and INN_ inputs to minimize the number of signals used to drive the

10 ______ /I/XI/M

^{0 =} Logic-low.

^{1 =} Logic-high.

device. In this case, whenever both the INP_ and INN_ inputs are low and the INC_ input is high, the active clamp circuit pulls the output to GND through the OCP_ and OCN_outputs (see Table 1 for more information).

Power-Supply Ramping and Gate-Source Short Circuit

The MAX4810/MAX4811/MAX4812 include a gate-source short circuit that is controlled by the enable input (EN_). When \overline{SHDN} is high and EN is low, a 60Ω switch shorts together the gate and source of the high-side output FET. At the same time, a similar switch shorts the gate and source of the low-side output FET (Table 1). The gate-source short circuit prevents accidental turn-on of the output FETs due to the ramping voltage on VPP_ and VNN_, and allows for faster ramping rates and smaller delay times between pulsing modes.

Shutdown Mode

SHDN is common to both channel 1 and channel 2 and powers up or down the device. Drive SHDN low to power down all internal circuits (except the clamp circuits). When SHDN is low, the device is in the lowest power state (1µA) and the gate-source short circuit is disabled. The device takes 1µs (typ) to become active when SHDN is disabled.

Thermal Protection

A thermal shutdown circuit with a typical threshold of $+150^{\circ}\text{C}$ prevents damage due to excessive power dissipation. When the junction temperature exceeds T_J = $+150^{\circ}\text{C}$, all outputs are disabled. Normal operation typically resumes after the IC's junction temperature drops below $+130^{\circ}\text{C}$.

Applications Information

AC-Coupling Capacitor Selection

The value of all AC-coupling capacitors (between CDP_{-} and CGP_{-} , and between CDN_{-} and CGN_{-}) should be between 1nF to 10nF. The voltage rating of the capacitor should be at least as high as VPP_{-} . The capacitors should be placed as close as possible to the device.

Because INP_ and part of INC_ are AC-coupled to the output devices, they cannot be driven high indefinitely when the device is active.

Power Dissipation

The power dissipation of the MAX4810/MAX4811/ MAX4812 consists of three major components caused by the current consumption from V_{CC} , V_{PP} , and V_{NN} . The sum of these components (P_{VCC} , P_{VPP} and

P_{VNN}) must be kept below the maximum power-dissipation limit. See the *Typical Operating Characteristics* section for more information on typical supply currents versus switching frequencies.

The device consumes most of the supply current from V_{CC} supply to charge and discharge internal nodes such as the gate capacitance of the high-side FET (C_P) and the low-side FET (C_N). Neglecting the small quiescent supply current and a small amount of current used to charge and discharge the capacitances at the internal gate clamp FETs, the power consumption can be estimated as follows:

$$\begin{aligned} P_{VCC} &= \left[\left(C_{N} \times V_{CC} \right)^{2} \times f_{|N} \right) + \left(C_{P} \times V_{CC} \right)^{2} \times f_{|N} \right] \times \left(BRF \times BTD \right) \\ f_{|N} &= f_{|NN} + f_{|NP} \end{aligned}$$

Where f_{INN} and f_{INP} are the switching frequency of the inputs INN, INP respectively, and where BRF is the burst repitition frequency and BTD is the burst time duration. The typical value of the gate capacitances of the power FET are $C_N = 0.2nF$, $C_P = 0.4nF$.

For an output load that has a resistance of R_L and capacitance of C_L , the MAX4810/MAX4811/MAX4812 power dissipation can be estimated as follows (assume square wave output and neglect the resistance of the switches):

$$\mathsf{P}_{\mathsf{VPP}} = \left\{ \left[\left(\mathsf{C}_{\mathsf{O}} + \mathsf{C}_{\mathsf{L}} \right) \times \mathfrak{f}_{\mathsf{IN}} \times \left(\mathsf{V}_{\mathsf{PP}_{-}} - \mathsf{V}_{\mathsf{NN}_{-}} \right)^{2} \right] + \left[\frac{\mathsf{V}_{\mathsf{PP}_{-}}^{2}}{\mathsf{R}_{\mathsf{L}}} \times \frac{1}{2} \right] \times \left(\mathsf{BRF} \times \mathsf{BTD} \right) \right\}$$

where Co is the output capacitance of the device.

Power Supplies and Bypassing

The MAX4810/MAX4811/MAX4812 operate from independent supply voltage sets (only V_{DD} and V_{SS} are common to both channels). The logic input circuit operates from a +2.7V to +6V single supply (V_{DD}). The level-shift driver dual supplies, V_{CC}_/V_{EE}_ operate from $\pm 4.75V$ to $\pm 12.6V$.

The V_{PP}/V_{NN} high-side and low-side supplies are driven from a single positive supply up to +220V, from a single negative supply up to -200V, or from ±110V dual supplies. Either V_{PP} or V_{NN} can be set at 0. Bypass each supply input to ground with a 0.1 μ F capacitor as close as possible to the device.

Depending on the load of the input, additional bypassing may be needed to keep the output of V_{NN} and V_{PP} stable during output transitions. For example, with

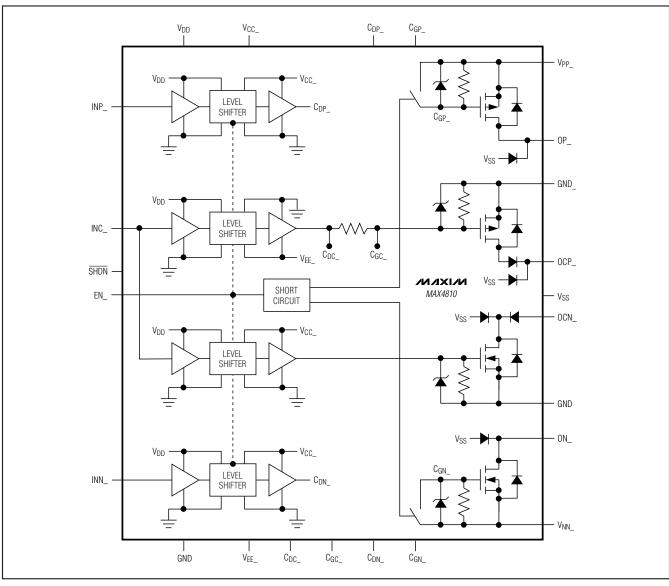


Figure 1. MAX4810 Simplified Functional Diagram for One Channel

 C_{OUT} = 100pF and R_{OUT} = 100 Ω load, additional 10µF (typ) capacitor is recommended. VSS is the substrate voltage and must be connected to a voltage equal to or more negative than the more negative voltage of V_{NN1} or $V_{NN2}.$

Exposed Pad and Layout Concerns

The MAX4810/MAX4811/MAX4812 provide an exposed pad (EP) underneath the TQFN package for improved thermal performance. EP is internally connected to Vss. Connect EP to Vss externally and do not run traces

under the package to avoid possible short circuits. To aid heat dissipation, connect EP to a similarly sized pad on the component side of the PCB. This pad should be connected through to the solder-side copper by several plated holes to a large heat spreading copper area to conduct heat away from the device.

The MAX4810/MAX4811/MAX4812 high-speed pulsers require low-inductance bypass capacitors to their supply inputs. High-speed PCB trace design practices are recommended. Pay particular attention to minimize

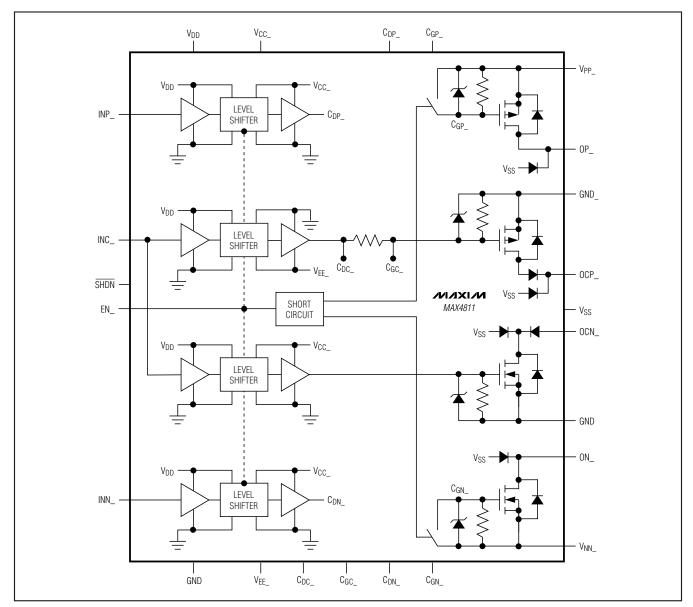


Figure 2. MAX4811 Simplified Functional Diagram for One Channel

trace lengths and use sufficient trace width to reduce inductance. Use of surface-mount components is recommended.

Supply Sequencing

Vss must be lower than or equal to the more negative voltage of V_{NN1} or V_{NN2} at all times. No other power-supply sequencing is required for the MAX4810/MAX4811/MAX4812.

Typical Application Circuits

Figures 8, 9, and 10 show typical applications for the MAX4810/MAX4811/MAX4812. Figure 8 shows the MAX4810 used in a bipolar pulsing connection. Figure 9 shows the MAX4811 in a five-level pulsing application, and Figure 10 shows the MAX4812 used in a unipolar application.

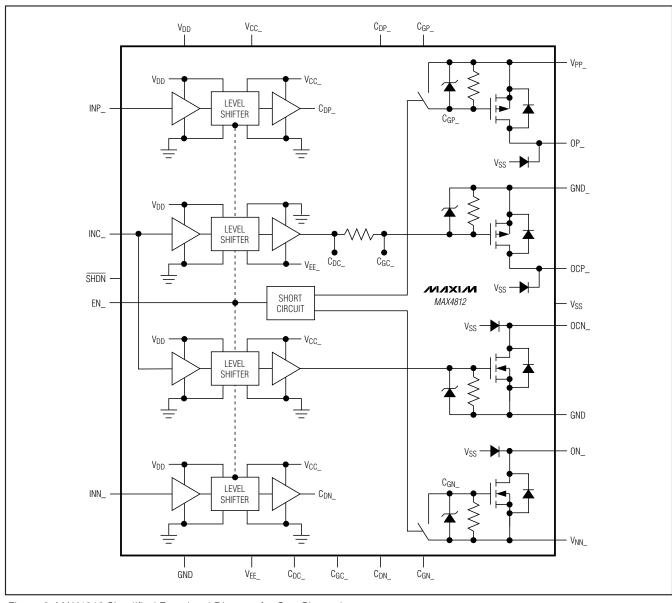


Figure 3. MAX4812 Simplified Functional Diagram for One Channel

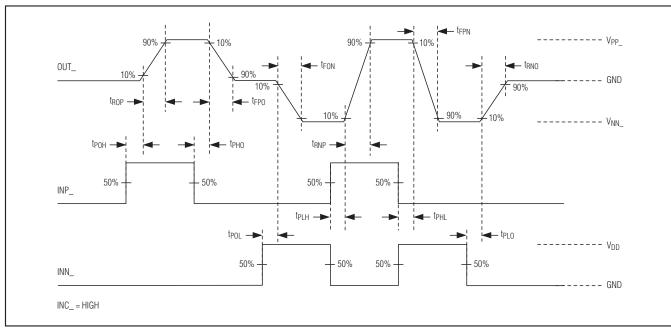


Figure 4. Detailed Timing ($R_L = 100\Omega$, $C_L = 100pF$)

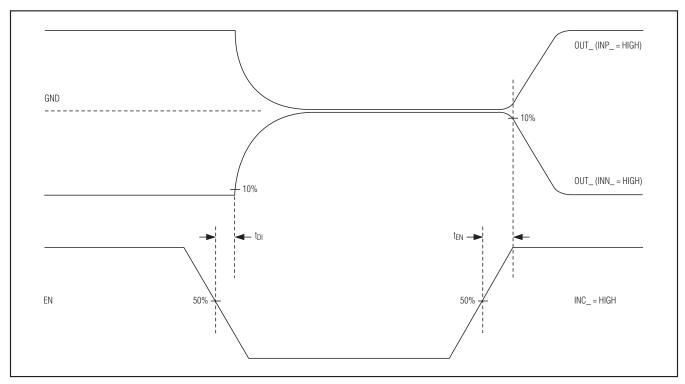


Figure 5. Enable Timing ($R_L = 100\Omega$, $C_L = 100pF$)

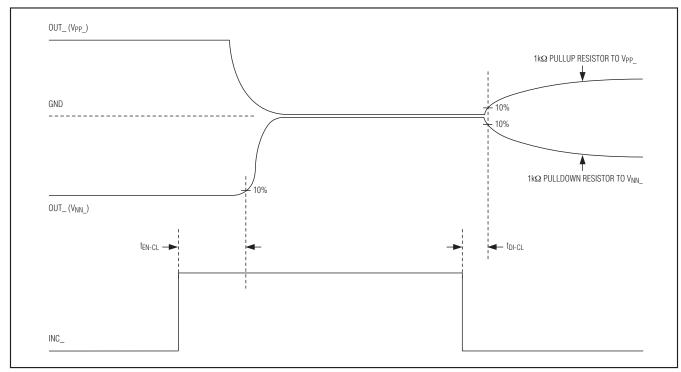


Figure 6. Active Clamp Timing

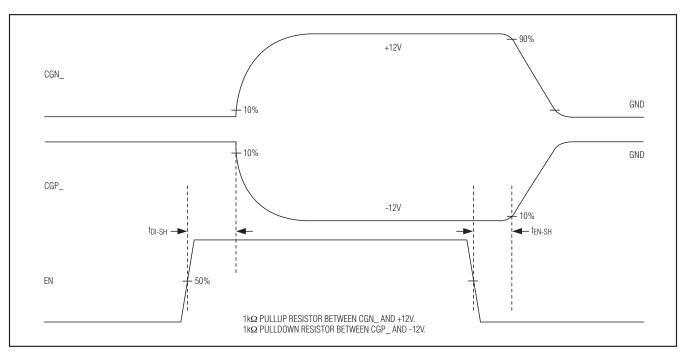


Figure 7. Short-Circuit Timing

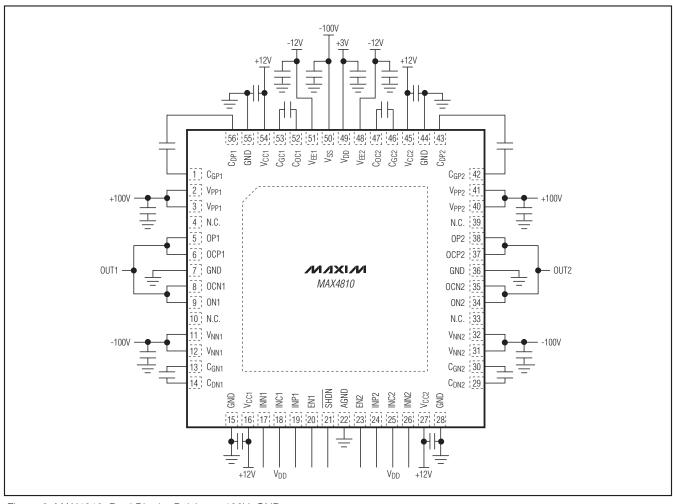


Figure 8. MAX4810: Dual Bipolar Pulsing, ±100V, GND

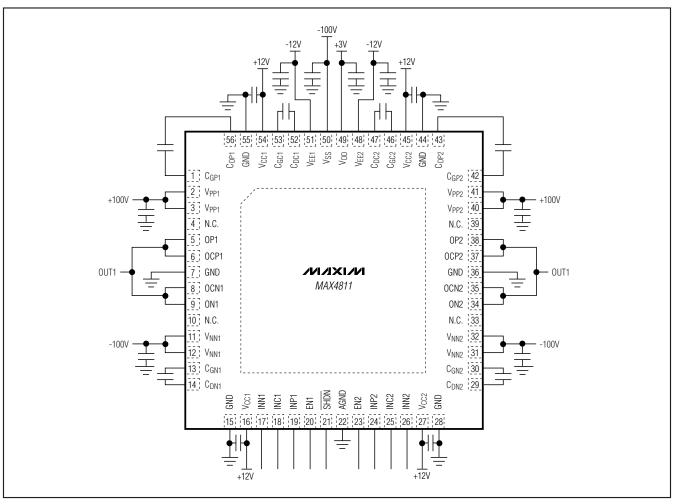


Figure 9. MAX4811: Five-Level Pulsing, ±100V, ±50V, GND

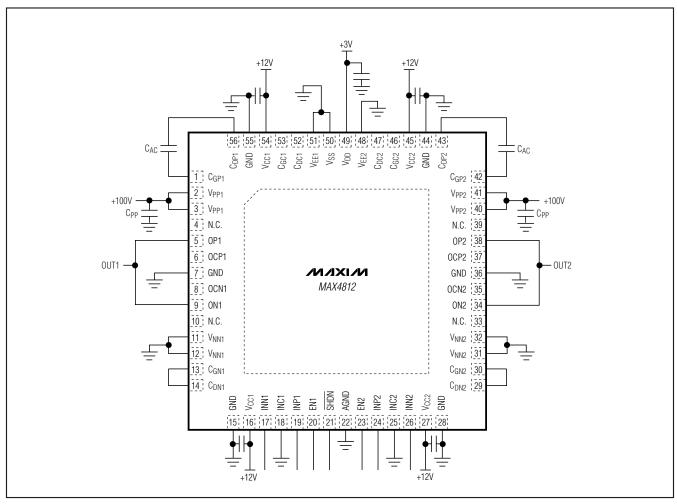


Figure 10. MAX4812: Dual Unipolar Pulsing, +100V, GND

Package Information

For the latest package outline information and land patterns, go to www.maxim-ic.com/packages.

PACKAGE TYPE	PACKAGE CODE	DOCUMENT NO.
56 TQFN	T5677-1	<u>21-0144</u>

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