

# SANYO Semiconductors **DATA SHEET**

**Monolithic Digital IC** 

# LB11948T

# PWM Constant Current Control 1-2 Phase Excitation Stepping Motor Driver

#### Overview

The LB11948T is a low saturation voltage output PWM current control bipolar drive stepping motor driver. It is optimal for use as the driver for the miniature low-voltage stepping motors used in portable electronic equipment such as portable thermal printers.

#### **Features**

- PWM current control (external excitation)
- Simultaneous on state prevention function (through current prevention)
- Thermal shutdown circuit
- Noise canceller function
- Low-power mode control pin

#### **Specifications**

#### Absolute Maximum Ratings at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
VS supply voltage	VS		-0.3 to +18	V
Logic system supply voltage	Vcc		-0.3 to +18	V
Peak output current	I <sub>O</sub> peak	tW ≤ 20μS	0.5	Α
Continuous output current	I <sub>O</sub> max		0.4	Α
Emitter output voltage	VE		1.0	V
Input voltage	V <sub>IN</sub>		-0.3 to V <sub>CC</sub>	V
Allowable power dissipation	Pd max	Mounted on the specified PCB*	1.2	W
Operating temperature	Topg		-20 to +85	°C
Storage temperature	Tstg		-40 to +150	°C

Note \*: Specified PCB: 114.3×76.1×1.6mm

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#### Recommended Operating Conditions at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
VS supply voltage	VS		3.0 to 15	V
V <sub>CC</sub> supply voltage	VCC		3.0 to 15	V
Reference voltage	VREF	V <sub>CC</sub> ≤ 4V	0.0 to 1.0	V
		V <sub>CC</sub> > 4V	0.0 to 1.5	V

# **Electrical Characteristics** $Ta = 25^{\circ}C$ , $VS = V_{CC} = 5V$ , VREF = 0.3V

Deservator	Symbol	Conditions		Unit		
Parameter	Symbol		min	typ	max	Onit
[Output Block]						
VS system supply current	I <sub>VS</sub> OFF	PH1 = PH2 = 0V, EN1 = EN2 = 3.0V, ST = 3.0V			5	μΑ
	I <sub>VS</sub> ON	PH1 = PH2 = EN1 = EN2 = 0V, ST = 3.0V	28	40	52	mA
	I <sub>VS</sub> wt	PH1 = PH2 = EN1 = EN2 = ST = 0V			1	μΑ
Output saturation voltage 1	V <sub>O</sub> (sat) 1	I <sub>O</sub> = +0.2A (source)		0.2	0.4	V
Output saturation voltage 2	V <sub>O</sub> (sat) 2	I <sub>O</sub> = +0.4A (source)		0.3	0.5	V
Output saturation voltage 3	V <sub>O</sub> (sat) 3	I <sub>O</sub> = -0.2A (sink)		0.2	0.4	V
Output saturation voltage 4	V <sub>O</sub> (sat) 4	I <sub>O</sub> = -0.4A (sink)		0.3	0.5	V
Output leakage current	I <sub>O</sub> 1 (leak)	$V_O = V_{BB}$ (sink)			50	μΑ
	I <sub>O</sub> 2 (leak)	V <sub>O</sub> = 0V (source)	-50			μΑ
Upper and lower side output diodes						
Forward voltage 1 (upper side)	VF1	I = 400mA	0.9	1.1	1.3	٧
Forward voltage 2 (lower side)	VF2	I = 400mA	0.9	1.1	1.3	٧
[Logic Block]						
V <sub>CC</sub> system supply current	I <sub>CC</sub> OFF	PH1 = PH2 = 0V, EN1 = EN2 = 3.0V, ST = 3.0V	6.5	10	13.5	mA
	I <sub>CC</sub> ON	PH1 = PH2 = EN1 = EN2 = 0V, ST = 3.0V	7	11	15	mA
	I <sub>CC</sub> wt	PH1 = PH2 = EN1 = EN2 = ST = 0V			1	μΑ
Input voltage	V <sub>I</sub> on		2.0			٧
	V <sub>I</sub> off				0.8	V
Input current	I <sub>IN</sub>	V <sub>IN</sub> = 5V	70	100	130	μΑ
Reference voltage : 1V	V1V	I <sub>O</sub> = 1mA	0.95	1	1.05	V
Current setting reactive current	IE		-22	-17	-10.5	mA
Reference current	IREF	VREF = 0.3V, VE = 0.3V	-1			μΑ
CR pin current 1	ICR1	CR = 0.5V	-2			μΑ
CR pin current 2	ICR2	CR = 3V	1.65	2.2	2.75	mA
Sense voltage 1	VSEN1	VREF = 0.5V	0.475	0.5	0.525	V
Thermal shutdown temperature *	TS	*		170		°C

Note \* : Design guarantee value

#### **Truth Table**

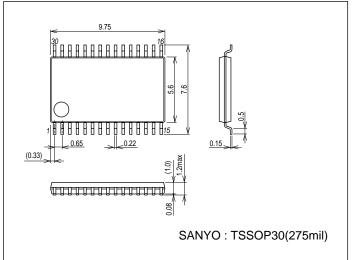
	Channel 1					Channel 2				
Input	Input		Output		Input		Output			
ST	PHASE1	ENABLE1	OUT1 <sup>-</sup>	OUT1	PHASE2	ENABLE2	OUT2 <sup>-</sup>	OUT2		
Н	L	L	Н	L	L	L	Н	L		
Н	Н	L	L	Н	Н	L	L	Н		
Н	*	Н	OFF	OFF	*	Н	OFF	OFF		
L	*	*	OFF	OFF	*	*	OFF	OFF		

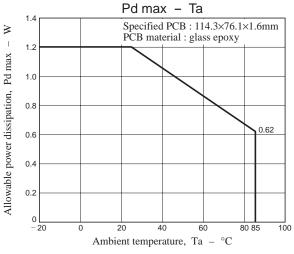
Note \*: Levels shown as an asterisk (\*) can be set to be either high or low.

#### **Package Dimensions**

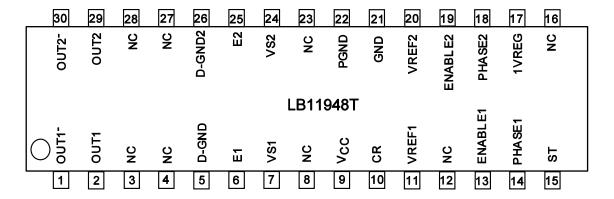
unit: mm (typ)

3259





# **Pin Assignment**

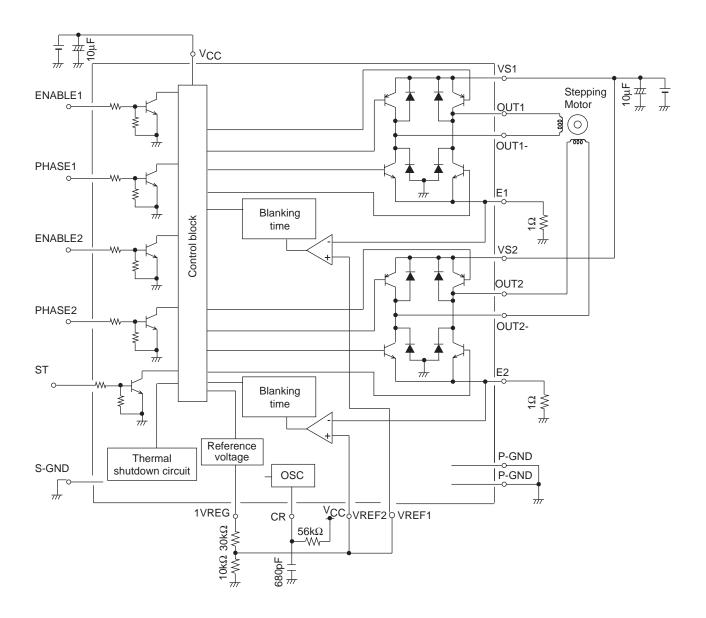


Top view

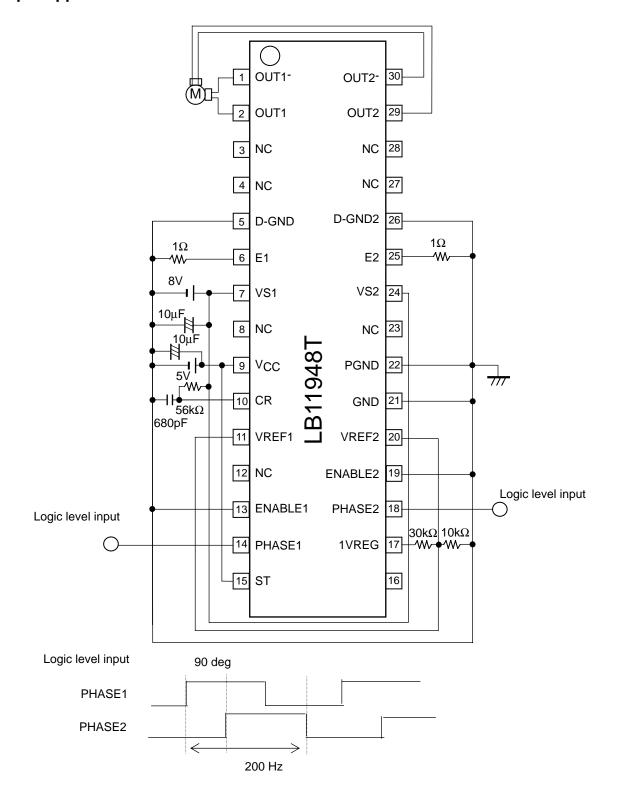
#### **Pin Functions**

	u	•
Pin No.	Pin Name	Description
1	OUT1	Output
2	OUT1	Output
3	NC	Unused
4	NC	Unused
5	D-GND	Lower side internal diode anode connection
6	E1	Constant current control sensing
		The motor current is set by the value of the sensing resistor Re connected between the E1 pin and ground.
		The current is set according to the following equation : IO = VREF/Re (A)
7	VS1	VS power supply
8	NC	Unused
9	Vcc	V <sub>CC</sub> power supply
10	CR	RC oscillator connection
11	VREF1	Current setting system reference voltage input
		VREF1 voltage range : 0 to 0.5V
12	NC	Unused
13	ENABLE1	Output is turned on when ENABLE1 is low, and the output is turned off (operating state) when ENABLE1 is high.
14	PHASE1	Logic level input : phase switching
		When PHASE1 = high: Output pin states: OUT1: high, OUT1-: low.
15	ST	When PHASE1 = low : Output pin states : OUT1 : low, OUT1 <sup>-</sup> : high.  Standby mode setting
15	31	When ST = high : the IC operates in normal operating mode.
		When ST = low : the IC operates in standby mode. The VS and $V_{CC}$ current drain levels are under 1 $\mu$ A in this mode.
16	NC	Unused
17	1VREG	1V regulator circuit output
		The LB11948 includes an internal 1V regulator circuit, and this pin is the output from that circuit. The VREF1 and VREF2
		reference voltages can be set by voltage dividing the 1V regulator output.
18	PHASE2	Logic level input : phase switching
		When PHASE2 = high: Output pin states: OUT2: high, OUT2 <sup>-</sup> : low.  When PHASE2 = low: Output pin states: OUT2: low, OUT2 <sup>-</sup> : high.
19	ENABLE2	Output is turned on when ENABLE2 is low, and the output is turned off (operating state) when ENABLE2 is high.
20	VREF2	Current setting reference voltage input
	******	VREF2 voltage range : 0 to 0.5V
21	GND	Ground (small signal circuit system ground)
22	PGND	Power system ground (high current circuit system ground)
23	NC	Unused
24	VS2	VS power supply
25	E2	Constant current control sensing
		The motor current is set by the value of the sensing resistor Re connected between the E2 pin and ground.
		The current is set according to the following equation : IO = VREF/Re (A)
26	D-GND2	Lower side internal diode anode connection
27	NC	Unused
28	NC	Unused
29	OUT2	Output
30	OUT2	Output

# **Block Diagram**



# **Sample Application Circuit**



# Drive Sequence Table 2 Phase Excitation Drive Sequence

Table 1 Clockwise drive

No.	PHASE1	ENABLE1	OUT1	OUT1 <sup>-</sup>	PHASE2	ENABLE2	OUT2	OUT2-
0	0	0	0	1	0	0	0	1
1	1	0	1	0	0	0	0	1
2	1	0	1	0	1	0	1	0
3	0	0	0	1	1	0	1	0

Table 2 Counterclockwise drive

No.	PHASE1	ENABLE1	OUT1	OUT1	PHASE2	ENABLE2	OUT2	OUT2
0	0	0	0	1	1	0	1	0
1	1	0	1	0	1	0	1	0
2	1	0	1	0	0	0	0	1
3	0	0	0	1	0	0	0	1

#### 1-2 Phase Excitation Drive Sequence

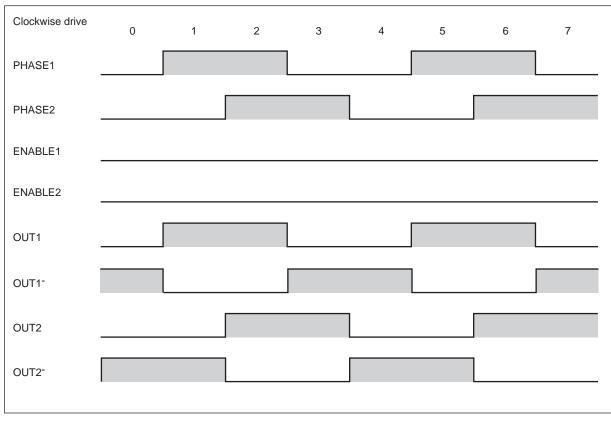
Table 3 Clockwise drive

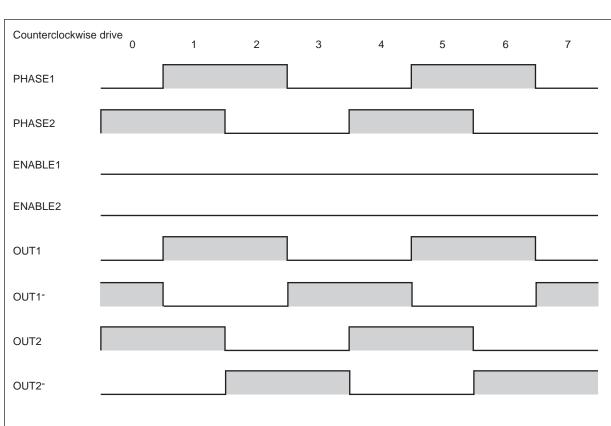
No.	PHASE1	ENABLE1	OUT1	OUT1 <sup>-</sup>	PHASE2	ENABLE2	OUT2	OUT2
0	0	0	0	1	0	1	OFF	OFF
1	0	0	0	1	0	0	0	1
2	1	1	OFF	OFF	0	0	0	1
3	1	0	1	0	0	0	0	1
4	1	0	1	0	1	1	OFF	OFF
5	1	0	1	0	1	0	1	0
6	0	1	OFF	OFF	1	0	1	0
7	0	0	0	1	1	0	1	0

Table 4 Counterclockwise drive

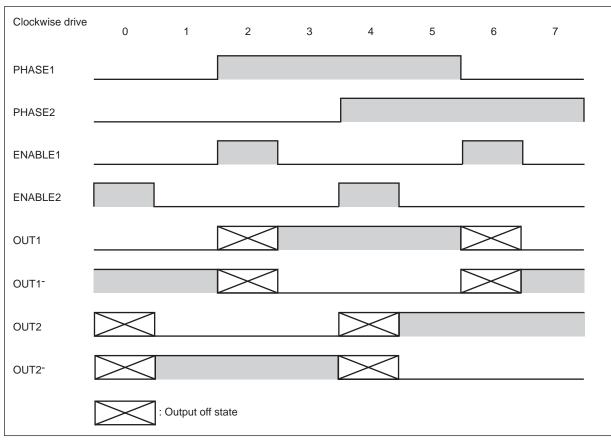
No.	PHASE1	ENABLE1	OUT1	OUT1 <sup>-</sup>	PHASE2	ENABLE2	OUT2	OUT2 <sup>-</sup>
0	0	0	0	1	1	1	OFF	OFF
1	0	0	0	1	1	0	1	0
2	1	1	OFF	OFF	1	0	1	0
3	1	0	1	0	1	0	1	0
4	1	0	1	0	0	1	OFF	OFF
5	1	0	1	0	0	0	0	1
6	0	1	OFF	OFF	0	0	0	1
7	0	0	0	1	0	0	0	1

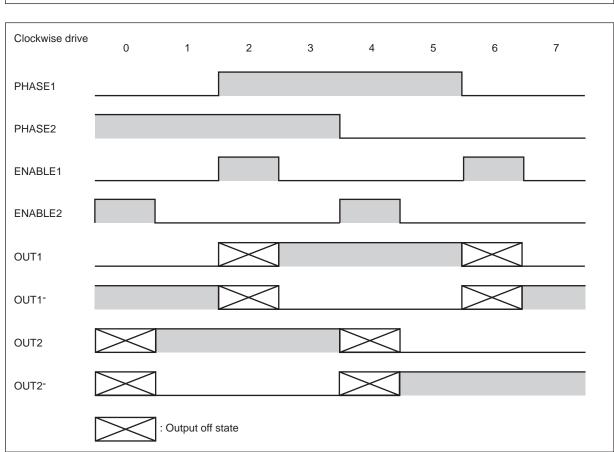
# 2 Phase Excitation Drive Sequence



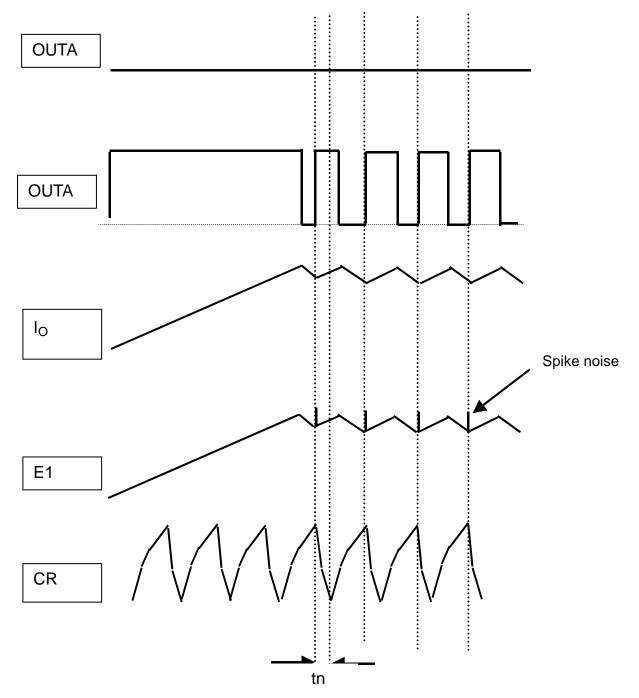


# 1-2 Phase Excitation Drive Sequence





# **Switching Operation Timing Chart**



tn: The noise canceller operating time

#### **Usage Notes**

(1) Simple Formulas for Determining Resister and Capacitor Values

The formula for setting the rising time (T1) and the falling time (T2) for the RC oscillator are shown below. (Refer to Fig. 1)

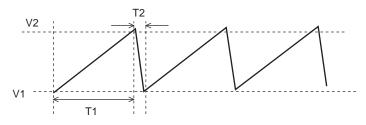


Fig. 1

Formulas

Oscillation period T = T1+T2 (sec)

Threshold voltages  $V1 = ((V_{CC} - V_{Set1}) \times 10.7k/48.7k) + V_{Set1}$  (V)

 $V2 = ((V_{CC} - V_{set2}) \times 42.7k/80.7k) + V_{set2}$  (V)

Vset1: VCE voltage of transistor for internal comparator hysteresis = 0.05V

Vset2: VCE voltage of reference resistance switching transistor of oscillation circuit = 0.1V

When charging:  $T1 = -C \times R \times ln \{(V_{CC} - V_2)/(V_{CC} - V_1)\}$  (sec)

When discharging:  $T2 = -C \times Rin \times ln (V1/V2)$  (sec)

Rin : Internal discharge resistance of the  $CR\ pin1.3k\Omega$ 

C : External capacitor R : External resistor

Oscillation frequency Fc = 1/T (Hz)

The T2 fall time serves as the noise canceling time (Tn). This time is a forced-on time for the output, and the output is not turned off even when the E pin voltage is higher than the sense voltage that has been preset by VREF.

#### (2) Constant current settings

The reference voltages of the VREF1 and VREF2 pins can be set by dividing the resistance voltage from the 1V regulator output pin (1VREG).

The output current is set using the VREF reference voltage applied to the VREF1 and VREF2 pins and the Re resistor connected between the E1 and E2 pins and ground. The bias current of the output transistor also flows from the E pins so that the Iout output current flowing to the motor is reduced by an amount equivalent to the bias current. In addition, in controlling the constant current, the voltage is sensed by the E pins (pad area on IC chip) so that the amount equivalent to the wire bonding resistance (rw) from the pad to the package pins is added to the current sensing resistance (Re).

Therefore, the formula for calculating the current setting is as shown below. (Refer to Fig. 2)

$$I_{OUT} = VREF / (Re + rw) - Ibias$$
 [A]

Re: Sensing resistance of resistor connected between E pins and ground

rw : Amount equivalent to wire bonding resistance from pad to pins  $rw = 50 - 100 m\Omega$ 

Ibias: Output transistor bias current

The Ibias current corresponds to the current setting reactive current (IE) in the specifications for the electrical characteristics.

Current setting reactive current IE ratings: Min: -22mA Typ: -17mA Max: -10.5mA

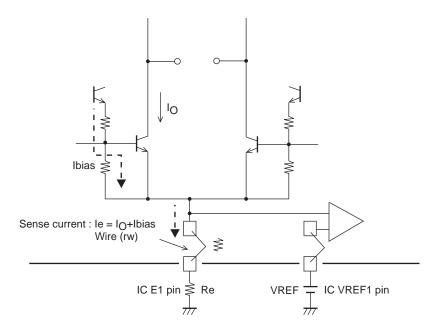


Fig. 2

#### (3) VREF pins

The VREF pins are the reference voltage input pins for the preset current, so take special care to ensure that they are not affected by noise. If these pins will be affected by noise, connect a capacitor to VREF1 and VREF2 pins.

#### (4) Notes on the Ground Pins

Since this IC switches large currents, the following notes on ground lines must be observed.

- The PCB pattern lines in areas that handle large currents must be as wide as possible so as to have low impedances, and must be kept as far as possible from the small signal systems.
- The ground terminals on the sensing resistors Re connected to the E pins (E1 and E2) must be connected as close as possible to the IC GND (pin 21), PGND (pin 22), or DGND (pins 5 and 26) pins as possible.
- The capacitors between V<sub>CC</sub> and ground and between V<sub>BB</sub> and ground must be as close as possible to the corresponding V<sub>CC</sub> and V<sub>BB</sub> pin in the pattern.

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