



## 16-Bit, Parallel Input Multiplying Digital-to-Analog Converter

### FEATURES

- $\pm 0.5$  LSB DNL
- $\pm 1$  LSB INL
- 16-Bit Monotonic
- Low Noise:  $10 \text{ nV}/\sqrt{\text{Hz}}$
- Low Power:  $I_{\text{DD}} = 2 \mu\text{A}$
- Analog Power Supply:  $+2.7 \text{ V}$  to  $+5.5 \text{ V}$
- 1.66 mA Full-Scale Current, with  $V_{\text{REF}} = 10 \text{ V}$
- Settling Time:  $0.5 \mu\text{s}$
- 4-Quadrant Multiplying Reference
- Reference Bandwidth: 8 MHz
- Reference Input:  $\pm 15 \text{ V}$
- Reference Dynamics:  $-105 \text{ THD}$
- SSOP-28 Package
- Industry-Standard Pin Configuration

### APPLICATIONS

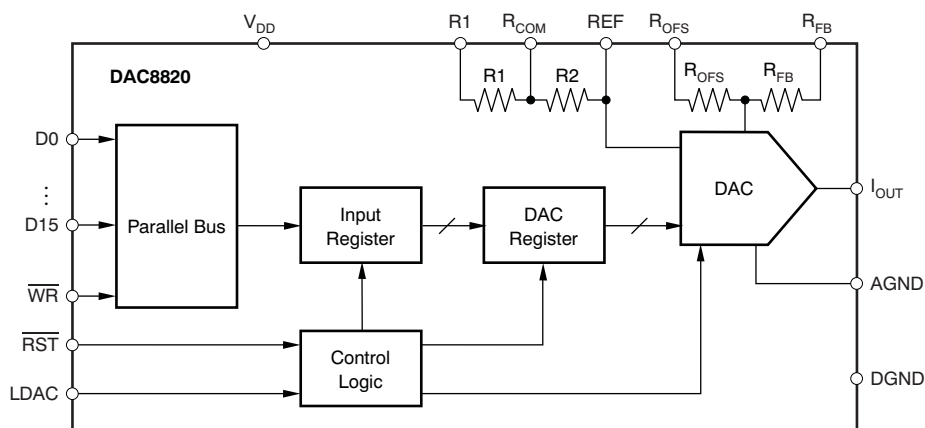
- Automatic Test Equipment
- Instrumentation
- Digitally Controlled Calibration
- Industrial Control PLCs

### DESCRIPTION

The DAC8820, a multiplying digital-to-analog converter (DAC), is designed to operate from a single 2.7 V to 5.5 V supply.

The applied external reference input voltage  $V_{\text{REF}}$  determines the full-scale output current. An internal feedback resistor ( $R_{\text{FB}}$ ) provides temperature tracking for the full-scale output when combined with an external, current-to-voltage (I/V) precision amplifier.

A parallel interface offers high-speed communications. The DAC8820 is packaged in a space-saving SSOP-28 package and has an industry-standard pinout.



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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

**ORDERING INFORMATION<sup>(1)</sup>**

PRODUCT	RELATIVE ACCURACY (LSB)	DIFFERENTIAL NONLINEARITY (LSB)	PACKAGE-LEAD (DESIGNATOR)	SPECIFIED TEMPERATURE RANGE	PACKAGE MARKING	ORDERING NUMBER	TRANSPORT MEDIA, QUANTITY
DAC8820IB	±2	±1	DB-28 (SSOP)	–40°C to +85°C	DAC8820	DAC8820IBDB	Tubes, 48
						DAC8820IBDBR	Tape and Reel, 2000
DAC8820IC	±1	±1	DB-28 (SSOP)	–40°C to +85°C	DAC8820	DAC8820ICDB	Tubes, 48
						DAC8820ICDBR	Tape and Reel, 2000

(1) For the most current package and ordering information see the Package Option Addendum at the end of this document, or see the TI web site at [www.ti.com](http://www.ti.com).

**ABSOLUTE MAXIMUM RATINGS**

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

	DAC8820	UNIT
V <sub>DD</sub> to GND	–0.3 to +7	V
Digital input voltage to GND	–0.3 to +V <sub>DD</sub> + 0.3	V
V (I <sub>OUT</sub> ) to GND	–0.3 to +V <sub>DD</sub> + 0.3	V
REF, R <sub>OFS</sub> , R <sub>FB</sub> , R <sub>1</sub> , R <sub>COM</sub> to AGND, DGND	±25	V
Operating temperature range	–40 to +85	°C
Storage temperature range	–65 to +150	°C
Junction temperature range (T <sub>J</sub> max)	+125	°C
Power dissipation	(T <sub>J</sub> max – T <sub>A</sub> ) / R <sub>θJA</sub>	W
Thermal impedance, R <sub>θJA</sub>	55	°C/W
ESD rating:		
Human Body Model (HBM)	4000	V
Charged Device Model (CDM)	1000	V

(1) Stresses above those listed under absolute maximum ratings may cause permanent damage to the device. Exposure to absolute maximum conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS

All specifications at  $-40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$ ,  $V_{\text{DD}} = +2.7\text{ V}$  to  $+5.5\text{ V}$ ,  $I_{\text{OUT}} = \text{virtual GND}$ ,  $\text{GND} = 0\text{ V}$ ,  $V_{\text{REF}} = 10\text{ V}$ , and  $T_{\text{A}} = \text{full operating temperature}$ , unless otherwise noted.

PARAMETER	CONDITIONS	DAC8820			UNITS
		MIN	TYP	MAX	
<b>STATIC PERFORMANCE</b>					
Resolution		16			Bits
Relative accuracy	DAC8820IB			$\pm 2$	LSB
Relative accuracy	DAC8820IC			$\pm 1$	LSB
Differential nonlinearity			$\pm 0.5$	$\pm 1$	LSB
Output leakage current	Data = 0000h, $T_{\text{A}} = +25^{\circ}\text{C}$			5	nA
Output leakage current	Data = 0000h, $T_{\text{A}} = T_{\text{MAX}}$			10	nA
Full-scale gain error	Unipolar, data = FFFFh		2	$\pm 16$	LSB
	Bipolar, data = FFFFh		2	$\pm 16$	LSB
Full-scale temperature coefficient			1	2	ppm/ $^{\circ}\text{C}$
Bipolar zero scale error	$T_{\text{A}} = +25^{\circ}\text{C}$			$\pm 5$	LSB
	$T_{\text{A}} = T_{\text{MAX}}$			$\pm 8$	LSB
PSRR	Power-supply rejection ratio; $V_{\text{DD}} = 5\text{ V} \pm 10\%$		$\pm 0.2$	$\pm 2.0$	LSB/V
<b>OUTPUT CHARACTERISTICS<sup>(1)</sup></b>					
Output current			1.66		mA
Output capacitance	Code dependent		50		pF
<b>REFERENCE INPUT</b>					
$V_{\text{REF}}$ Range		-15		15	V
$R_{\text{REF}}$	Input resistance (unipolar)	4.5	6	7.5	k $\Omega$
Input capacitance			5		pF
R1/R2	R1/R2 resistance (bipolar)	9	12	15	k $\Omega$
$R_{\text{OFS}}$ , $R_{\text{FB}}$	Feedback and offset resistance	9	12	15	k $\Omega$
<b>LOGIC INPUTS AND OUTPUT<sup>(1)</sup></b>					
Input low voltage	$V_{\text{IL}}$	$V_{\text{DD}} = +2.7\text{ V}$		0.6	V
	$V_{\text{IL}}$	$V_{\text{DD}} = +5\text{ V}$		0.8	V
Input high voltage	$V_{\text{IH}}$	$V_{\text{DD}} = +2.7\text{ V}$	2.1		V
	$V_{\text{IH}}$	$V_{\text{DD}} = +5\text{ V}$	2.4		V
Input leakage current	$I_{\text{IL}}$		0.001	1	$\mu\text{A}$
Input capacitance	$C_{\text{IL}}$			8	pF
<b>INTERFACE TIMING, <math>V_{\text{DD}} = +5.0\text{V}</math><sup>(1)</sup> (See Figure 40 and Table 1)</b>					
	$t_{\text{DS}}$	Data to $\overline{\text{WR}}$ setup time	20		ns
	$t_{\text{DH}}$	Data to $\overline{\text{WR}}$ hold time	0		ns
	$t_{\text{WR}}$	$\overline{\text{WR}}$ pulse width	20		ns
	$t_{\text{LDAC}}$	LDAC pulse width	20		ns
Data setup time	$t_{\text{RST}}$	$\overline{\text{RST}}$ pulse width	20		ns
Data hold time	$t_{\text{LWD}}$	$\overline{\text{WR}}$ to LDAC delay time	0		ns
<b>INTERFACE TIMING, <math>V_{\text{DD}} = +2.7\text{V}</math><sup>(1)</sup> (See Figure 40 and Table 1)</b>					
	$t_{\text{DS}}$	Data to $\overline{\text{WR}}$ setup time	35		ns
	$t_{\text{DH}}$	Data to $\overline{\text{WR}}$ hold time	0		ns
	$t_{\text{WR}}$	$\overline{\text{WR}}$ pulse width	35		ns
	$t_{\text{LDAC}}$	LDAC pulse width	35		ns
Data setup time	$t_{\text{RST}}$	$\overline{\text{RST}}$ pulse width	35		ns
Data hold time	$t_{\text{LWD}}$	$\overline{\text{WR}}$ to LDAC delay time	0		ns

(1) Specified by design and characterization; not production tested.

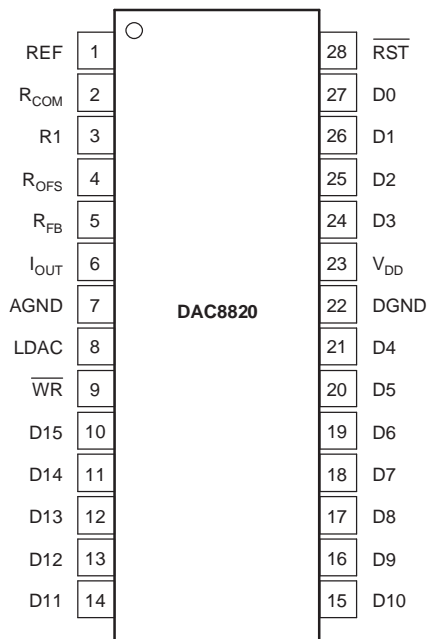
**ELECTRICAL CHARACTERISTICS (continued)**

All specifications at  $-40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$ ,  $V_{\text{DD}} = +2.7\text{ V}$  to  $+5.5\text{ V}$ ,  $I_{\text{OUT}} = \text{virtual GND}$ ,  $\text{GND} = 0\text{ V}$ ,  $V_{\text{REF}} = 10\text{ V}$ , and  $T_{\text{A}} = \text{full operating temperature}$ , unless otherwise noted.

PARAMETER	CONDITIONS	DAC8820			UNITS
		MIN	TYP	MAX	
<b>POWER REQUIREMENTS</b>					
$V_{\text{DD}}$		2.7		5.5	V
$I_{\text{DD}}$ (normal operation)	Logic inputs = 0 V			5	$\mu\text{A}$
$V_{\text{DD}} = +4.5\text{ V}$ to $+5.5\text{ V}$	$V_{\text{IH}} = V_{\text{DD}}$ and $V_{\text{IL}} = \text{GND}$		3	5	$\mu\text{A}$
$V_{\text{DD}} = +2.7\text{ V}$ to $+3.6\text{ V}$	$V_{\text{IH}} = V_{\text{DD}}$ and $V_{\text{IL}} = \text{GND}$		1	2.5	$\mu\text{A}$
<b>AC CHARACTERISTICS<sup>(2)</sup></b>					
Output current settling time			0.5		$\mu\text{s}$
Reference multiplying BW	$V_{\text{REF}} = 5\text{ V}_{\text{PP}}$ , Data = FFFFh		8		MHz
DAC glitch impulse	$V_{\text{REF}} = 0\text{ V}$ to $10\text{ V}$ , Data = 7FFFh to 8000h to 7FFFh		2		nV-s
Feedthrough error $V_{\text{OUT}}/V_{\text{REF}}$	Data = 0000h, $V_{\text{REF}} = 10\text{ kHz}$ , $\pm 10V_{\text{PP}}$		-70		dB
Digital feedthrough	LDAC = Logic low, $V_{\text{REF}} = -10\text{ V}$ to $+10\text{ V}$ Any code change		1		nV-s
Total harmonic distortion	$V_{\text{REF}} = 6V_{\text{RMS}}$ , Data = FFFFh, $f = 1\text{ kHz}$		-105		dB
Output spot noise voltage			10		$\text{nV}/\sqrt{\text{Hz}}$

(2) Specified by design and characterization; not production tested.

**PIN ASSIGNMENTS**



PIN #	NAME	DESCRIPTION
2	$R_{\text{COM}}$	Center tap of two 4-quadrant resistors (R1 and R2).
3	R1	4-quadrant resistor (R1).
4	$R_{\text{OFS}}$	Bipolar offset resistor
5	$R_{\text{FB}}$	Internal matching feedback resistor
6	$I_{\text{OUT}}$	DAC current output
7	AGND	Analog ground
8	LDAC	Digital input load DAC control. When LDAC is high, data is loaded from input register into a DAC register, updating the DAC output.
9	$\overline{\text{WR}}$	Write control digital input. Active low. When $\overline{\text{WR}}$ is taken to logic low, data is loaded from the digital input pins (D0–D15) into a 16-bit input register.
10–21	D15–D4	Digital input data bits. D15 is MSB.
22	DGND	Digital ground
23	$V_{\text{DD}}$	Positive power supply
24–27	D3–D0	Digital Input data bits. D0 is LSB.
28	$\overline{\text{RST}}$	Reset. Active low. When $\overline{\text{RST}}$ is taken to logic low, the DAC output and all internal registers are set to zero code for the DAC8820.

**TERMINAL FUNCTIONS**

PIN #	NAME	DESCRIPTION
1	REF	Reference input and 4-quadrant Resistor (R2).

**TYPICAL CHARACTERISTICS:  $V_{DD} = +5\text{ V}$**

At  $T_A = +25^\circ\text{C}$ , unless otherwise noted.

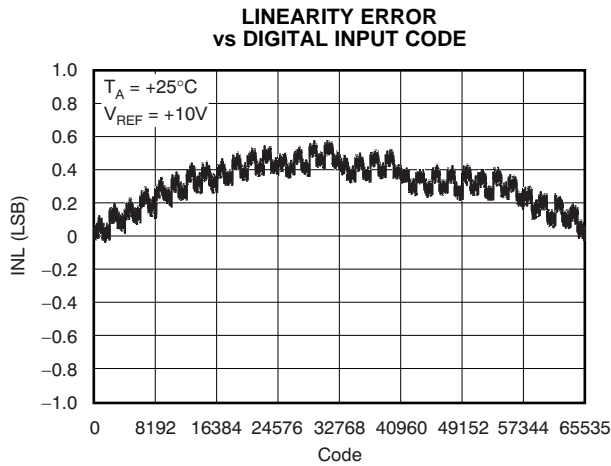


Figure 1.

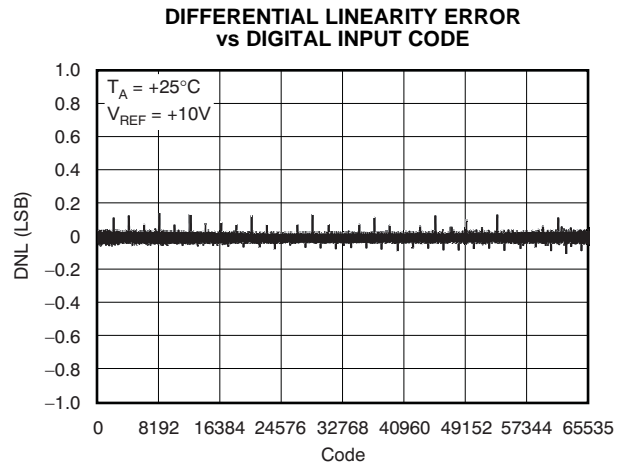


Figure 2.

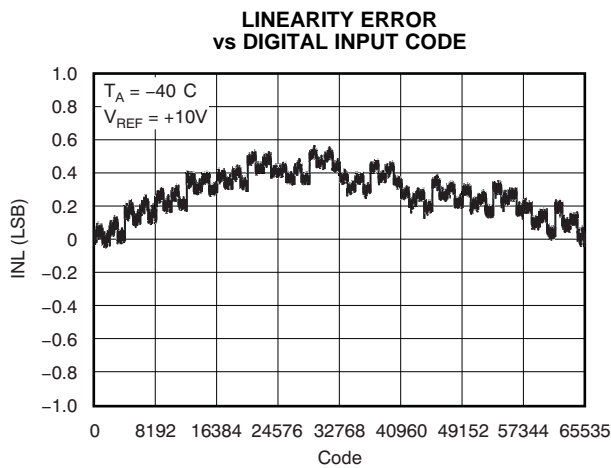


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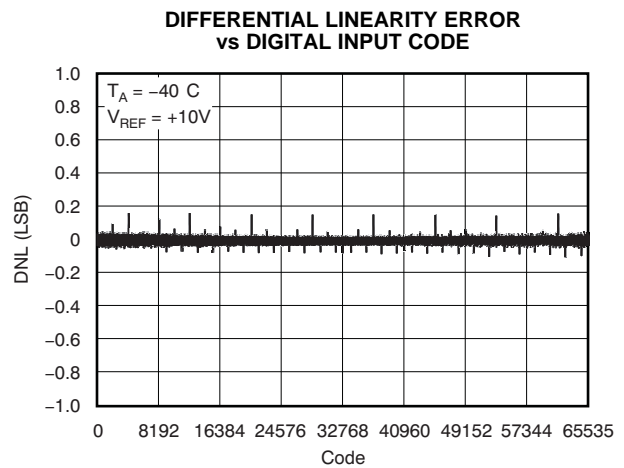


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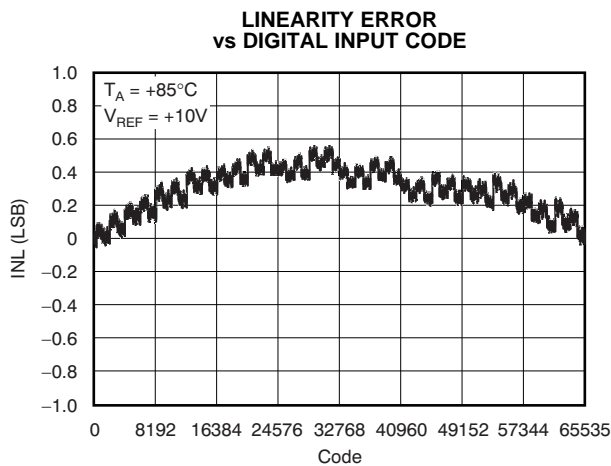


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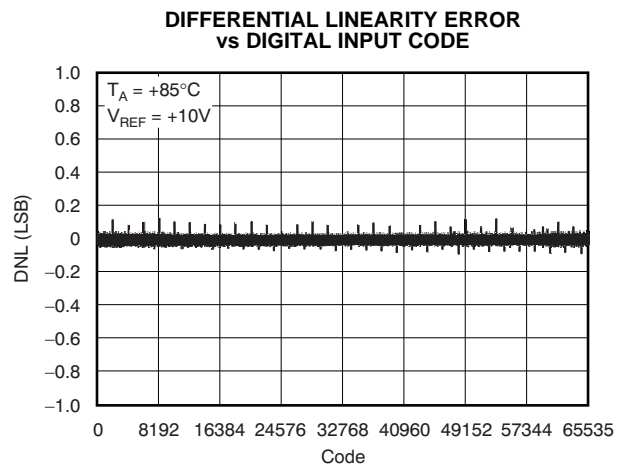


Figure 6.

TYPICAL CHARACTERISTICS:  $V_{DD} = +5\text{ V}$  (continued)

At  $T_A = +25^\circ\text{C}$ , unless otherwise noted.

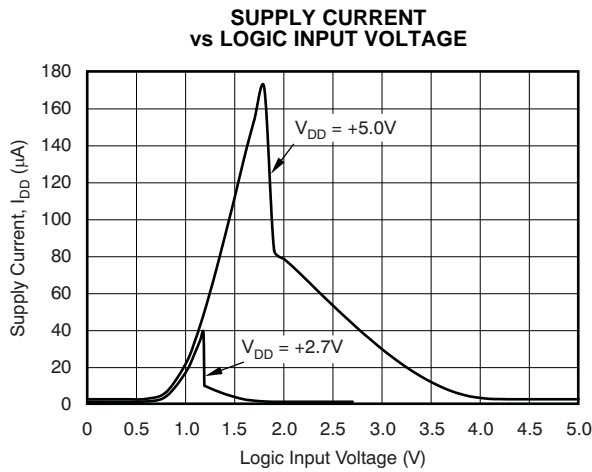


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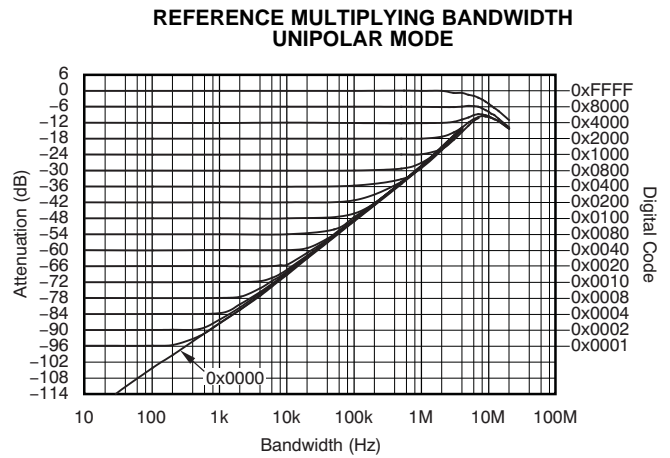


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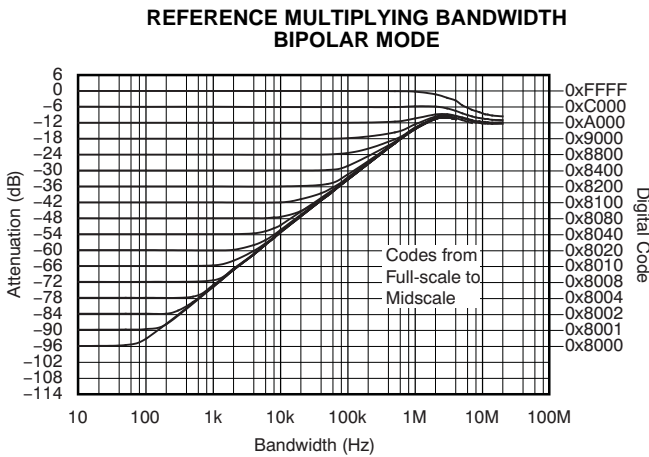


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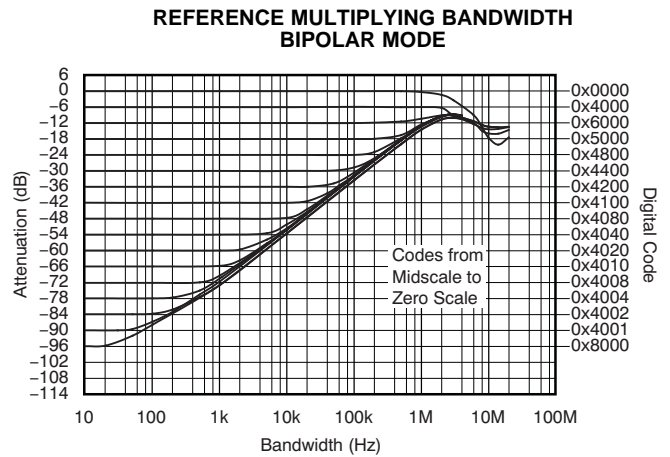


Figure 10.

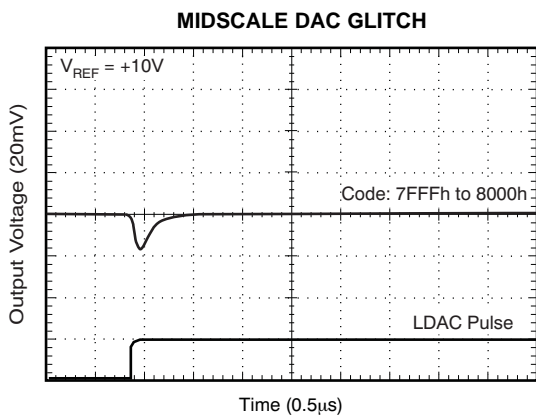


Figure 11.

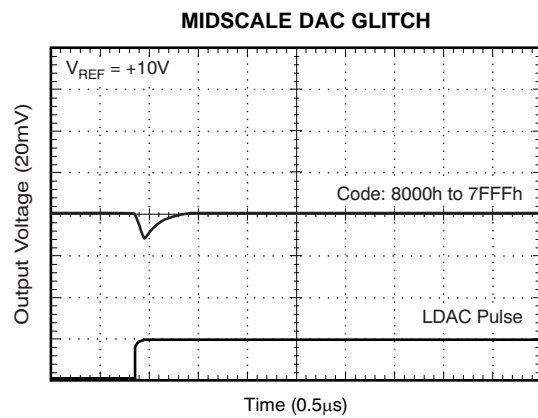
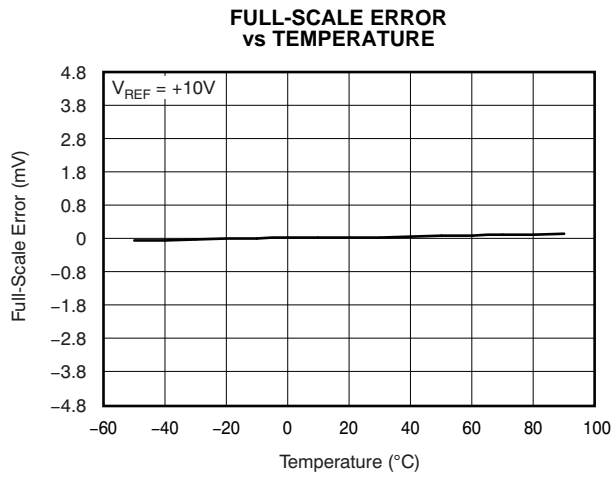


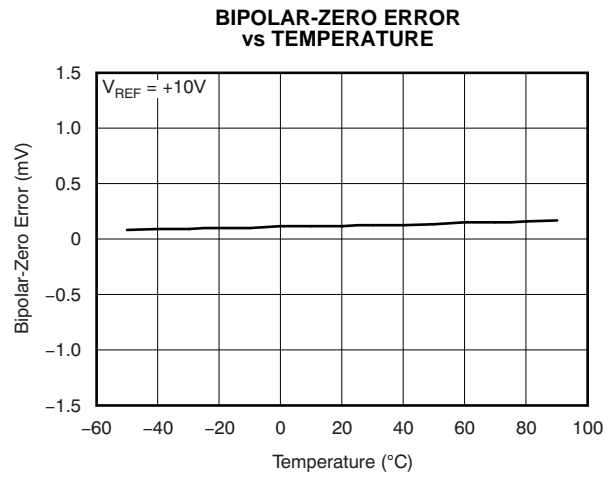
Figure 12.

**TYPICAL CHARACTERISTICS:  $V_{DD} = +5\text{ V}$  (continued)**

At  $T_A = +25^\circ\text{C}$ , unless otherwise noted.



**Figure 13.**



**Figure 14.**

**TYPICAL CHARACTERISTICS:  $V_{DD} = +2.7\text{ V}$**

At  $T_A = +25^\circ\text{C}$ , unless otherwise noted.

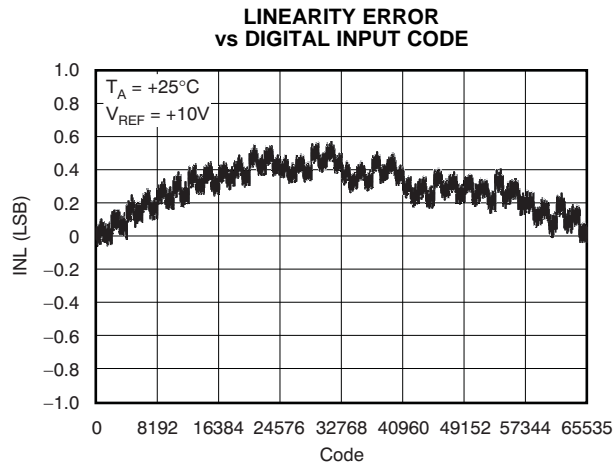


Figure 15.

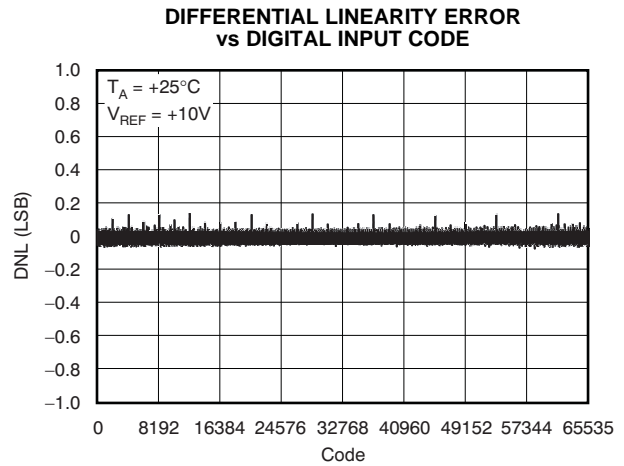


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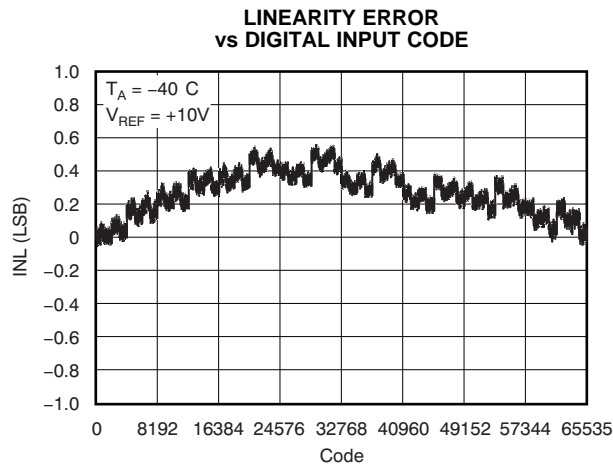


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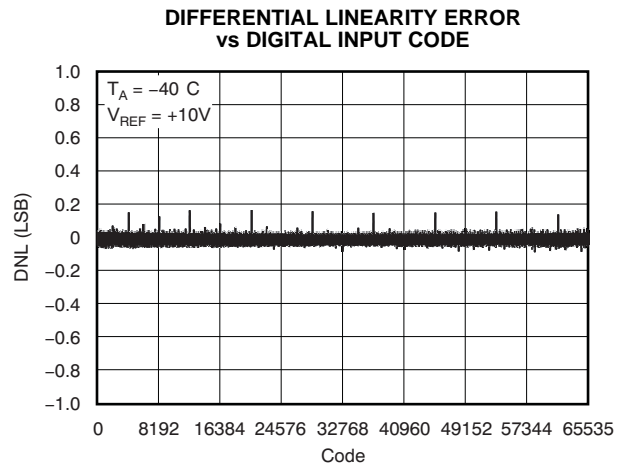


Figure 18.

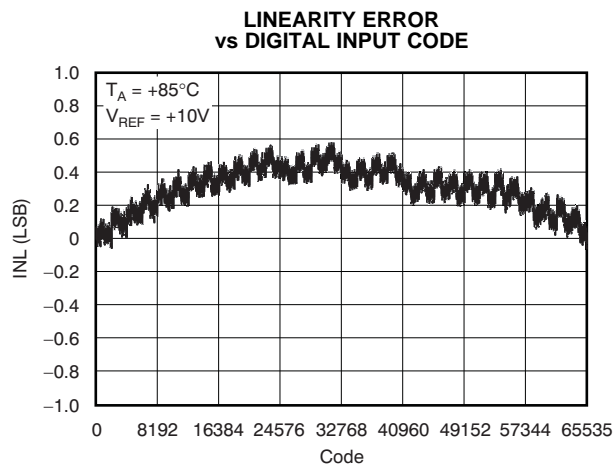


Figure 19.

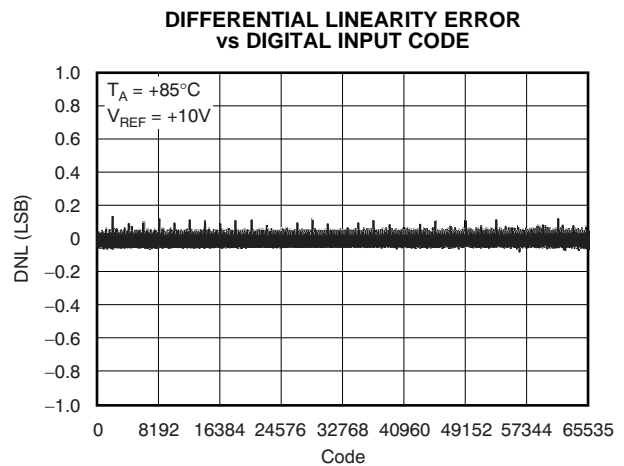
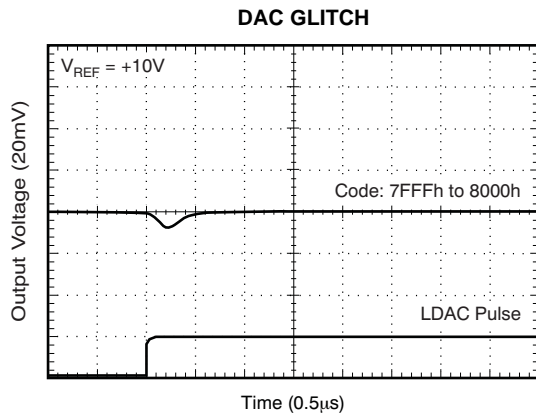


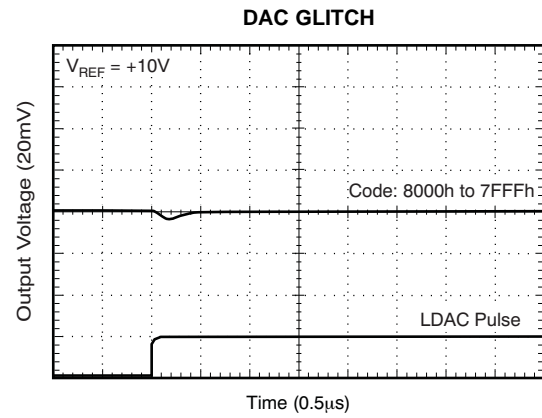
Figure 20.

**TYPICAL CHARACTERISTICS:  $V_{DD} = +2.7\text{ V}$  (continued)**

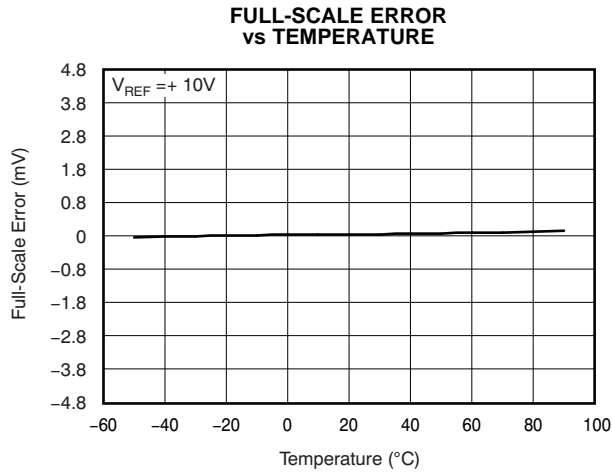
At  $T_A = +25^\circ\text{C}$ , unless otherwise noted.



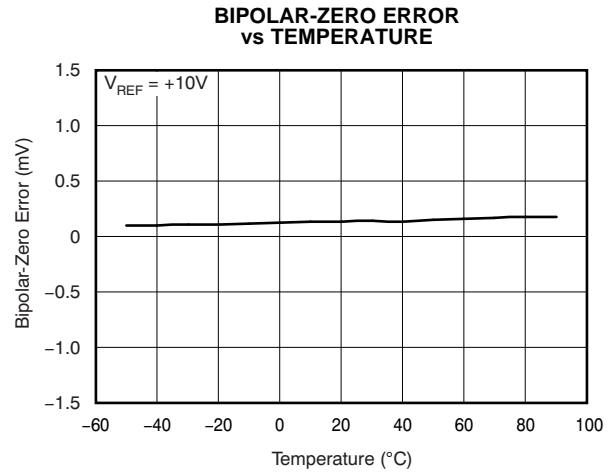
**Figure 21.**



**Figure 22.**



**Figure 23.**



**Figure 24.**

**TYPICAL CHARACTERISTICS**

At  $T_A = +25^\circ\text{C}$ , unless otherwise noted.

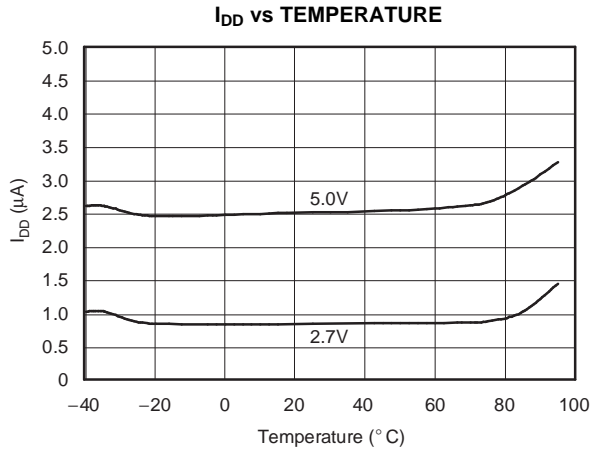


Figure 25.

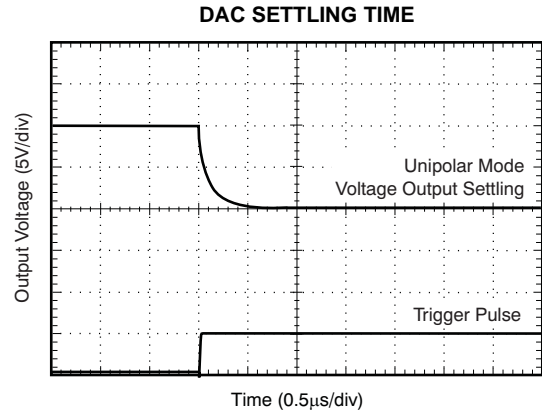


Figure 26.

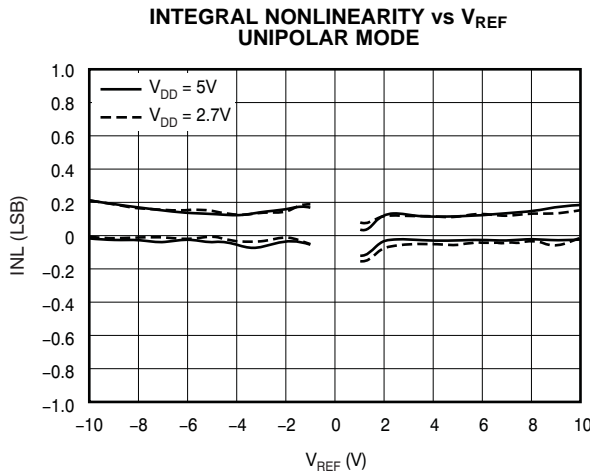


Figure 27.

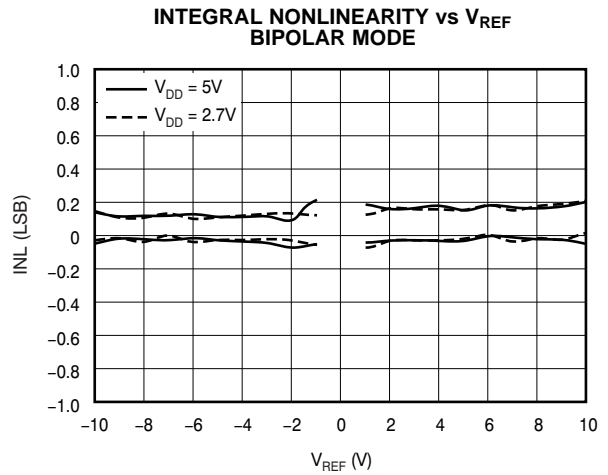


Figure 28.

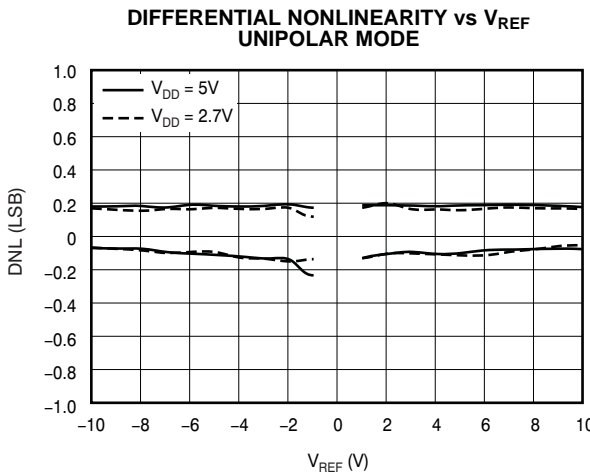


Figure 29.

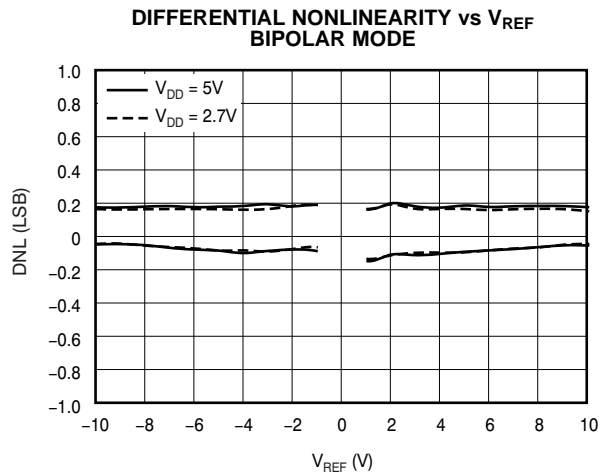


Figure 30.

TYPICAL CHARACTERISTICS (continued)

At  $T_A = +25^\circ\text{C}$ , unless otherwise noted.

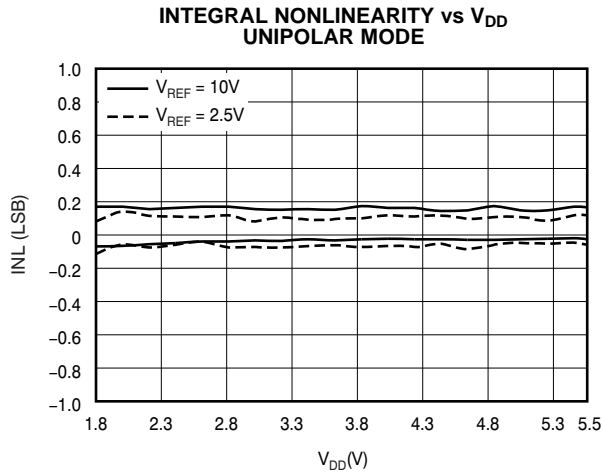


Figure 31.

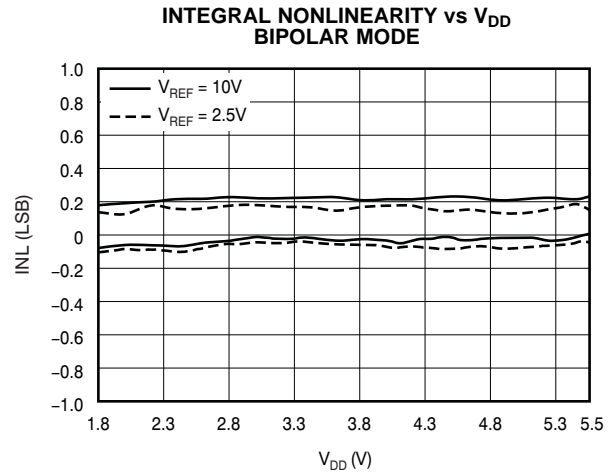


Figure 32.

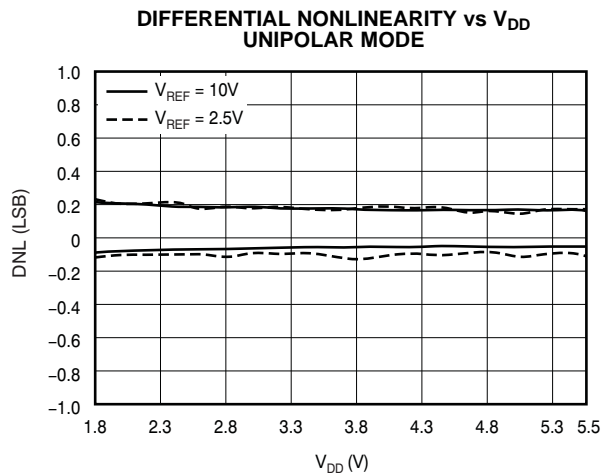


Figure 33.

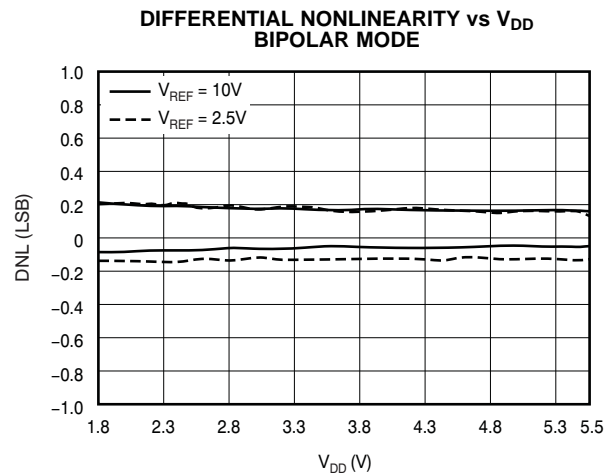


Figure 34.

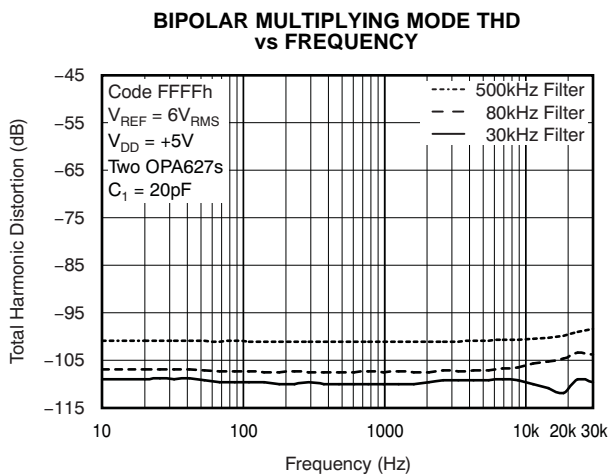


Figure 35.

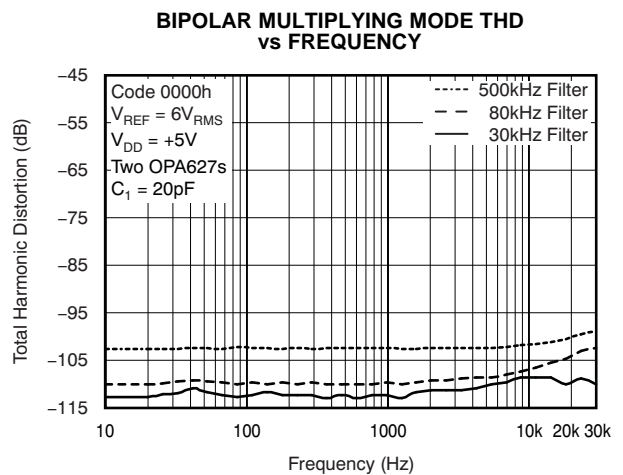


Figure 36.

**TYPICAL CHARACTERISTICS (continued)**

At  $T_A = +25^\circ\text{C}$ , unless otherwise noted.

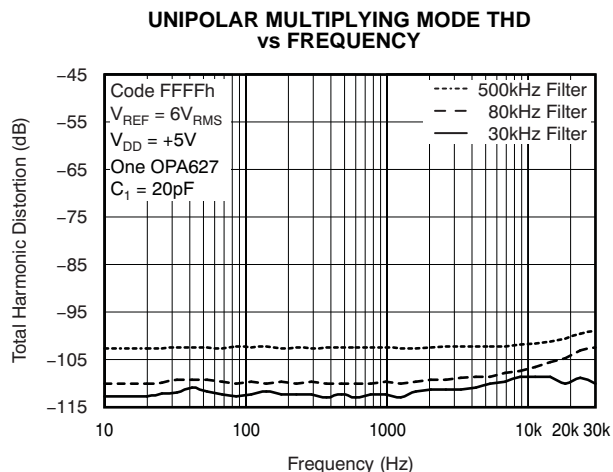


Figure 37.

**THEORY OF OPERATION**

The DAC8820 is a multiplying, single-channel current output, 16-bit DAC. The architecture, illustrated in Figure 38, is an R-2R ladder configuration with the three MSBs segmented. Each 2R leg of the ladder is either switched to GND or to the  $I_{OUT}$  terminal. The  $I_{OUT}$  terminal of the DAC is held at a virtual GND potential by the use of an external I/V converter op amp. The R-2R ladder is connected to an external reference input ( $V_{REF}$ ) that determines the DAC full-scale current. The R-2R ladder presents a code independent load impedance to the external reference of  $6\text{ k}\Omega \pm 25\%$ . The external reference voltage can vary in a range of  $-15\text{ V}$  to  $+15\text{ V}$ , thus providing bipolar  $I_{OUT}$  current operation. By using an external I/V converter op amp and the  $R_{FB}$  resistor in the DAC8820, an output voltage range of  $-V_{REF}$  to  $+V_{REF}$  can be generated.

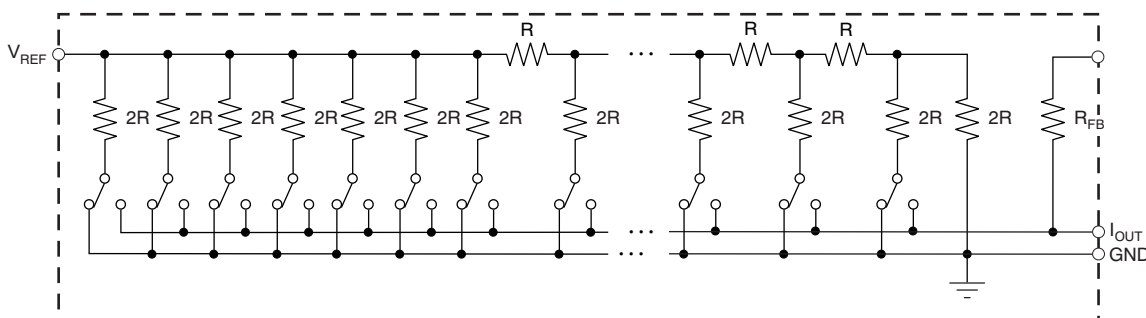


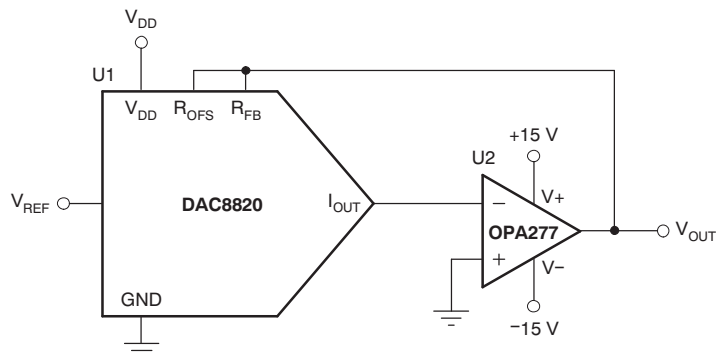
Figure 38. Equivalent R-2R DAC Circuit

The DAC output voltage is determined by  $V_{REF}$  and the digital data (D) according to Equation 1:

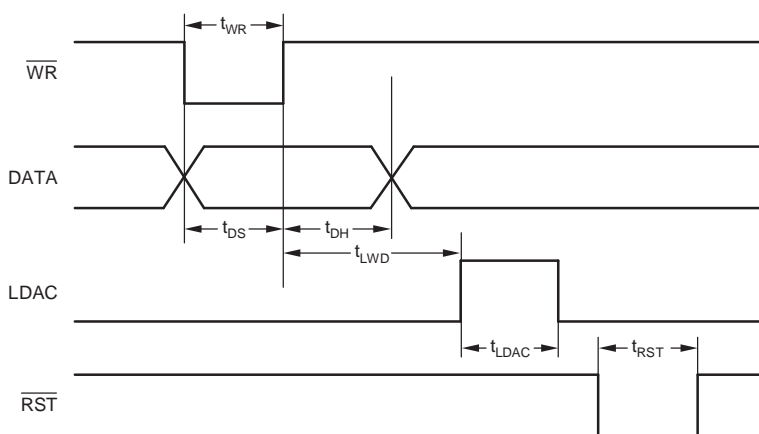
$$V_{OUT} = -V_{REF} \times \frac{D}{65536} \tag{1}$$

Each DAC code determines the 2R-leg switch position to either GND or  $I_{OUT}$ . The external I/V converter op amp noise gain will also change because the DAC output impedance (as seen looking into the  $I_{OUT}$  terminal) changes versus code. Because of this, the external I/V converter op amp must have a sufficiently low offset voltage such that the amplifier offset is not modulated by the DAC  $I_{OUT}$  terminal impedance change. External op amps with large offset voltages can produce INL errors in the transfer function of the DAC8820 because of offset modulation versus DAC code. For best linearity performance of the DAC8820, an op amp (OPA277) is recommended, as shown in Figure 39. This circuit allows  $V_{REF}$  to swing from  $-10\text{ V}$  to  $+10\text{ V}$ .

**THEORY OF OPERATION (continued)**





**Figure 39. Voltage Output Configuration**



**Figure 40. DAC8820 Timing Diagram**

**Table 1. Function of Control Inputs**

CONTROL INPUTS			REGISTER OPERATION
RST	WR	LDAC	
0	X	X	Asynchronous operation. Reset the input and DAC register to a predetermined value. The DAC8820 is reset to all 0s.
1	0	0	Load the input register with all 16 data bits.
1	1	1	Load the DAC register with the contents of the input register.
1	0	1	The input and DAC register are transparent.
1			LDAC and WR are tied together and programmed as a pulse. The 16 data bits are loaded into the input register on the falling edge of the pulse and then loaded into the DAC register on the rising edge of the pulse.
1	1	0	No register operation.

## APPLICATION INFORMATION

### Multiplying Mode THD vs Frequency

Figure 35 and Figure 36 show the DAC8820 bipolar 4-quadrant multiplying mode total harmonic distortion (THD) versus frequency. Figure 35 shows the bipolar multiplying mode THD with the DAC8820 set to a full-scale code of FFFFh. Figure 36 shows the bipolar multiplying mode THD with the DAC8820 set to a minus full scale code of 0000h. In both graphs, two OPA627s are used for both the DAC output op amp and the reference inverting amplifier. A  $6V_{RMS}$  sine wave is used for the reference input  $V_{REF}$  and is swept in frequency from 10 Hz to 30 kHz. The THD levels versus frequency are illustrated at various DAC output filtering levels using an external ac-coupled low-pass filter.

Figure 37 illustrates the DAC8820 unipolar 2-quadrant multiplying mode THD vs. frequency. The DAC8820 is set to a full-scale code of FFFFh. A single OPA627 is used for the DAC output op amp.

### Stability Circuit

For a current-to-voltage (I/V) design, as shown in Figure 41, the DAC8820 current output ( $I_{OUT}$ ) and the connection with the inverting node of the op amp should be as short as possible and laid out according to correct printed circuit board (PCB) layout design. For each code change there is a step function. If the gain bandwidth product (GBP) of the op amp is limited and parasitic capacitance is excessive at the inverting node, then gain peaking is possible. Therefore, a compensation capacitor C1 (4 pF to 20 pF, typ) can be added to the design for circuit stability, as shown in Figure 41.

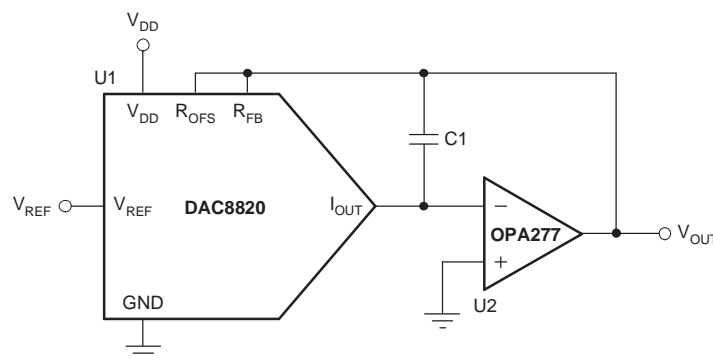


Figure 41. Gain Peaking Prevention Circuit with Compensation Capacitor

### Bipolar Output Circuit

The DAC8820, as a 4-quadrant multiplying DAC, can be used to generate a bipolar output. The polarity of the full-scale output ( $I_{OUT}$ ) is the inverse of the input reference voltage at  $V_{REF}$ .

Using a dual op amp, such as the OPA2277, full 4-quadrant operation can be achieved with minimal components. Figure 42 demonstrates a  $\pm 10 V_{OUT}$  circuit with a fixed +10 V reference.

APPLICATION INFORMATION (continued)

$$V_{OUT} = \left( \frac{D}{32,768} - 1 \right) \times V_{REF} \quad (2)$$

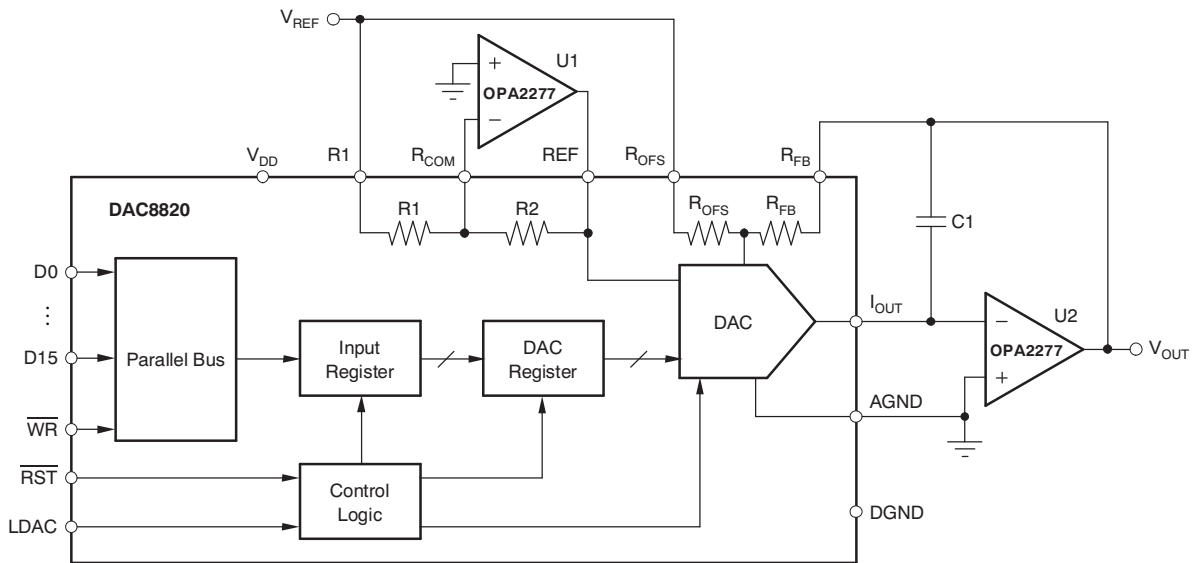


Figure 42. Bipolar Output Circuit

Programmable Current Source Circuit

A DAC8820 can be integrated into the circuit in Figure 43 to implement an improved Howland current pump for precise V/I conversions. Bidirectional current flow and high-voltage compliance are two features of the circuit. With a matched resistor network, the load current of the circuit is shown by Equation 3:

$$I_L = \frac{(R_2 + R_3) / R_1}{R_3} \times V_{REF} \times D \quad (3)$$

The value of R3 in the previous equation can be reduced to increase the output current drive of U3. U3 can drive ±20 mA in both directions with voltage compliance limited up to 15 V by the U3 voltage supply. Elimination of the circuit compensation capacitor (C1) in the circuit is not suggested as a result of the change in the output impedance (Z<sub>O</sub>), according to Equation 4:

$$Z_O = \frac{R_1' R_3 (R_1 + R_2)}{R_1 (R_2' + R_3') - R_1' (R_2 + R_3)} \quad (4)$$

As shown in Equation 4, Z<sub>O</sub> with matched resistors is infinite and the circuit is optimum for use as a current source. However, if unmatched resistors are used, Z<sub>O</sub> is positive or negative with negative output impedance being a potential cause of oscillation. Therefore, by incorporating C1 into the circuit, possible oscillation problems are eliminated. The value of C1 can be determined for critical applications; for most applications, however, a value of several pF is suggested.

APPLICATION INFORMATION (continued)

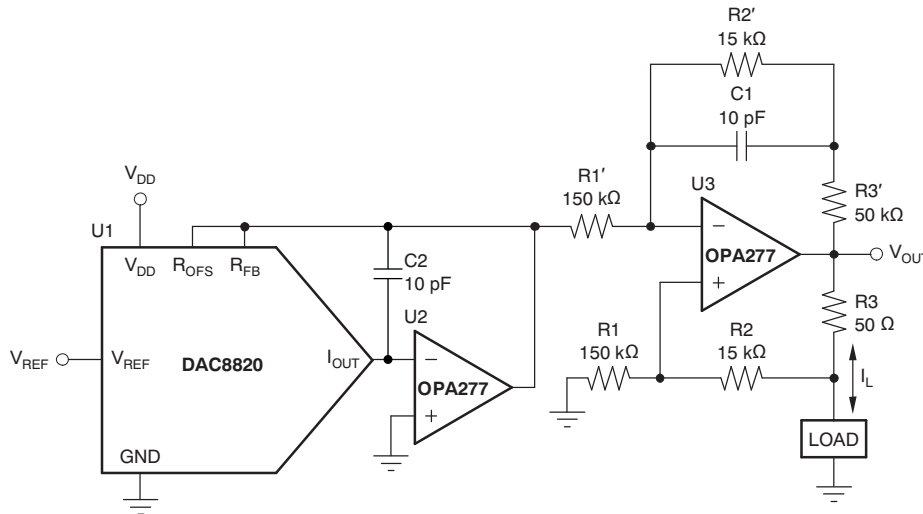


Figure 43. Programmable Bidirectional Current Source Circuit

Cross-Reference

The DAC8820 has an industry-standard pinout. [Table 2](#) provides the cross-reference information.

Table 2. Cross-Reference

PRODUCT	BIT	INL (LSB)	DNL (LSB)	SPECIFIED TEMPERATURE RANGE	PACKAGE DESCRIPTION	PACKAGE OPTION	CROSS-REFERENCE PART
DAC8820IBDB	16	±2	±1	–40°C to +85°C	SSOP-28	SSOP-28	LTC1597BIG
DAC8820ICDB	16	±1	±1	–40°C to +85°C	SSOP-28	SSOP-28	LTC1597AIG

## Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

<b>Changes from B Revision (March, 2006) to C Revision</b>	<b>Page</b>
• Changed to "current-to-voltage" from "voltage-to-current" .....	1
• Added bipolar zero scale error specification.....	3

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**PACKAGING INFORMATION**

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
DAC8820IBDB	ACTIVE	SSOP	DB	28	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
DAC8820IBDBG4	ACTIVE	SSOP	DB	28	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
DAC8820IBDBR	ACTIVE	SSOP	DB	28	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
DAC8820IBDBRG4	ACTIVE	SSOP	DB	28	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
DAC8820ICDB	ACTIVE	SSOP	DB	28	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
DAC8820ICDBG4	ACTIVE	SSOP	DB	28	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
DAC8820ICDBR	ACTIVE	SSOP	DB	28	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
DAC8820ICDBRG4	ACTIVE	SSOP	DB	28	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR

<sup>(1)</sup> The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

<sup>(2)</sup> Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

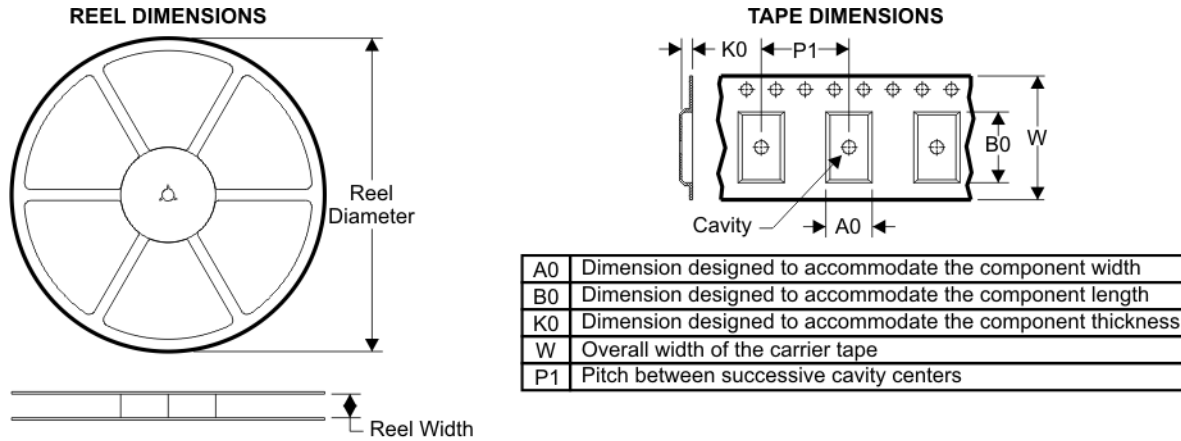
**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(3)</sup> MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

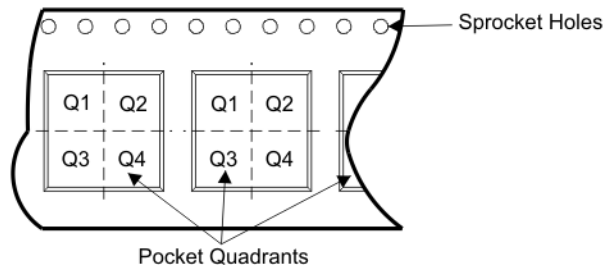
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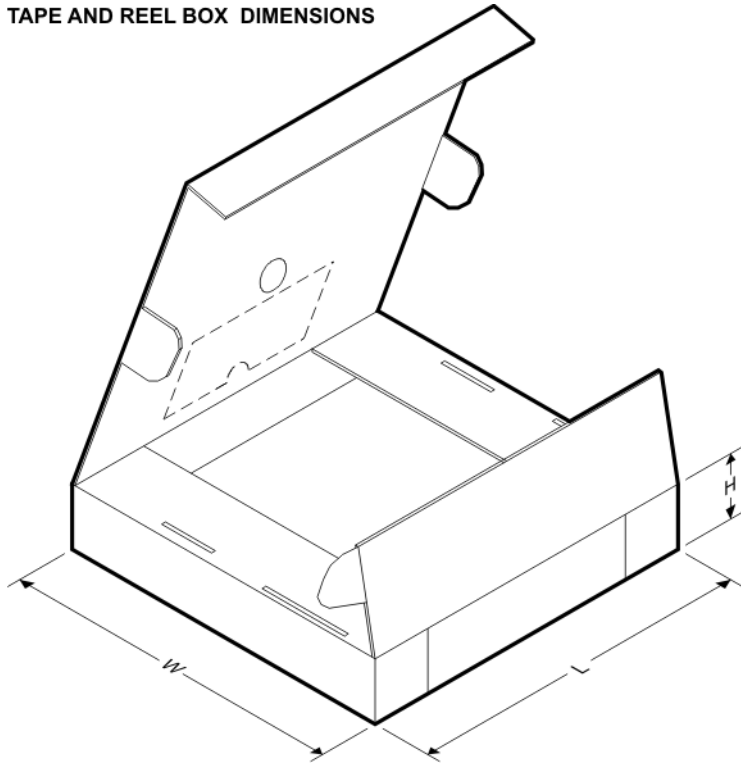


**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**



Device	Package	Pins	Site	Reel Diameter (mm)	Reel Width (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
DAC8820IBDBR	DB	28	SITE 60	330	16	8.1	10.4	2.5	12	16	Q1
DAC8820ICDBR	DB	28	SITE 60	330	16	8.1	10.4	2.5	12	16	Q1

**TAPE AND REEL BOX DIMENSIONS**



Device	Package	Pins	Site	Length (mm)	Width (mm)	Height (mm)
DAC8820IBDBR	DB	28	SITE 60	346.0	346.0	33.0
DAC8820ICDBR	DB	28	SITE 60	346.0	346.0	33.0

DB (R-PDSO-G\*\*)

PLASTIC SMALL-OUTLINE

28 PINS SHOWN



- NOTES: A. All linear dimensions are in millimeters.  
 B. This drawing is subject to change without notice.  
 C. Body dimensions do not include mold flash or protrusion not to exceed 0,15.  
 D. Falls within JEDEC MO-150

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