LM4562

LM4562 Dual High Performance, High Fidelity Audio Operational Amplifier



Literature Number: SNAS326I



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Dual High Performance, High Fidelity Audio Operational Amplifier

General Description

The LM4562 is part of the ultra-low distortion, low noise, high slew rate operational amplifier series optimized and fully specified for high performance, high fidelity applications. Combining advanced leading-edge process technology with state-of-the-art circuit design, the LM4562 audio operational amplifiers deliver superior audio signal amplification for outstanding audio performance. The LM4562 combines extremely low voltage noise density (2.7nV/ $\sqrt{\text{Hz}}$) with vanishingly low THD+N (0.00003%) to easily satisfy the most demanding audio applications. To ensure that the most challenging loads are driven without compromise, the LM4562 has a high slew rate of $\pm 20\text{V}/\mu\text{s}$ and an output current capability of $\pm 26\text{mA}$. Further, dynamic range is maximized by an output stage that drives $2k\Omega$ loads to within 1V of either power supply voltage and to within 1.4V when driving 600 Ω loads.

The LM4562's outstanding CMRR (120dB), PSRR (120dB), and $V_{\rm OS}$ (0.1mV) give the amplifier excellent operational amplifier DC performance.

The LM4562 has a wide supply range of $\pm 2.5 \text{V}$ to $\pm 17 \text{V}$. Over this supply range the LM4562's input circuitry maintains excellent common-mode and power supply rejection, as well as maintaining its low input bias current. The LM4562 is unity gain stable. This Audio Operational Amplifier achieves outstanding AC performance while driving complex loads with values as high as 100pF.

The LM4562 is available in 8–lead narrow body SOIC, 8–lead Plastic DIP, and 8–lead Metal Can TO-99. Demonstration boards are available for each package.

Key Specifications

■ Power Supply Voltage Range ±2.5V to ±17V

THD+N ($A_V = 1, V_{OUT} = 3V_{BMS}, f_{IN} = 1kHz$)

| $R_L = 2k\Omega$ | 0.00003% (typ) |
|--|---------------------------------|
| $R_L = 600\Omega$ | 0.00003% (typ) |
| ■ Input Noise Density | 2.7nV/ $\sqrt{\text{Hz}}$ (typ) |
| ■ Slew Rate | ±20V/μs (typ) |
| ■ Gain Bandwidth Product | 55MHz (typ) |
| ■ Open Loop Gain (R _L = 600Ω) | 140dB (typ) |
| ■ Input Bias Current | 10nA (typ) |
| ■ Input Offset Voltage | 0.1mV (typ) |
| ■ DC Gain Linearity Error | 0.000009% |

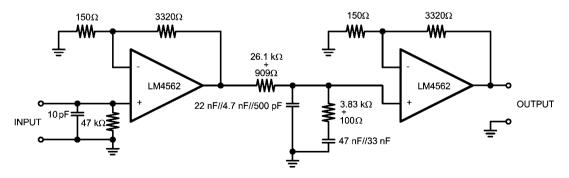
Features

- Easily drives 600Ω loads
- Optimized for superior audio signal fidelity
- Output short circuit protection
- PSRR and CMRR exceed 120dB (typ)
- SOIC, DIP, TO-99 metal can packages

Applications

- Ultra high quality audio amplification
- High fidelity preamplifiers
- High fidelity multimedia
- State of the art phono pre amps
- High performance professional audio
- High fidelity equalization and crossover networks
- High performance line drivers
- High performance line receivers
- High fidelity active filters

Typical Application

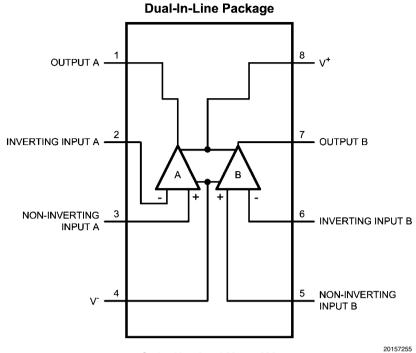


Note: 1% metal film resistors, 5% polypropylene capacitors

Passively Equalized RIAA Phono Preamplifier

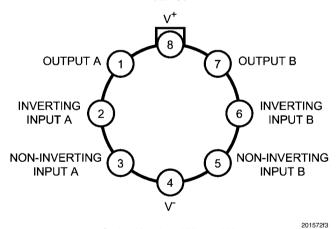
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Connection Diagrams



Order Number LM4562MA See NS Package Number — M08A Order Number LM4562NA See NS Package Number — N08E

Metal Can



Order Number LM4562HA See NS Package Number — H08C

Absolute Maximum Ratings (Note 1, Note

2)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Power Supply Voltage $(V_S = V^+ - V^-)$ 36V Storage Temperature -65° C to 150° C Input Voltage $(V^-) - 0.7V$ to $(V^+) + 0.7V$ Output Short Circuit (Note 3) Continuous Power Dissipation Internally Limited ESD Susceptibility (Note 4) 2000V

ESD Susceptibility (Note 5) Pins 1, 4, 7 and 8 200V Pins 2, 3, 5 and 6 100V Junction Temperature 150°C Thermal Resistance θ_{JA} (SO) 145°C/W θ_{JA} (NA) 102°C/W θ_{JA} (HA) 150°C/W θ_{JC} (HA) 35°C/W Temperature Range $\mathsf{T}_{\mathsf{MIN}} \leq \mathsf{T}_{\mathsf{A}} \leq \mathsf{T}_{\mathsf{MAX}}$ -40°C $\leq T_A \leq 85$ °C Supply Voltage Range $\pm 2.5 \text{V} \le \text{V}_{\text{S}} \le \pm 17 \text{V}$

Electrical Characteristics for the LM4562 (*Note 1, Note 2*) The specifications apply for $V_S = \pm 15V$,

 $\rm R_L$ = 2kΩ, $\rm f_{IN}$ = 1kHz, $\rm T_A$ = 25°C, unless otherwise specified.

| | | | LM4562 | | Units |
|---------------------------|---------------------------------------|---|----------|------------|---------------|
| Symbol | Parameter | Conditions | Typical | Limit | (Limits) |
| | | | (Note 6) | (Note 7) | (Lillits) |
| | | $A_V = 1$, $V_{OUT} = 3V_{rms}$ | | | |
| THD+N | Total Harmonic Distortion + Noise | $R_L = 2k\Omega$ | 0.00003 | | % (max) |
| | | $R_L = 600\Omega$ | 0.00003 | 0.00009 | |
| IMD | Intermedial Distantian | $A_V = 1$, $V_{OUT} = 3V_{RMS}$ | 0.00005 | | 0/ |
| טואו | Intermodulation Distortion | Two-tone, 60Hz & 7kHz 4:1 | 0.00005 | | % |
| GBWP | Gain Bandwidth Product | | 55 | 45 | MHz (min |
| SR | Slew Rate | | ±20 | ±15 | V/µs (min |
| | | $V_{OUT} = 1V_{P-P}, -3dB$ | | | |
| FPBW | Full Power Bandwidth | referenced to output magnitude | 10 | | MHz |
| | | at f = 1kHz | | | |
| t _s | Settling time | $A_V = -1$, 10V step, $C_L = 100pF$ | 1.2 | | μs |
| 'S | | 0.1% error range | 1.2 | 1.2 | |
| | Equivalent Input Noise Voltage | f _{BW} = 20Hz to 20kHz | 0.34 | 0.65 | μV_{RMS} |
| 2 | | | | | (max) |
| P _n | Equivalent Input Naise Density | f = 1kHz | 2.7 | 4.7 | nV/√Hz |
| | Equivalent Input Noise Density | f = 10Hz | 6.4 | | (max) |
| i _n | Current Noise Density | f = 1kHz | 1.6 | | pAJ√Hz |
| | | f = 10Hz | 3.1 | | par√nz |
| V _{OS} | Offset Voltage | | ±0.1 | ±0.7 | mV (max |
| ΔV _{OS} /ΔTemp | Average Input Offset Voltage Drift vs | -40°C ≤ T _A ≤ 85°C | 0.2 | | μV/°C |
| ΔV _{OS} /ΔTemp | Temperature | | | | μν/ Ο |
| PSRR | Average Input Offset Voltage Shift vs | $\Delta V_S = 20V \text{ (Note 8)}$ | 120 | 110 | dB (min) |
| - Orini | Power Supply Voltage | $\Delta V_{\rm S} = 20V$ (Note 8) | 120 | 110 | dB (IIIII) |
| ISO _{CH-CH} | Channel-to-Channel Isolation | f _{IN} = 1kHz | 118 | | dB |
| СН-СН | Charmer-to-Charmer isolation | f _{IN} = 20kHz | 112 | | ub. |
| В | Input Bias Current | $V_{CM} = 0V$ | 10 | 72 | nA (max) |
| ΔI _{OS} /ΔTemp | Input Bias Current Drift vs | 40°C < T < 95°C | 0.1 | | nA/°C |
| Δι _{OS} /Δ ι emp | Temperature | -40°C ≤ T _A ≤ 85°C | 0.1 | | TIA/ O |
| os | Input Offset Current | $V_{CM} = 0V$ | 11 | 65 | nA (max) |
| V _{IN-CM} | Common-Mode Input Voltage Range | | +14.1 | (V+) - 2.0 | V (min) |
| IN-CM | | | -13.9 | (V-) + 2.0 | V (111111) |
| CMRR | Common-Mode Rejection | -10V <vcm<10v< td=""><td>120</td><td>110</td><td>dB (min)</td></vcm<10v<> | 120 | 110 | dB (min) |
| 7 | Differential Input Impedance | | 30 | | kΩ |
| Z _{IN} | Common Mode Input Impedance | -10V <vcm<10v< td=""><td>1000</td><td></td><td>МΩ</td></vcm<10v<> | 1000 | | МΩ |

| | | | LM4562 | | |
|---------------------|-------------------------------------|--|----------|----------|----------|
| Symbol | Parameter | Conditions | Typical | Limit | Units |
| | | | (Note 6) | (Note 7) | (Limits) |
| | | $-10V$ <vout<10v, r<sub="">L = 600Ω</vout<10v,> | 140 | 125 | |
| A _{VOL} | Open Loop Voltage Gain | $-10V$ <vout<10v, r<sub="">L = $2k\Omega$</vout<10v,> | 140 | | dB (min) |
| | | $-10V$ <vout<10v, r<sub="">L = 10kΩ</vout<10v,> | 140 | | |
| | | $R_L = 600\Omega$ | ±13.6 | ±12.5 | |
| V _{OUTMAX} | Maximum Output Voltage Swing | $R_L = 2k\Omega$ | ±14.0 | | V (min) |
| | | $R_L = 10k\Omega$ | ±14.1 | | |
| I _{OUT} | Output Current | $R_L = 600\Omega, V_S = \pm 17V$ | ±26 | ±23 | mA (min) |
| I _{OUT-CC} | Instantaneous Short Circuit Current | | +53 | | mA |
| | | f _{IN} = 10kHz | -42 | | |
| R _{OUT} | Output Impedance | Closed-Loop | 0.01 | | Ω |
| 301 | | Open-Loop | 13 | | |
| C _{LOAD} | Capacitive Load Drive Overshoot | 100pF | 16 | | % |
| I _S | Total Quiescent Current | I _{OUT} = 0mA | 10 | 12 | mA (max) |

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur.

Note 2: Operating Ratings indicate conditions for which the device is functional, but do not guarantee specific performance limits. For guaranteed specifications and test conditions, see the Electrical Characteristics. The guaranteed specifications apply only for the test conditions listed. Some performance characteristics may degrade when the device is not operated under the listed test conditions.

Note 3: Amplifier output connected to GND, any number of amplifiers within a package.

Note 4: Human body model, 100pF discharged through a 1.5k Ω resistor.

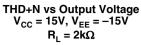
Note 5: Machine Model ESD test is covered by specification EIAJ IC-121-1981. A 200pF cap is charged to the specified voltage and then discharged directly into the IC with no external series resistor (resistance of discharge path must be under 50Ω).

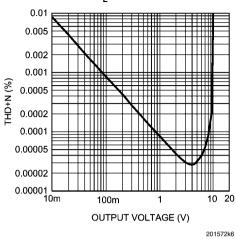
Note 6: Typical specifications are specified at +25°C and represent the most likely parametric norm.

Note 7: Tested limits are guaranteed to National's AOQL (Average Outgoing Quality Level).

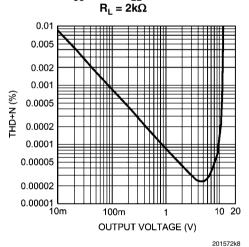
 $\textbf{Note 8: PSRR is measured as follows: V_{OS} is measured at two supply voltages, $\pm 5V$ and $\pm 15V$. PSRR = |$20log($\Delta V_{OS}$/$\Delta V_S$)|$ l.}$

Typical Performance Characteristics

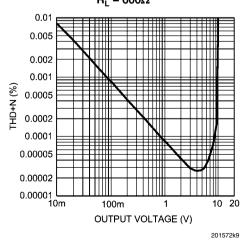




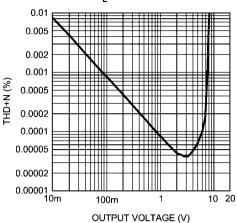
THD+N vs Output Voltage $V_{CC} = 17V$, $V_{EE} = -17V$



THD+N vs Output Voltage V_{CC} = 15V, V_{EE} = -15V R_L = 600 Ω

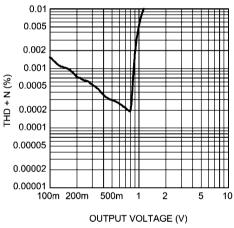


THD+N vs Output Voltage V_{CC} = 12V, V_{EE} = -12V R_L = 2k Ω



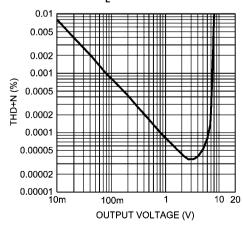
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THD+N vs Output Voltage V_{CC} = 2.5V, V_{EE} = -2.5V R_L = $2k\Omega$

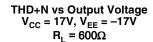


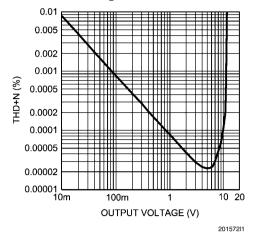
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THD+N vs Output Voltage $V_{CC} = 12V, \, V_{EE} = -12V \\ R_L = 600\Omega$

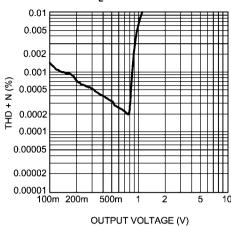


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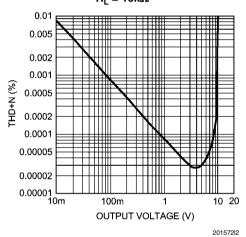


THD+N vs Output Voltage $\begin{aligned} \text{V}_{\text{CC}} &= 2.5 \text{V}, \, \text{V}_{\text{EE}} = -2.5 \text{V} \\ \text{R}_{\text{L}} &= 600 \Omega \end{aligned}$

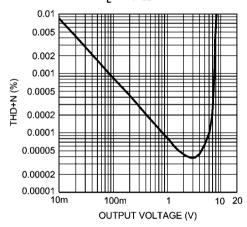


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THD+N vs Output Voltage V_{CC} = 15V, V_{EE} = -15V R_L = 10k Ω

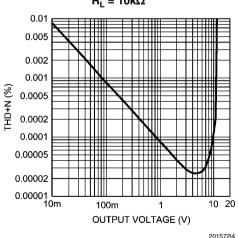


THD+N vs Output Voltage $V_{CC} = 12V, \, V_{EE} = -12V \\ R_L = 10k\Omega$

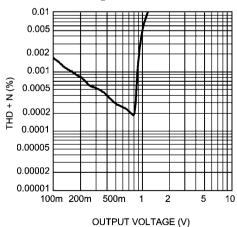


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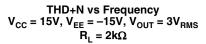
THD+N vs Output Voltage V_{CC} = 17V, V_{EE} = -17V R_L = 10k Ω

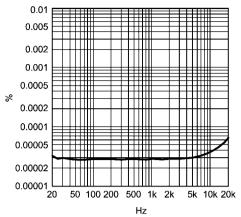


THD+N vs Output Voltage V_{CC} = 2.5V, V_{EE} = -2.5V R_L = 10k Ω



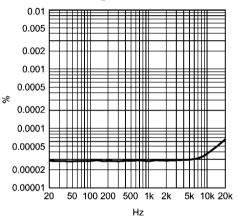
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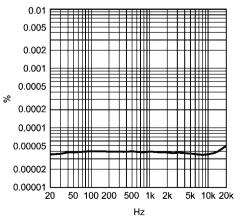
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THD+N vs Frequency V_{CC} = 17V, V_{EE} = -17V, V_{OUT} = $3V_{RMS}$ R_L = $2k\Omega$



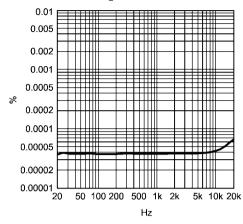
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THD+N vs Frequency
$$\begin{aligned} &V_{CC} = 12V, \, V_{EE} = -12V, \, V_{OUT} = 3V_{RMS} \\ &R_L = 600\Omega \end{aligned}$$



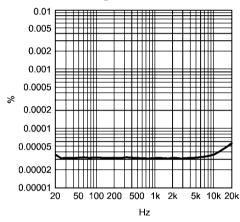
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$$\begin{aligned} & \text{THD+N vs Frequency} \\ V_{\text{CC}} &= 12 V, \, V_{\text{EE}} = -12 V, \, V_{\text{OUT}} = 3 V_{\text{RMS}} \\ & R_1 = 2 k \Omega \end{aligned}$$



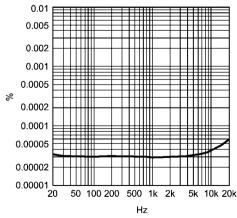
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THD+N vs Frequency
$$\begin{aligned} V_{CC} = 15V, \, V_{EE} = -15V, \, V_{OUT} = 3V_{RMS} \\ R_L = 600\Omega \end{aligned}$$

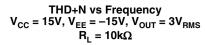


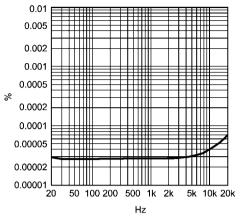
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THD+N vs Frequency
$$\begin{aligned} V_{CC} &= 17V, \, V_{EE} = -17V, \, V_{OUT} = 3V_{RMS} \\ R_L &= 600\Omega \end{aligned}$$



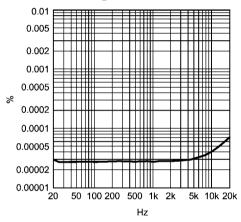
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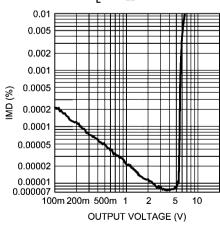
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THD+N vs Frequency V_{CC} = 17V, V_{EE} = -17V, V_{OUT} = $3V_{RMS}$ R_L = $10k\Omega$



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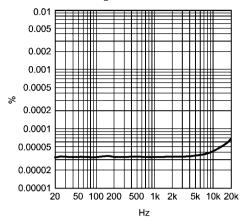
$$\begin{split} & \text{IMD vs Output Voltage} \\ & \text{V}_{\text{CC}} = 12\text{V}, \, \text{V}_{\text{EE}} = -12\text{V} \\ & \text{R}_{\text{L}} = 2\text{k}\Omega \end{split}$$



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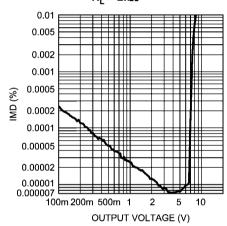
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THD+N vs Frequency
$$\begin{aligned} V_{CC} &= 12V, \, V_{EE} = -12V, \, V_{OUT} = 3V_{RMS} \\ R_{_{I}} &= 10k\Omega \end{aligned}$$



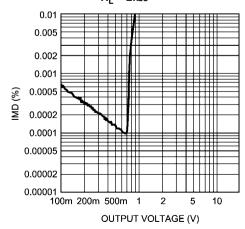
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$$\begin{split} & \text{IMD vs Output Voltage} \\ & \text{V}_{\text{CC}} = 15\text{V}, \, \text{V}_{\text{EE}} = -15\text{V} \\ & \text{R}_{\text{L}} = 2\text{k}\Omega \end{split}$$

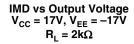


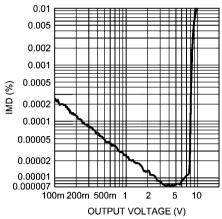
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IMD vs Output Voltage V_{CC} = 2.5V, V_{EE} = -2.5V R_L = $2k\Omega$



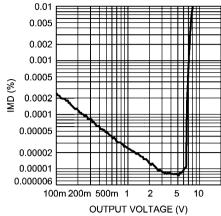
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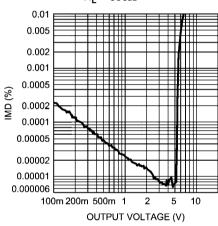
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IMD vs Output Voltage V_{CC} = 15V, V_{EE} = -15V R_L = 600Ω



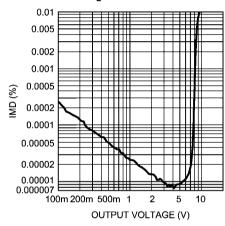
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$\begin{array}{c} \text{IMD vs Output Voltage} \\ V_{\text{CC}} = 12V, \, V_{\text{EE}} = -12V \\ R_{\text{L}} = 600\Omega \end{array}$



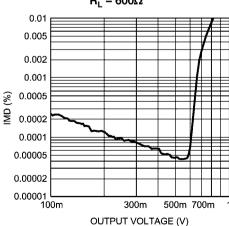
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IMD vs Output Voltage V_{CC} = 17V, V_{EE} = -17V R_L = 600 Ω



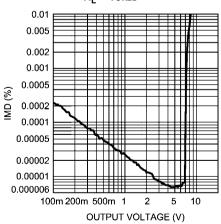
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IMD vs Output Voltage V_{CC} = 2.5V, V_{EE} = -2.5V R_L = 600 Ω

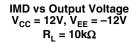


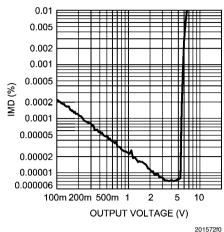
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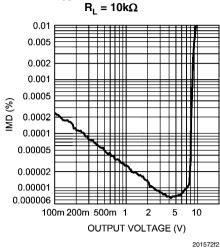
IMD vs Output Voltage V_{CC} = 15V, V_{EE} = -15V R_L = 10k Ω



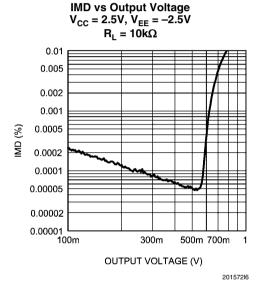
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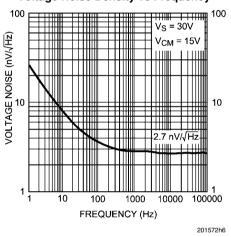


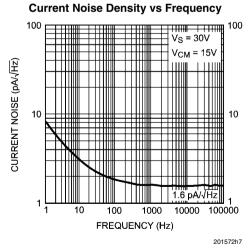


IMD vs Output Voltage $V_{CC} = 17V$, $V_{EE} = -17V$

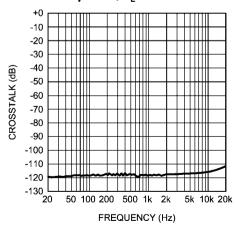


Voltage Noise Density vs Frequency

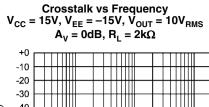


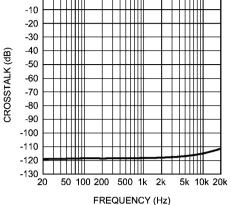


Crosstalk vs Frequency V_{CC} = 15V, V_{EE} = -15V, V_{OUT} = $3V_{RMS}$ $A_V = 0$ dB, $R_L = 2k\Omega$

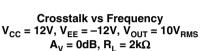


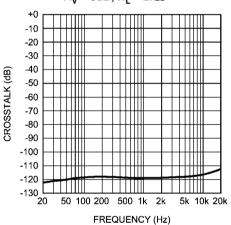
201572c8



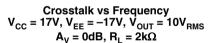


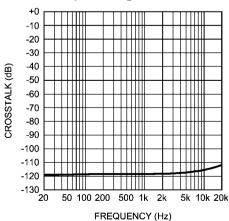
201572c9





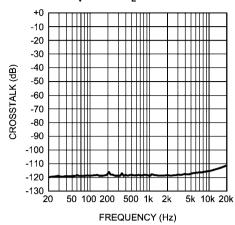
201572c7





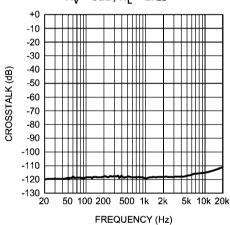
201572d1

$$\begin{array}{c} \text{Crosstalk vs Frequency} \\ V_{\text{CC}} = 12V, V_{\text{EE}} = -12V, V_{\text{OUT}} = 3V_{\text{RMS}} \\ A_{V} = 0dB, R_{L} = 2k\Omega \end{array}$$



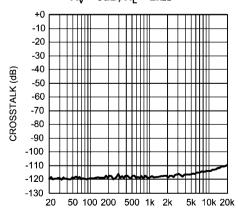
201572c6

 $\begin{aligned} & \text{Crosstalk vs Frequency} \\ V_{\text{CC}} &= 17V, V_{\text{EE}} = -17V, V_{\text{OUT}} = 3V_{\text{RMS}} \\ A_{_{V}} &= 0\text{dB}, R_{_{L}} = 2k\Omega \end{aligned}$



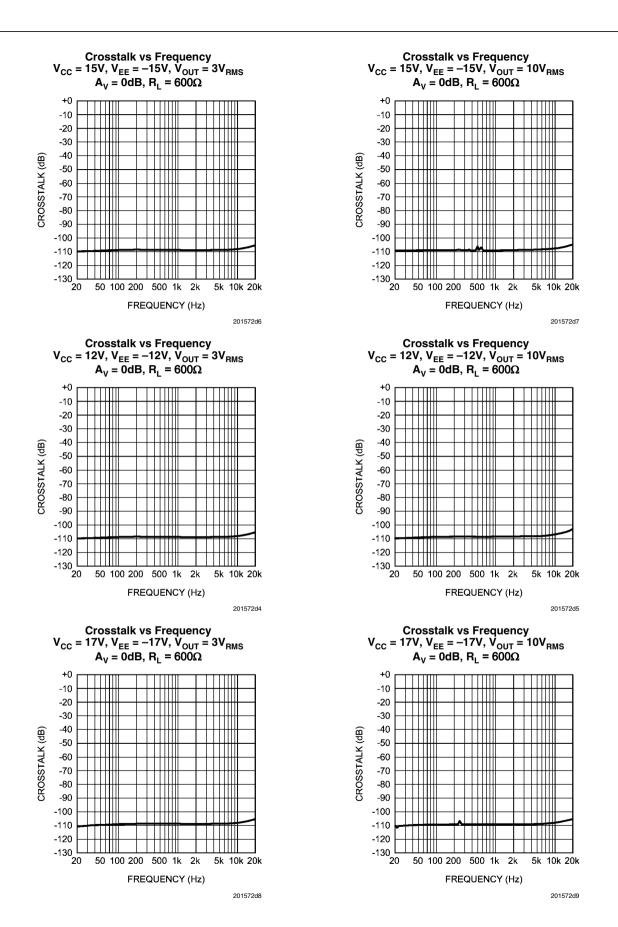
201572d0

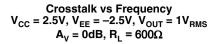
Crosstalk vs Frequency
$$V_{CC}$$
 = 2.5V, V_{EE} = -2.5V, V_{OUT} = 1 V_{RMS} A_V = 0dB, R_I = 2k Ω

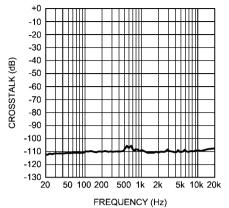


FREQUENCY (Hz)

201572n8

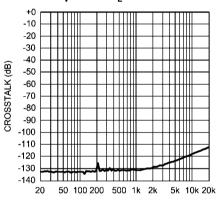






201572d2

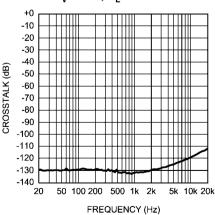
Crosstalk vs Frequency V_{CC} = 15V, V_{EE} = -15V, V_{OUT} = 10V $_{RMS}$ $A_{_{V}}$ = 0dB, $R_{_{L}}$ = 10k Ω



FREQUENCY (Hz)

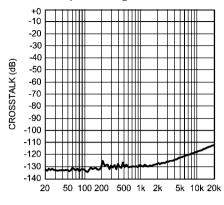
201572n7

 $\begin{array}{l} \text{Crosstalk vs Frequency} \\ \text{V}_{\text{CC}} = 12\text{V}, \, \text{V}_{\text{EE}} = -12\text{V}, \, \text{V}_{\text{OUT}} = 10\text{V}_{\text{RMS}} \\ \text{A}_{\text{V}} = 0\text{dB}, \, \text{R}_{\text{L}} = 10\text{k}\Omega \end{array}$



201572n6

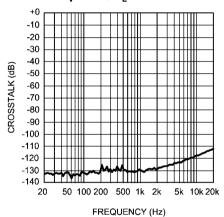
 $\begin{array}{c} \text{Crosstalk vs Frequency} \\ V_{\text{CC}} = 15\text{V}, V_{\text{EE}} = -15\text{V}, V_{\text{OUT}} = 3V_{\text{RMS}} \\ A_{\text{V}} = 0\text{dB}, \ R_{\text{L}} = 10\text{k}\Omega \end{array}$



FREQUENCY (Hz)

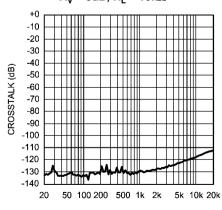
20157200

Crosstalk vs Frequency V_{CC} = 12V, V_{EE} = -12V, V_{OUT} = 3 V_{RMS} A_{v} = 0dB, R_{L} = 10k Ω



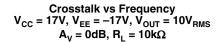
201572n9

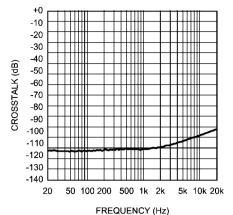
 $\begin{aligned} & \text{Crosstalk vs Frequency} \\ V_{\text{CC}} &= 17\text{V}, V_{\text{EE}} = -17\text{V}, V_{\text{OUT}} = 3\text{V}_{\text{RMS}} \\ & \text{A}_{\text{V}} = 0\text{dB}, \, \text{R}_{\text{L}} = 10\text{k}\Omega \end{aligned}$



FREQUENCY (Hz)

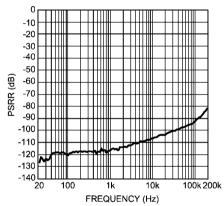
201572n5





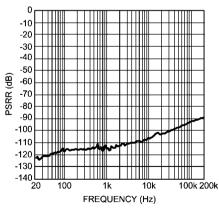
201572n3

$\begin{array}{c} \text{PSRR+ vs Frequency} \\ \text{V}_{\text{CC}} = 15\text{V}, \, \text{V}_{\text{EE}} = -15\text{V} \\ \text{R}_{\text{L}} = 10\text{k}\Omega, \, \text{f} = 200\text{kHz}, \, \text{V}_{\text{RIPPLE}} = 200\text{mVpp} \end{array}$



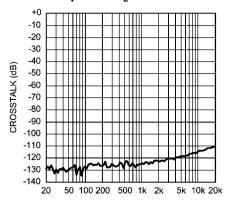
201572p1

 $\begin{array}{c} \text{PSRR+ vs Frequency} \\ \text{V}_{\text{CC}} = 15\text{V}, \, \text{V}_{\text{EE}} = -15\text{V} \\ \text{R}_{\text{L}} = 2\text{k}\Omega, \, \text{f} = 200\text{kHz}, \, \text{V}_{\text{RIPPLE}} = 200\text{mVpp} \end{array}$



201572p2

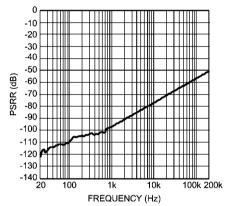
Crosstalk vs Frequency
$$V_{CC}$$
 = 2.5V, V_{EE} = -2.5V, V_{OUT} = 1 V_{RMS} A_V = 0dB, R_I = 10k Ω



FREQUENCY (Hz)

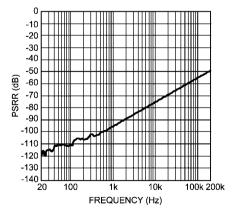
201572n4

PSRR- vs Frequency V_{CC} = 15V, V_{EE} = -15V R_L = 10k Ω , f = 200kHz, V_{RIPPLE} = 200mVpp



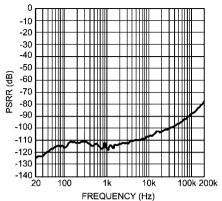
201572p4

 $\begin{array}{c} \text{PSRR- vs Frequency} \\ \text{V}_{\text{CC}} = 15\text{V}, \, \text{V}_{\text{EE}} = -15\text{V} \\ \text{R}_{\text{L}} = 2k\Omega, \, \text{f} = 200\text{kHz}, \, \text{V}_{\text{RIPPLE}} = 200\text{mVpp} \end{array}$



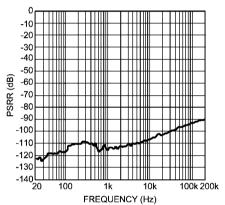
201572p5





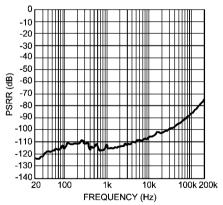
201572p0

PSRR+ vs Frequency V_{CC} = 12V, V_{EE} = -12V R_L = 10k Ω , f = 200kHz, V_{RIPPLE} = 200mVpp



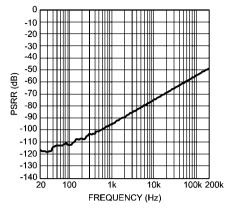
201572p7

 $\label{eq:psrr} \begin{array}{c} \text{PSRR+ vs Frequency} \\ \text{V}_{\text{CC}} = 12\text{V}, \, \text{V}_{\text{EE}} = -12\text{V} \\ \text{R}_{\text{L}} = 2\text{k}\Omega, \, \text{f} = 200\text{kHz}, \, \text{V}_{\text{RIPPLE}} = 200\text{mVpp} \end{array}$



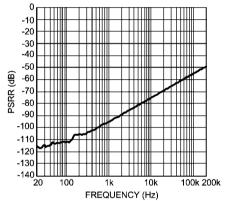
201572p8

PSRR- vs Frequency $V_{CC}=15V,\,V_{EE}=-15V$ $R_L=600\Omega,\,f=200kHz,\,V_{RIPPLE}=200mVpp$



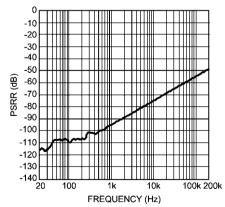
201572p3

 $\begin{array}{c} \text{PSRR- vs Frequency} \\ \text{V}_{\text{CC}} = 12\text{V}, \, \text{V}_{\text{EE}} = -12\text{V} \\ \text{R}_{\text{L}} = 10\text{k}\Omega, \, \text{f} = 200\text{kHz}, \, \text{V}_{\text{RIPPLE}} = 200\text{mVpp} \end{array}$



201572q0

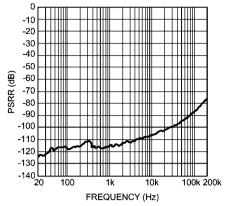
 $\begin{array}{c} \text{PSRR- vs Frequency} \\ \text{V}_{\text{CC}} = 12\text{V}, \, \text{V}_{\text{EE}} = -12\text{V} \\ \text{R}_{\text{L}} = 2\text{k}\Omega, \, \text{f} = 200\text{kHz}, \, \text{V}_{\text{RIPPLE}} = 200\text{mVpp} \end{array}$



201572q1

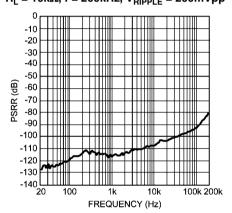
15





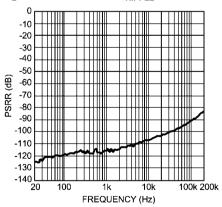
201572p6

 $\begin{array}{c} \text{PSRR+ vs Frequency} \\ \text{V}_{\text{CC}} = 17\text{V}, \, \text{V}_{\text{EE}} = -17\text{V} \\ \text{R}_{\text{L}} = 10\text{k}\Omega, \, \text{f} = 200\text{kHz}, \, \text{V}_{\text{RIPPLE}} = 200\text{mVpp} \end{array}$



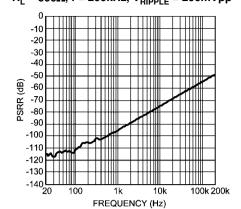
201572q9

 $\begin{array}{c} \text{PSRR+ vs Frequency} \\ \text{V}_{\text{CC}} = 17\text{V}, \, \text{V}_{\text{EE}} = -17\text{V} \\ \text{R}_{\text{L}} = 2\text{k}\Omega, \, \text{f} = 200\text{kHz}, \, \text{V}_{\text{RIPPLE}} = 200\text{mVpp} \end{array}$



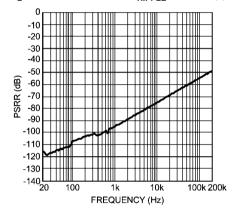
201572r0

 $\begin{array}{c} \text{PSRR- vs Frequency} \\ \text{V}_{\text{CC}} = 12\text{V}, \, \text{V}_{\text{EE}} = -12\text{V} \\ \text{R}_{\text{L}} = 600\Omega, \, \text{f} = 200\text{kHz}, \, \text{V}_{\text{RIPPLE}} = 200\text{mVpp} \end{array}$



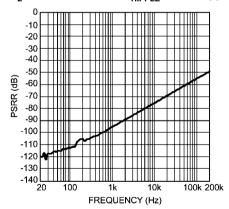
201572p9

PSRR- vs Frequency $V_{CC} = 17V,\, V_{EE} = -17V$ $R_L = 10k\Omega,\, f = 200kHz,\, V_{RIPPLE} = 200mVpp$



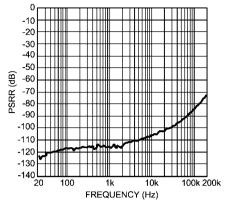
201572r2

 $\begin{array}{c} \text{PSRR- vs Frequency} \\ \text{V}_{\text{CC}} = 17\text{V, V}_{\text{EE}} = -17\text{V} \\ \text{R}_{\text{L}} = 2\text{k}\Omega, \, \text{f} = 200\text{kHz, V}_{\text{RIPPLE}} = 200\text{mVpp} \end{array}$



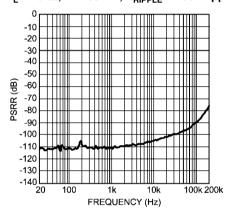
201572r3





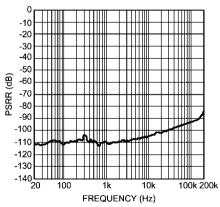
201572q8

PSRR+ vs Frequency V_{CC} = 2.5V, V_{EE} = -2.5V R_L = 10k Ω , f = 200kHz, V_{RIPPLE} = 200mVpp



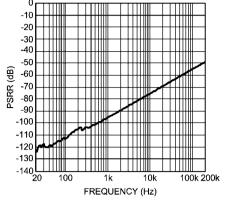
201572q3

 $\begin{aligned} & \text{PSRR+ vs Frequency} \\ & \text{V}_{\text{CC}} = 2.5\text{V}, \text{V}_{\text{EE}} = -2.5\text{V} \\ & \text{R}_{\text{L}} = 2\text{k}\Omega, \text{f} = 200\text{kHz}, \text{V}_{\text{RIPPLE}} = 200\text{mVpp} \end{aligned}$



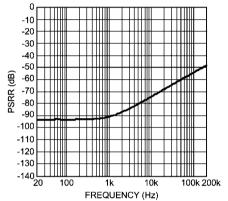
201572q4

 $\begin{array}{c} \text{PSRR- vs Frequency} \\ \text{V}_{\text{CC}} = 17\text{V}, \, \text{V}_{\text{EE}} = -17\text{V} \\ \text{R}_{\text{L}} = 600\Omega, \, \text{f} = 200\text{kHz}, \, \text{V}_{\text{RIPPLE}} = 200\text{mVpp} \end{array}$



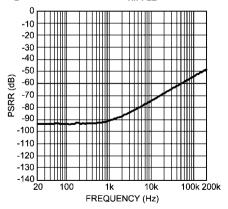
201572r1

 $\begin{array}{c} \text{PSRR- vs Frequency} \\ \text{V}_{\text{CC}} = 2.5\text{V}, \text{V}_{\text{EE}} = -2.5\text{V} \\ \text{R}_{\text{L}} = 10\text{k}\Omega, \text{f} = 200\text{kHz}, \text{V}_{\text{RIPPLE}} = 200\text{mVpp} \end{array}$

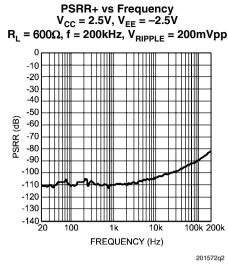


201572q6

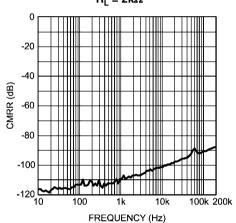
 $\begin{array}{c} \text{PSRR- vs Frequency} \\ \text{V}_{\text{CC}} = 2.5\text{V}, \, \text{V}_{\text{EE}} = -2.5\text{V} \\ \text{R}_{\text{L}} = 2\text{k}\Omega, \, \text{f} = 200\text{kHz}, \, \text{V}_{\text{RIPPLE}} = 200\text{mVpp} \end{array}$



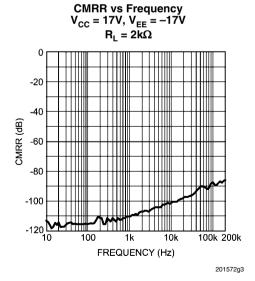
201572q7



CMRR vs Frequency V_{CC} = 15V, V_{EE} = -15V R_L = 2k Ω

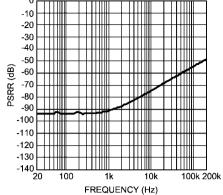


201572g0



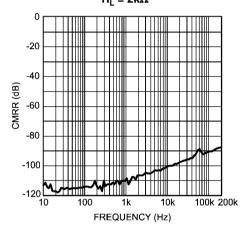
PSRR- vs Frequency $V_{CC} = 2.5V$, $V_{EE} = -2.5V$





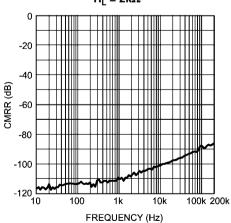
201572q5

CMRR vs Frequency V_{CC} = 12V, V_{EE} = -12V R_L = 2k Ω

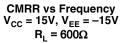


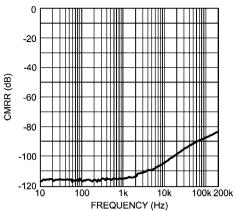
201572f7

CMRR vs Frequency V_{CC} = 2.5V, V_{EE} = -2.5V R_L = $2k\Omega$



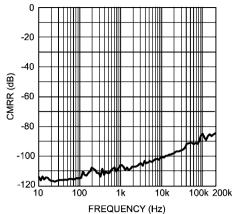
201572f4





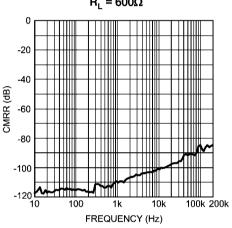
20157209

CMRR vs Frequency V_{CC} = 12V, V_{EE} = -12V R_L = 600 Ω



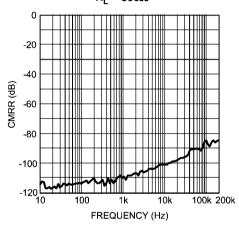
201572f9

CMRR vs Frequency V_{CC} = 17V, V_{EE} = -17V R_L = 600 Ω



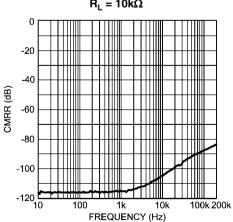
201572g5

CMRR vs Frequency
$$V_{CC}$$
 = 2.5V, V_{EE} = -2.5V R_L = 600 Ω



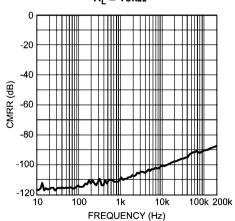
201572f6

CMRR vs Frequency V_{CC} = 15V, V_{EE} = -15V R_L = 10k Ω

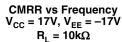


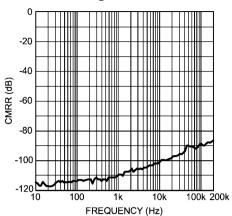
20157208

CMRR vs Frequency V_{CC} = 12V, V_{EE} = -12V R_L = 10k Ω



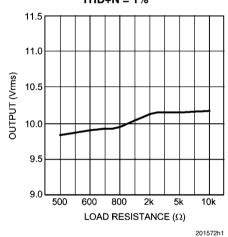
201572f8



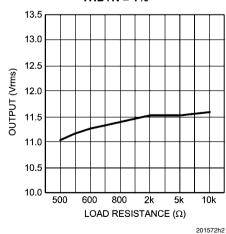


201572g4

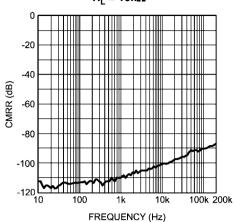
Output Voltage vs Load Resistance $V_{DD} = 15V, \, V_{EE} = -15V$ THD+N = 1%



Output Voltage vs Load Resistance V_{DD} = 17V, V_{EE} = -17V THD+N = 1%

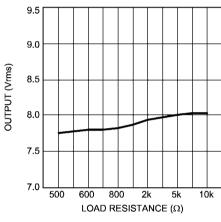


CMRR vs Frequency V_{CC} = 2.5V, V_{EE} = -2.5V R_L = 10k Ω



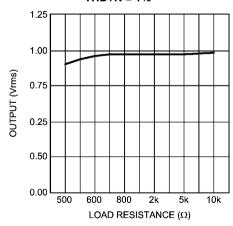
201572f5

Output Voltage vs Load Resistance V_{DD} = 12V, V_{EE} = -12V THD+N = 1%



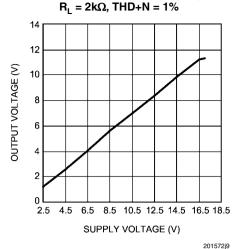
201572h0

Output Voltage vs Load Resistance V_{DD} = 2.5V, V_{EE} = -2.5V THD+N = 1%

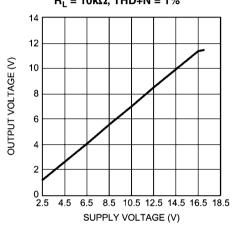


201572g9

Output Voltage vs Supply Voltage



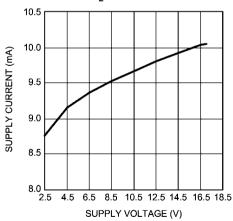
Output Voltage vs Supply Voltage $R_L = 10k\Omega$, THD+N = 1%



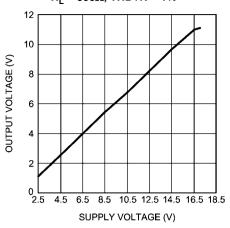
201572k0

201572i5

Supply Current vs Supply Voltage $R_L = 600\Omega$

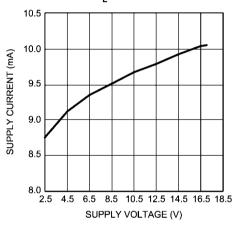


Output Voltage vs Supply Voltage $R_1 = 600\Omega$, THD+N = 1%



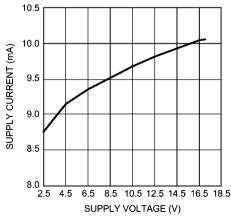
201572j8

Supply Current vs Supply Voltage $R_L = 2k\Omega \label{eq:RL}$



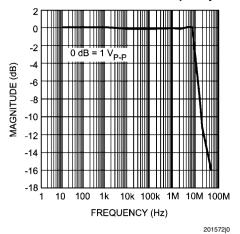
201572i6

Supply Current vs Supply Voltage $R_L = 10k\Omega$

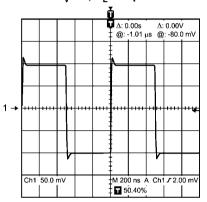


201572j7

Full Power Bandwidth vs Frequency

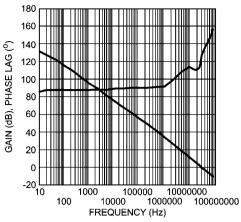


Small-Signal Transient Response $A_V = 1$, $C_L = 10pF$



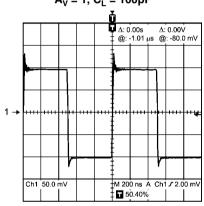
201572i7

Gain Phase vs Frequency



201572j1

Small-Signal Transient Response $A_V = 1$, $C_L = 100pF$



201572i8

Application Information

DISTORTION MEASUREMENTS

The vanishingly low residual distortion produced by LM4562 is below the capabilities of all commercially available equipment. This makes distortion measurements just slightly more difficult than simply connecting a distortion meter to the amplifier's inputs and outputs. The solution, however, is quite simple: an additional resistor. Adding this resistor extends the resolution of the distortion measurement equipment.

The LM4562's low residual distortion is an input referred internal error. As shown in Figure 1, adding the 10Ω resistor connected between the amplifier's inverting and non-inverting inputs changes the amplifier's noise gain. The result is that

the error signal (distortion) is amplified by a factor of 101. Although the amplifier's closed-loop gain is unaltered, the feedback available to correct distortion errors is reduced by 101, which means that measurement resolution increases by 101. To ensure minimum effects on distortion measurements, keep the value of R1 low as shown in Figure 1.

This technique is verified by duplicating the measurements with high closed loop gain and/or making the measurements at high frequencies. Doing so produces distortion components that are within the measurement equipment's capabilities. This datasheet's THD+N and IMD values were generated using the above described circuit connected to an Audio Precision System Two Cascade.

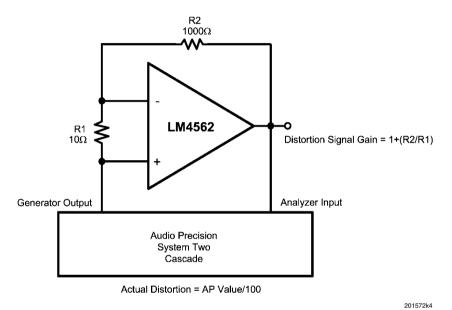
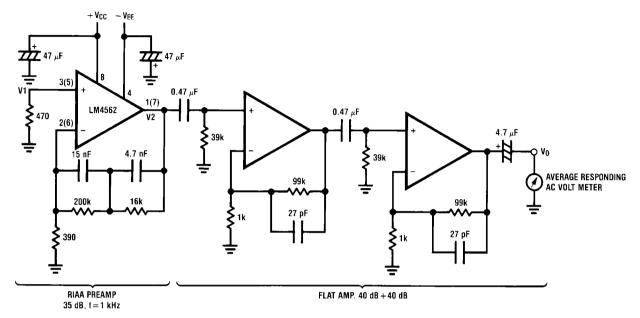


FIGURE 1. THD+N and IMD Distortion Test Circuit

The LM4562 is a high speed op amp with excellent phase margin and stability. Capacitive loads up to 100pF will cause little change in the phase characteristics of the amplifiers and are therefore allowable.

Capacitive loads greater than 100pF must be isolated from the output. The most straightforward way to do this is to put a resistor in series with the output. This resistor will also prevent excess power dissipation if the output is accidentally shorted.

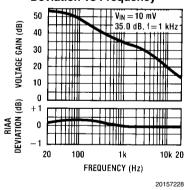


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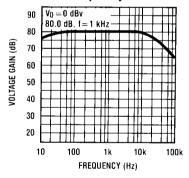
Complete shielding is required to prevent induced pick up from external sources. Always check with oscilloscope for power line noise.

Noise Measurement Circuit Total Gain: 115 dB @f = 1 kHz Input Referred Noise Voltage: $e_n = V0/560,000$ (V)

RIAA Preamp Voltage Gain, RIAA Deviation vs Frequency

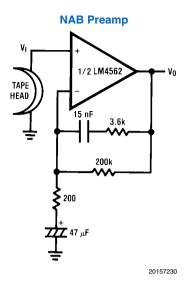


Flat Amp Voltage Gain vs Frequency

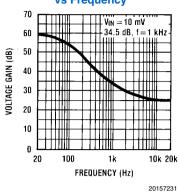


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TYPICAL APPLICATIONS

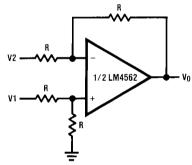


NAB Preamp Voltage Gain vs Frequency



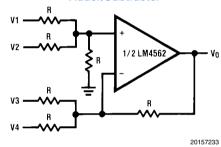
 $A_V = 34.5$ F = 1 kHz $E_n = 0.38 \text{ }\mu\text{V}$ A Weighted

Balanced to Single Ended Converter



20157232

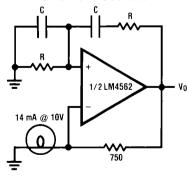
Adder/Subtracter



$$V_0 = V1 + V2 - V3 - V4$$

V_O = V1-V2

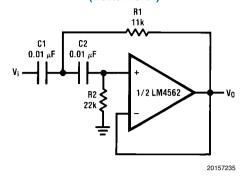
Sine Wave Oscillator



20157234

$$f_0 = \frac{1}{2\pi RC}$$

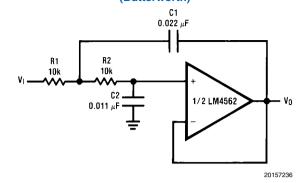
Second Order High Pass Filter (Butterworth)



$$R1 = \frac{\sqrt{2}}{2\omega_{\rm p}C}$$

Illustration is $f_0 = 1 \text{ kHz}$

Second Order Low Pass Filter (Butterworth)

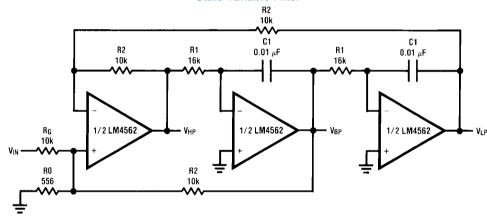


$$C1 = \frac{\sqrt{2}}{\omega_0 R}$$

$$C2 = \frac{C1}{2}$$

Illustration is $f_0 = 1 \text{ kHz}$

State Variable Filter

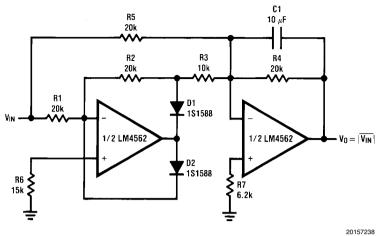


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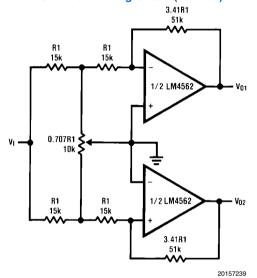
$$f_0 = \frac{1}{2\pi C 1 R 1}, Q = \frac{1}{2} \left(1 + \frac{R2}{R0} + \frac{R2}{RG} \right), A_{BP} = QA_{LP} = QA_{LH} = \frac{R2}{RG}$$

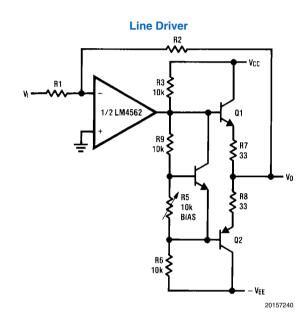
Illustration is $f_0 = 1 \text{ kHz}$, Q = 10, $A_{BP} = 1$

AC/DC Converter

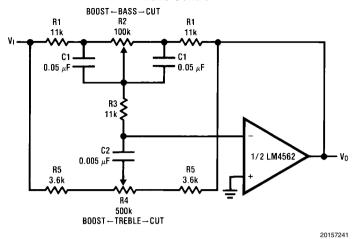


2 Channel Panning Circuit (Pan Pot)





Tone Control



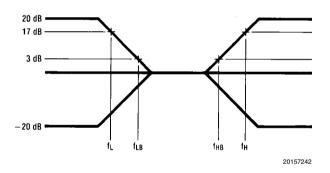
$$\begin{split} f_L &\approx \frac{1}{2\pi\,R\,2\,C\,1}, \, f_{LB} \approx \frac{1}{2\pi\,R\,1\,C\,1} \\ f_H &\approx \frac{1}{2\pi\,R\,5\,C\,2}, \, f_{HB} \approx \frac{1}{2\pi\,(R\,1 + R\,5 + 2R\,3)\,C\,2} \end{split}$$

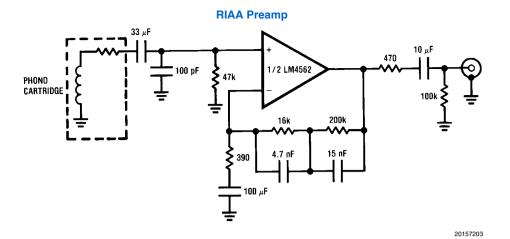
Note: The equations started above are simplifications, providing guidance of general -3dB point values, when the potentiometers are at their null position.

Illustration is:

$$f_L \approx 32 \text{ Hz}, f_{LB} \approx 320 \text{ Hz}$$

 $f_H \approx 11 \text{ kHz}, f_{HB} \approx 1.1 \text{ kHz}$





 $A_v = 35 \text{ dB}$

 $E_n = 0.33 \, \mu V$

S/N = 90 dB

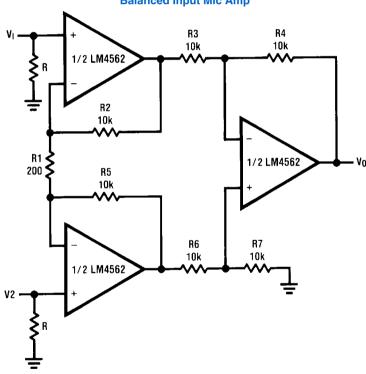
f = 1 kHz

A Weighted

A Weighted, V_{IN} = 10 mV

@f = 1 kHz

Balanced Input Mic Amp



20157243

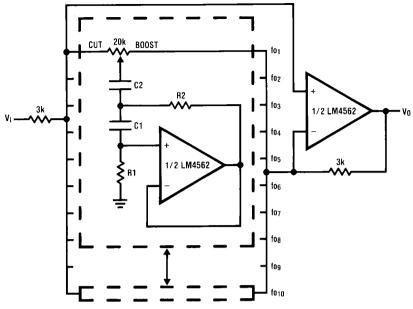
If R2 = R5, R3 = R6, R4 = R7

$$V0 = \left(1 + \frac{2R2}{R1}\right) \frac{R4}{R3} (V2 - V1)$$

Illustration is:

V0 = 101(V2 - V1)

10 Band Graphic Equalizer



20157244

| fo (Hz) | C ₁ | C ₂ | R ₁ | R ₂ |
|---------|----------------|----------------|----------------|----------------|
| 32 | 0.12µF | 4.7µF | 75kΩ | 500Ω |
| 64 | 0.056µF | 3.3µF | 68kΩ | 510Ω |
| 125 | 0.033µF | 1.5µF | 62kΩ | 510Ω |
| 250 | 0.015µF | 0.82µF | 68kΩ | 470Ω |
| 500 | 8200pF | 0.39µF | 62kΩ | 470Ω |
| 1k | 3900pF | 0.22µF | 68kΩ | 470Ω |
| 2k | 2000pF | 0.1µF | 68kΩ | 470Ω |
| 4k | 1100pF | 0.056µF | 62kΩ | 470Ω |
| 8k | 510pF | 0.022µF | 68kΩ | 510Ω |
| 16k | 330pF | 0.012µF | 51kΩ | 510Ω |

Note 9: At volume of change = $\pm 12 \text{ dB}$

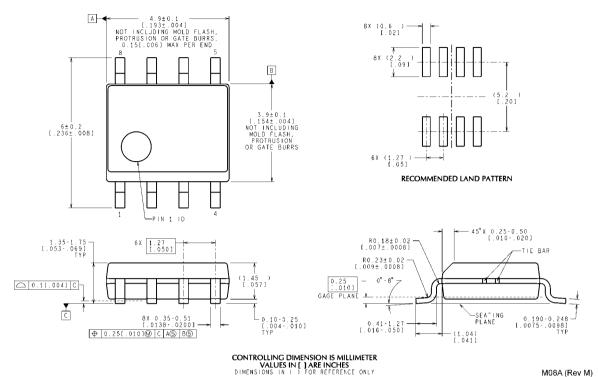
Q = 1.7

Reference: "AUDIO/RADIO HANDBOOK", National Semiconductor, 1980, Page 2-61

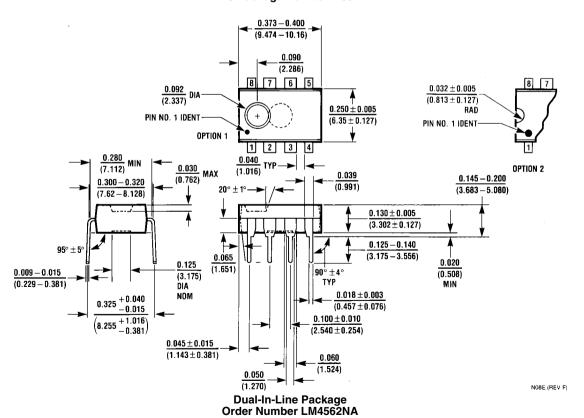
Revision History

| Rev | Date | Description |
|-----|----------|---|
| 1.0 | 08/16/06 | Initial release. |
| 1.1 | 08/22/06 | Updated the Instantaneous Short Circuit Current specification. |
| 1.2 | 09/12/06 | Updated the three ±15V CMRR Typical Performance Curves. |
| 1.3 | 09/26/06 | Updated interstage filter capacitor values on page 1 Typical Application schematic. |
| 1.4 | 05/03/07 | Added the "general note" under the EC table. |
| 1.5 | 10/17/07 | Replaced all the PSRR curves. |
| 1.6 | 01/26/10 | Edited the equations on page 28 (under Tone Control). |

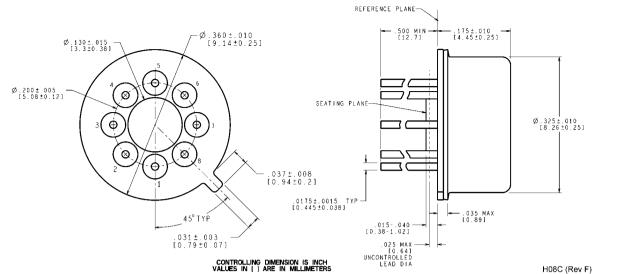
Physical Dimensions inches (millimeters) unless otherwise noted



Narrow SOIC Package Order Number LM4562MA NS Package Number M08A



NS Package Number N08E



TO-99 Metal Can Package Order Number LM4562HA NS Package Number H08C

Notes

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