TL1454A, TL1454AY DUAL-CHANNEL PULSE-WIDTH-MODULATION (PWM) CONTROL CIRCUIT

SLVS423 A- MAY 2002 - REVISED SEPTEMBER 2002

- Two Complete PWM Control Circuits
- Outputs Drive MOSFETs Directly
- Oscillator Frequency . . . 50 kHz to 2 MHz
- 3.6-V to 20-V Supply-Voltage Range
- Low Supply Current . . . 3.5 mA Typ
- Adjustable Dead-Time Control, 0% to 100%
- 1.26-V Reference

description

The TL1454A is a dual-channel pulse-width-modulation (PWM) control circuit, primarily intended for low-power, dc/dc converters. Applications include LCD displays, backlight inverters, notebook computers, and other products requiring small, high-frequency, dc/dc converters.

D, N OR PW PACKAGE (TOP VIEW) 16 REF CT 15 SCP RT Π DTC1 14 DTC2 13 N2+ IN1+ □ 12 IN2-IN1 – ∏ 5 11 \ COMP2 COMP1 [10 V_{CC} GND 17 OUT1 [9 OUT2

Each PWM channel has its own error amplifier, PWM comparator, dead-time control comparator, and MOSFET driver. The voltage reference, oscillator, undervoltage lockout, and short-circuit protection are common to both channels.

Channel 1 is configured to drive n-channel MOSFETs in step-up or flyback converters, and channel 2 is configured to drive p-channel MOSFETs in step-down or inverting converters. The operating frequency is set with an external resistor and an external capacitor, and dead time is continuously adjustable from 0 to 100% duty cycle with a resistive divider network. Soft start can be implemented by adding a capacitor to the dead-time control (DTC) network. The error-amplifier common-mode input range includes ground, which allows the TL1454A to be used in ground-sensing battery chargers as well as voltage converters.

AVAILABLE OPTIONS

	PACKAGED DEVICES [†]					CUID FORM
TA	SMALL OUTLINE (D)	PLASTIC DIP (N)	TSSOP (PW)	SSOP (DB)	SOP-EIAJ (NS)	CHIP FORM (Y)
-20°C to 85°C	TL1454ACD	TL1454ACN	TL1454ACPWR	TL1454ACDB	TL1454ACNS	TL1454AY

† The D, DB and NS packages are available taped and reeled. Add the suffix R to the device name (e.g., TL1454ACDR). The PW package is available only left-end taped and reeled (indicated by the R suffix on the device type; e.g., TL1454ACPWR).



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functional block diagram RT CT | 10 2 16 REF 1.26 V Voltage 1.8 V 2.5 V REF To Internal Circuitry 1.2 V GND VCC osc **PWM** COMP1 Comparator 1 IN1+ 8 OUT1 Amplifier 1 **PWM** COMP2 11 Comparator 2 IN2+ 13 VCC IN2-Error Amplifier 2 UVLO 9 OUT2 and SCP Latch SCP Comparator 2 0.65 V 0.65 V 1 V SCP Comparator 1 1.27 V

14

3

DTC1 DTC2

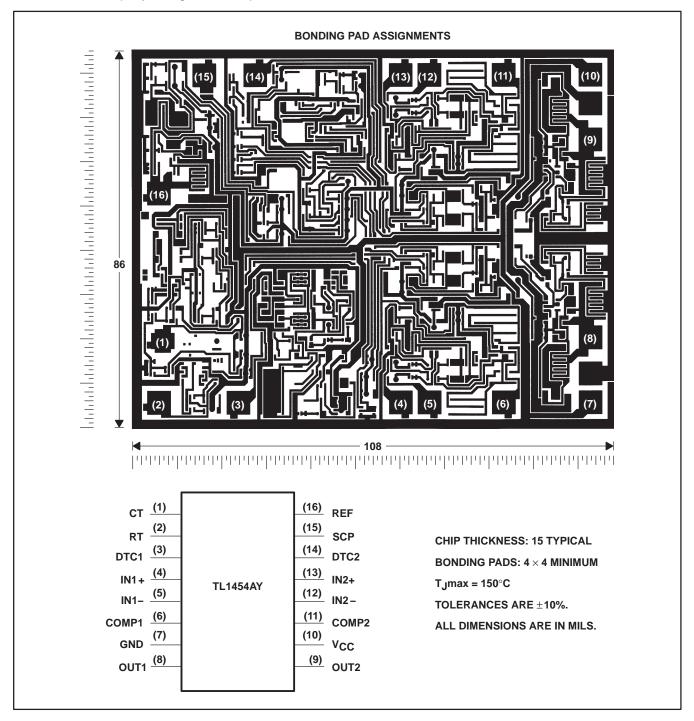
15

SCP



TL1454AY chip information

This device, when properly assembled, displays characteristics similar to the TL1454AC. Thermal compression or ultrasonic bonding may be used on the doped aluminum bonding pads. The chips may be mounted with conductive epoxy or a gold-silicon preform.



theory of operation

reference voltage

A linear regulator operating from V_{CC} generates a 2.5-V supply for the internal circuits and the 1.26-V reference, which can source a maximum of 1 mA for external loads. A small ceramic capacitor (0.047 μ F to 0.1 μ F) between REF and ground is recommended to minimize noise pickup.

error amplifier

The error amplifier generates the error signal used by the PWM to adjust the power-switch duty cycle for the desired converter output voltage. The signal is generated by comparing a sample of the output voltage to the voltage reference and amplifying the difference. An external resistive divider connected between the converter output and ground, as shown in Figure 1, is generally required to obtain the output voltage sample.

The amplifier output is brought out on COMP to allow the frequency response of the amplifier to be shaped with an external RC network to stabilize the feedback loop of the converter. DC loading on the COMP output is limited to $45 \, \mu A$ (the maximum amplifier source current capability).

Figure 1 illustrates the sense-divider network and error-amplifier connections for converters with positive output voltages. The divider network is connected to the noninverting amplifier input because the PWM has a phase inversion; the duty cycle decreases as the error-amplifier output increases.

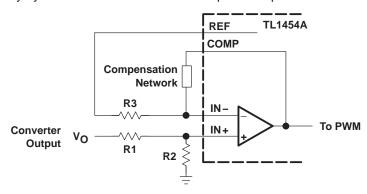


Figure 1. Sense Divider/Error Amplifier Configuration for Converters with Positive Outputs

The output voltage is given by:

$$V_{O} = V_{ref} \left(1 + \frac{R1}{R2} \right)$$

where $V_{ref} = 1.26 \text{ V}$.

The dc source resistance of the error-amplifier inputs should be 10 k Ω or less and approximately matched to minimize output voltage errors caused by the input-bias current. A simple procedure for determining appropriate values for the resistors is to choose a convenient value for R3 (10 k Ω or less) and calculate R1 and R2 using:

$$R_1 = \frac{R_3 V_0}{V_0 - V_{ref}}$$

$$R_2 = \frac{R_3 V_0}{V_{ref}}$$



error amplifier

R1 and R2 should be tight-tolerance ($\pm 1\%$ or better) devices with low and/or matched temperature coefficients to minimize output voltage errors. A device with a $\pm 5\%$ tolerance is suitable for R3.

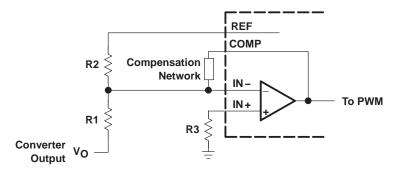


Figure 2. Sense Divider/Error Amplifier Configuration for Converters with Negative Outputs

Figure 2 shows the divider network and error-amplifier configuration for negative output voltages. In general, the comments for positive output voltages also apply for negative outputs. The output voltage is given by:

$$V_O = -\frac{R_1 V_{ref}}{R_2}$$

The design procedure for choosing the resistor value is to select a convenient value for R2 (instead of R3 in the procedure for positive outputs) and calculate R1 and R3 using:

$$R_1 = -\frac{R_2 V_0}{V_{ref}}$$

$$R_3 = \frac{R_1 R_2}{R_1 + R_2}$$

Values in the $10-k\Omega$ to $20-k\Omega$ range work well for R2. R3 can be omitted and the noninverting amplifier connected to ground in applications where the output voltage tolerance is not critical.

oscillator

The oscillator frequency can be set between 50 kHz and 2 MHz with a resistor connected between RT and GND and a capacitor between CT and GND (see Figure 3). Figure 6 is used to determine R_T and C_T for the desired operating frequency. Both components should be tight-tolerance, temperature-stable devices to minimize frequency deviation. A 1% metal-film resistor is recommended for R_T , and a 10%, or better, NPO ceramic capacitor is recommended for C_T .

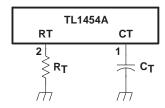


Figure 3. Oscillator Timing



dead-time control (DTC) and soft start

The two PWM channels have independent dead-time control inputs so that the maximum power-switch duty cycles can be limited to less then 100%. The dead-time is set with a voltage applied to DTC; the voltage is typically obtained from a resistive divider connected between the reference and ground as shown in Figure 4. Soft start is implemented by adding a capacitor between REF and DTC.

The voltage, V_{DT}, required to limit the duty cycle to a maximum value is given by:

$$V_{DT} = V_{O(max)} - D(V_{O(max)} - V_{O(min)}) - 0.65$$

where V_{O(max)} and V_{O(min)} are obtained from Figure 9, and D is the maximum duty cycle.

Predicting the regulator startup or rise time is complicated because it depends on many variables, including: input voltage, output voltage, filter values, converter topology, and operating frequency. In general, the output will be in regulation within two time constants of the soft-start circuit. A five-to-ten millisecond time constant usually works well for low-power converters.

The DTC input can be grounded in applications where achieving a 100% duty cycle is desirable, such as a buck converter with a very low input-to-output differential voltage. However, grounding DTC prevents the implementation of soft start, and the output voltage overshoot at power-on is likely to be very large. A better arrangement is to omit R_{DT1} (see Figure 4) and choose R_{DT2} = 47 k Ω . This configuration ensures that the duty cycle can reach 100% and still allows the designer to implement soft start using C_{SS} .

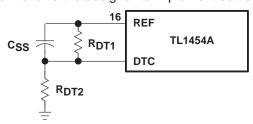


Figure 4. Dead-Time Control and Soft Start

PWM comparator

Each of the PWM comparators has dual inverting inputs. One inverting input is connected to the output of the error amplifier; the other inverting input is connected to the DTC terminal. Under normal operating conditions, when either the error-amplifier output or the dead-time control voltage is higher than that for the PWM triangle wave, the output stage is set inactive (OUT1 low and OUT2 high), turning the external power stage off.

undervoltage-lockout (UVLO) protection

The undervoltage-lockout circuit turns the output circuit off and resets the SCP latch whenever the supply voltage drops too low (to approximately 2.9 V) for proper operation. A hysteresis voltage of 200 mV eliminates false triggering on noise and chattering.

short-circuit protection (SCP)

The TL1454A SCP function prevents damage to the power switches when the converter output is shorted to ground. In normal operation, SCP comparator 1 clamps SCP to approximately 185 mV. When one of the converter outputs is shorted, the error amplifier output (COMP) will be driven below 1 V to maximize duty cycle and force the converter output back up. When the error amplifier output drops below 1 V, SCP comparator 1 releases SCP, and capacitor, C_{SCP} , which is connected between SCP and GND, begins charging. If the error-amplifier output rises above 1 V before C_{SCP} is charged to 1 V, SCP comparator 1 discharges C_{SCP} and normal operation resumes. If C_{SCP} reaches 1 V, SCP comparator 2 turns on and sets the SCP latch, which turns off the output drives and resets the soft-start circuit. The latch remains set until the supply voltage is lowered to 2 V or less, or C_{SCP} is discharged externally.



short-circuit protection (SCP) (continued)

The SCP time-out period must be greater than the converter start-up time or the converter will not start. Because high-value capacitor tolerances tend to be $\pm 20\%$ or more and IC resistor tolerances are loose as well, it is best to choose an SCP time-out period 10-to-15 times greater than the converter startup time. The value of C_{SCP} may be determined using Figure 6, or it can be calculated using:

$$C_{SCP} = \frac{T_{SCP}}{80.3}$$

where C_{SCP} is in μF and T_{SCP} is the time-out period in ms.

output stage

The output stage of the TL1454A is a totem-pole output with a maximum source/sink current rating of 40 mA and a voltage rating of 20 V. The output is controlled by a complementary output AND gate and is turned on (sourcing current for OUT1, sinking current for OUT2) when all the following conditions are met: 1) the oscillator triangle wave voltage is higher than both the DTC voltage and the error-amplifier output voltage, 2) the undervoltage-lockout circuit is inactive, and 3) the short-circuit protection circuit is inactive.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, V _{CC} (see Note 1)	23 V
Error amplifier input voltage: IN1+, IN1-, IN2+, IN2	
Output voltage: OUT1, OUT2	20 V
Continuous output current: OUT1, OUT2	±200 mA
Peak output current: OUT1, OUT2	1 A
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature range, T _A : C suffix	–20°C to 85°C
Storage temperature range, T _{stq}	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltage values are with respect to network GND.

DISSIPATION RATING TABLE

PACKAGE	$T_{\mbox{$A$}} \leq 25^{\circ}\mbox{$C$}$ POWER RATING	DERATING FACTOR ABOVE T _A = 25°C	T _A = 70°C POWER RATING	T _A = 85°C POWER RATING
D	950 mW	7.6 mW/°C	608 mW	494 mW
DB	1000 mW	8.0 mW/°C	640 mW	520 mW
N	1250 mW	10.0 mW/°C	800 mW	650 mW
NS	1953 mW	15.6 mW/°C	1250 mW	1015 mW
PW	500 mW	4.0 mW/°C	320 mW	260 mW



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recommended operating conditions

		MIN	MAX	UNIT
Supply voltage, V _{CC}		3.6	20	V
Error amplifier common-mode input voltage		-0.2	1.45	V
Output voltage, VO			20	V
Output current, IO			±40	mA
COMP source current			-45	μΑ
COMP sink current			100	μΑ
Reference output current			1	mA
COMP dc load resistance		100		kΩ
Timing capacitor, C _T		10	4000	pF
Timing resistor, R _T		5.1	100	kΩ
Oscillator frequency		50	2000	kHz
Operating free-air temperature, TA	TL1454AC	-20	85	°C

electrical characteristics over recommended operating free-air temperature range, V_{CC} = 6 V, f_{osc} = 500 kHz (unless otherwise noted)

reference

	DADAMETED	TEST SONDIT	TONO.	1	UNIT		
	PARAMETER	TEST CONDIT	IONS	MIN	TYP	MAX	UNII
V	Output wells as DEE	$I_O = 1 \text{ mA},$	T _A = 25°C	1.22	1.26	1.32	.,
Vref	/ref Output voltage, REF Input regulation	$I_O = 1 \text{ mA}$		1.20		1.34	V
	Input regulation	$V_{OC} = 3.6 \text{ V to } 20 \text{ V},$	$I_O = 1 \text{ mA}$		2	6	mV
	Output regulation	I _O = 0.1 mA to 1 mA			1	7.5	mV
	Output valtage change with temperature	$T_A = T_{A(min)}$ to 25°C,	$I_O = 1 \text{ mA}$	-12.5	-1.25	12.5	\/
	Output voltage change with temperature	$T_A = 25^{\circ}C \text{ to } 85^{\circ}C,$	$I_O = 1 \text{ mA}$	-12.5	-2.5	12.5	mV
los	Short-circuit output current	V _{ref} = 0 V			30		mA

undervoltage lockout (UVLO)

	DADAMETED	TEST COMPITIONS	TL1454A			LINUT
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
VIT+	Positive-going threshold voltage			2.9		V
VIT-	Negative-going threshold voltage	T _A = 25°C		2.7		V
V _{hys}	Hysteresis, V _{IT+} - V _{IT-}		100	200		mV

short-circuit protection (SCP)

	DADAMETED	TEST SOMBITIONS	TL1454A			
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
VIT	Input threshold voltage	T _A = 25°C	0.93	1	1.07	V
v _{stby} †	Standby voltage	N. a. a. allana	140	185	230	mV
V _I (latched)	Latched-mode input voltage	No pullup		60	120	mV
VIT(COMP)	Comparator threshold voltage	COMP1, COMP2		1		V
	Input source current	$T_A = 25^{\circ}C$, $V_{O(SCP)} = 0$	-5	-15	-20	μΑ

[†] This symbol is not presently listed within EIA/JEDEC standards for semiconductor symbology.



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electrical characteristics over recommended operating free-air temperature range, V_{CC} = 6 V, f_{osc} = 500 kHz (unless otherwise noted) (continued)

oscillator

	DARAMETER	TEST SOND	TEST CONDITIONS		TL1454A		
	PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
fosc	Frequency	$C_T = 120 \text{ pF},$	$R_T = 10 \text{ k}\Omega$		500		kHz
	Standard deviation of frequency				50		kHz
	Frequency change with voltage	$V_{CC} = 3.6 \text{ V to } 20 \text{ V},$	T _A = 25°C		10		kHz
	Francisco de como cráth tomo quetoro	$T_A = T_{A(min)}$ to 25°C			-2	±30	1.11=
	Frequency change with temperature	T _A = 25°C to 85°C			-10	±30	kHz
	Maximum ramp voltage				1.8		V
	Minimum ramp voltage				1.1		V

dead-time control (DTC)

	2424455	T-07 0011710110	Т			
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
.,	V _{IT} Input threshold voltage	Duty cycle = 0%	0.98	1.1	1.22	.,
VIT		Duty cycle = 100%	0.38	0.5	0.62	V
V _I (latched)	Latched-mode input voltage			1.2		V
I _{IB}	Common-mode input bias current	DTC1, IN1+ ≈ 1.2 V			4	μΑ
	Latched-mode (source) current	T _A = 25°C		-100		μΑ

error-amplifier

	24244555	TEST SOUDITIONS	TL1454A			
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{IO}	Input offset voltage				6	mV
I _{IO}	Input offset current	$V_O = 1.25 \text{ V}, V_{IC} = 1.25 \text{ V}$			100	nA
I _{IB}	Input bias current			-160	-500	nA
VICR	Input voltage range	V _{CC} = 3.6 V to 20 V	-0.2 to 1.40			V
Ay	Open-loop voltage gain	$R_{FB} = 200 \text{ k}\Omega$	70	80		dB
	Unity-gain bandwidth			3		MHz
CMRR	Common-mode rejection ratio		60	80		dB
VOM(max)	Positive output voltage swing		2.3	2.43		
VOM(min)	Negative output voltage swing			0.63	0.8	V
I _{O+}	Output sink current	$V_{ID} = -0.1 \text{ V}, V_{O} = 1.20 \text{ V}$	0.1	0.5	·	mA
I _O -	Output source current	$V_{ID} = 0.1 \text{ V}, \qquad V_{O} = 1.80 \text{ V}$	-45	-70		μΑ

output

	DADAMETER	TEST CONDITIONS	TL14			
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
		$I_O = -8 \text{ mA}$	V _{CC} -2	4.5		
	OH High-level output voltage	$I_O = -8 \text{ mA } @ V_{CC} = >10 \text{ V}$	V _{CC} -2.3 V			.,
VOH		$I_O = -40 \text{ mA}$	V _{CC} -2	4.4		V
		$I_O = 40 \text{ mA } @ V_{CC} = >10 \text{ V}$	V _{CC} -2.3 V			
.,		IO = 8 mA		0.1	0.4	.,
VOL	Low-level output voltage	I _O = 40 mA		1.8	2.5	V
t _{rv}	Output voltage rise time	C. 2000 T. T. 25°C		220		20
t _{fV}	Output voltage fall time	$C_L = 2000 \text{ pF}, T_A = 25^{\circ}\text{C}$		220		ns



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electrical characteristics over recommended operating free-air temperature range, V_{CC} = 6 V, f_{osc} = 500 kHz (unless otherwise noted) (continued)

supply current

	DADAMETED	TEST CONDITIONS		TL1454A		
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
ICC(stby)	Standby supply current	RT open, CT = 1.5 V, No load, V _O (COMP1, COMP2) = 1.25 V,		3.1	6	mA
ICC(average)	Average supply current	$R_T = 10 \text{ k}\Omega,$ $C_T = 120 \text{ pF},$ 50% duty cycle, Outputs open		3.5	7	mA

electrical characteristics, V_{CC} = 6 V, f_{osc} = 500 kHz, T_A = 25°C (unless otherwise noted)

reference

	PARAMETER	TEST SOMBITIONS	TL1			
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{ref}	Output voltage, REF	$I_O = 1 \text{ mA}$		1.26		V
	Input regulation	$V_{OC} = 3.6 \text{ V to } 20 \text{ V}, \qquad I_{O} = 1 \text{ mA}$		2		mV
	Output regulation	$I_O = 0.1 \text{ mA to } 1 \text{ mA}$		1		mV
	Output voltage change with temperature	$I_O = 1 \text{ mA}$	-1.25			mV
	Output voltage change with temperature	$I_O = 1 \text{ mA}$		-2.5		IIIV
los	Short-circuit output current	V _{ref} = 0 V		30		mA

undervoltage lockout (UVLO)

	242445752	TEST COMPLETIONS	TI			
PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{IT+}	Positive-going threshold voltage			2.9		V
VIT-	Negative-going threshold voltage	7		2.7		V
V _{hys}	Hysteresis, V _{IT+} - V _{IT}	7		200		mV

short-circuit protection (SCP)

	DARAMETER	TEGT CONDITIONS	TL			
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
VIT	Input threshold voltage			1		V
v _{stby} †	Standby voltage	No pullup		185		mV
VI(latched)	Latched-mode input voltage	No pullup		60		mV
VIT(COMP)	Comparator threshold voltage	COMP1, COMP2		1		V
	Input source current	$V_O(SCP) = 0$		-15		μΑ

[†] This symbol is not presently listed within EIA/JEDEC standards for semiconductor symbology.

oscillator

	PARAMETER	TEGT CONDITIONS	TL1454AY	
	PARAMETER	TEST CONDITIONS	MIN TYP MAX	UNIT
fosc	Frequency	$C_T = 120 \text{ pF}, \qquad R_T = 10 \text{ k}\Omega$	500	kHz
	Standard deviation of frequency		50	kHz
	Frequency change with voltage	V _{CC} = 3.6 V to 20 V	10	kHz
	Francisco de anno vitto tanno contino	$T_A = T_{A(min)}$ to $25^{\circ}C$	-2	1-11-
	Frequency change with temperature	$T_A = 25^{\circ}C \text{ to } 85^{\circ}C$	-10	kHz
	Maximum ramp voltage		1.8	V
	Minimum ramp voltage		1.1	V



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electrical characteristics, V_{CC} = 6 V, f_{OSC} = 500 kHz, T_A = 25°C (unless otherwise noted) (continued) dead-time control (DTC)

	DADAMETED	TEST SOUDITIONS	TL	UNIT		
	PARAMETER	TEST CONDITIONS	MIN TYP MAX			
.,	Input threshold voltage	Duty cycle = 0%		1.1		
VIT		Duty cycle = 100%		0.5		V
V _{I(latched)}	Latched-mode input voltage			1.2		V
	Latched-mode (source) current			-100		μΑ

error-amplifier

	DADAMETED	TEOT 04	NIDITIONS	TL	,	LINUT	
	PARAMETER	TEST CC	ONDITIONS	MIN	TYP	MAX	UNIT
I _{IB}	Input bias current	V _O = 1.25 V,	V _{IC} = 1.25 V		-160		nA
Ay	Open-loop voltage gain	$R_{FB} = 200 \text{ k}\Omega$			80		dB
	Unity-gain bandwidth				3		MHz
CMRR	Common-mode rejection ratio				80		dB
V _{OM(max)}	Positive output voltage swing				2.43		V
VOM(min)	Negative output voltage swing				0.63		V
I _{O+}	Output sink current	$V_{ID} = -0.1 V$,	V _O = 1.20 V		0.5		mA
IO-	Output source current	$V_{ID} = 0.1 V$,	V _O = 1.80 V		-70		μΑ

output

	242445752	TEST SOURITIONS	Τι	TL1454AY				
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT		
VOH	High lavel autout valtage	IO = -8 mA	4.5	V				
	High-level output voltage	$I_O = -40 \text{ mA}$		4.4				
,,	Law law law and a set a set to a life and	I _O = 8 mA		0.1		V		
VOL	Low-level output voltage	I _O = 40 mA		1.8				
t _{rv}	Output voltage rise time	C. 2000 pF		220				
tfV	Output voltage fall time	C _L = 2000 pF		220		ns		

supply current

	DADAMETED	TEST COMPLIANC	TL	UNIT		
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
ICC(stby)	Standby supply current	RT open, CT = 1.5 V, No load, V _O (COMP1, COMP2) = 1.25 V,		3.1		mA
ICC(average)	Average supply current	$R_T = 10 \text{ k}\Omega,$ $C_T = 120 \text{ pF},$ 50% duty cycle, Outputs open		3.5		mA



PARAMETER MEASUREMENT INFORMATION

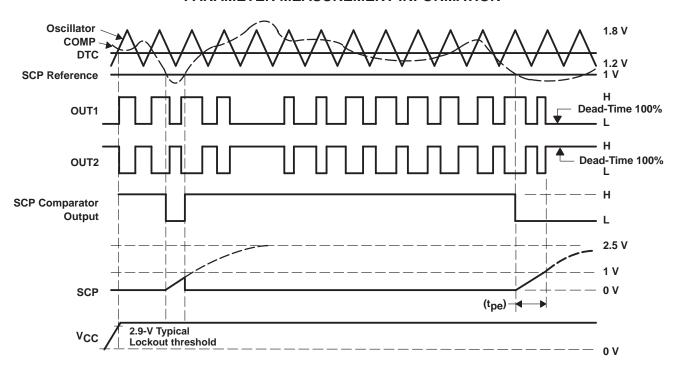
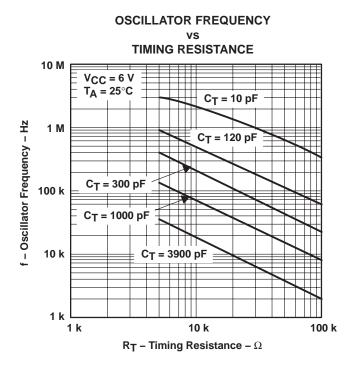


Figure 5. Timing Diagram



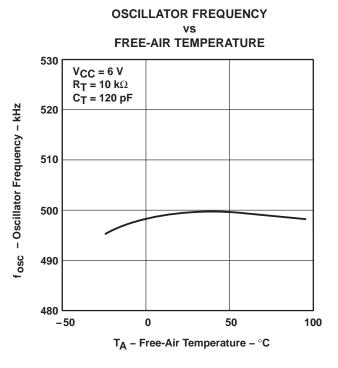
OSCILLATOR PERIOD

TYPICAL CHARACTERISTICS



TIMING CAPACITANCE 102 V_{CC} = 6 V $R_T = 5.1 \text{ k}\Omega$ T_A = 25°C t - Oscillation Period - μ s 101 10⁰ 10 - 1100 101 102 103 104 105 C_T - Timing Capacitance - pF

Figure 6



PWM TRIANGLE WAVEFORM AMPLITUDE vs

Figure 7

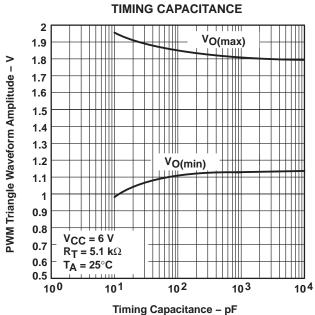


Figure 8 Figure 9

DTC INPUT THRESHOLD VOLTAGE FREE-AIR TEMPERATURE $V_{CC} = 6 V$ $R_T = 5.1 \text{ k}\Omega$ $C_T = 1000 pF$ 1.2 DTC Input Threshold Voltage - V V_{IT} (0% Duty Cycle) 0.8 0.6 VIT (100% Duty Cycle) 0.4 50 100 -50 T_A – Free-Air Temperature – $^{\circ}C$

Figure 10

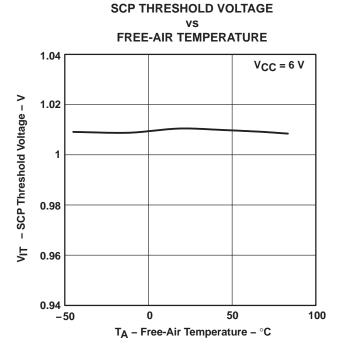


Figure 12

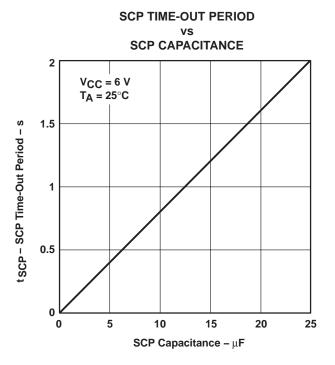


Figure 11

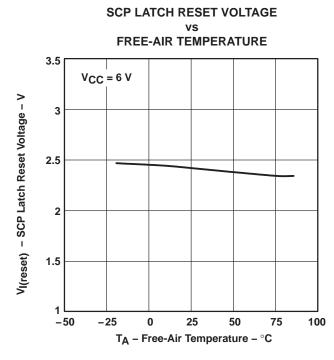
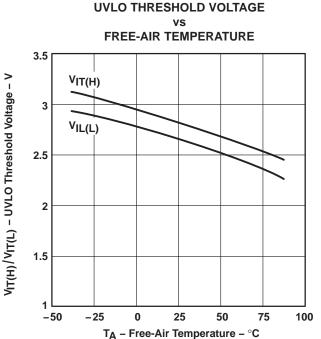


Figure 13

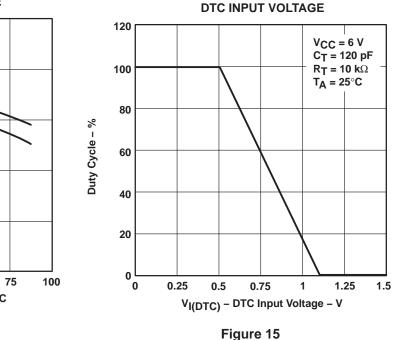


DUTY CYCLE

TYPICAL CHARACTERISTICS

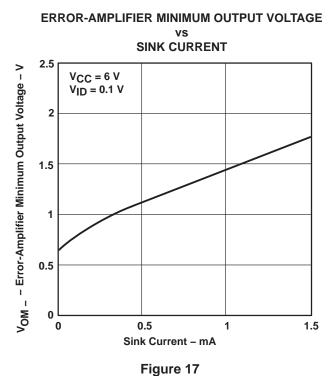


40 20 0 0.25 0 Figure 14



ERROR-AMPLIFIER MAXIMUM OUTPUT VOLTAGE vs **SOURCE CURRENT** V_{OM+} - Error-Amplifier Maximum Output Voltage - V 2.5 V_{CC} = 6 V V_{ID} = 0.1 V $T_A = 25^{\circ}C$ 2 1.5 1 0.5 0 80 120 0 40 Source Current - µA

Figure 16



ERROR AMPLIFIER MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE SWING vs **FREQUENCY** 2.5 $V_{CC} = 6 V$ Vo(PP) - Error Amplifier Maximum Peak-to-Peak Output Voltage Swing - V TA = 25°C 1.5 0.5 10 k 100 k 1M 10 M 100 M 1 k f - Frequency - Hz

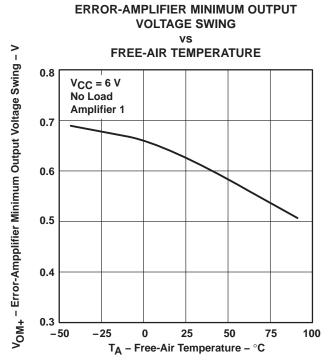


Figure 18 Figure 19

ERROR AMPLIFIER OPEN-LOOP GAIN AND PHASE SHIFT

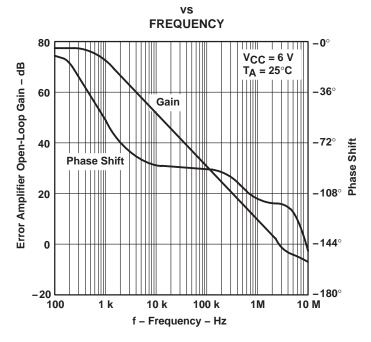


Figure 20



ERROR-AMPLIFIER POSITIVE OUTPUT VOLTAGE SWING

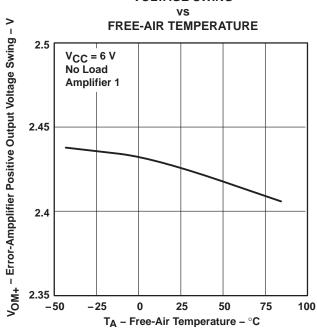


Figure 21

HIGH-LEVEL OUTPUT VOLTAGE

vs OUTPUT CURRENT

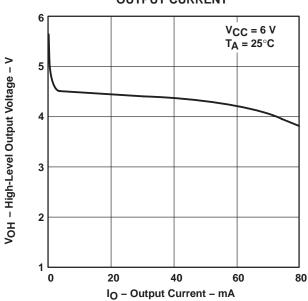


Figure 22

HIGH-LEVEL OUTPUT VOLTAGE

VS

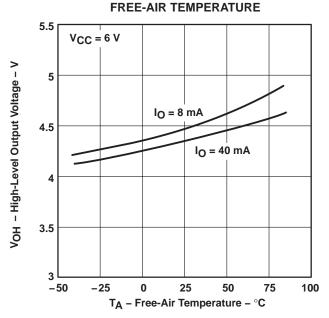


Figure 23

Figure 24

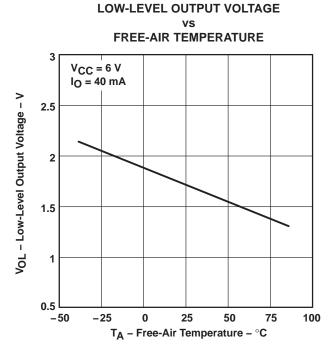


Figure 26

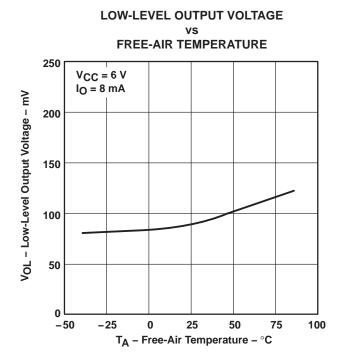


Figure 25

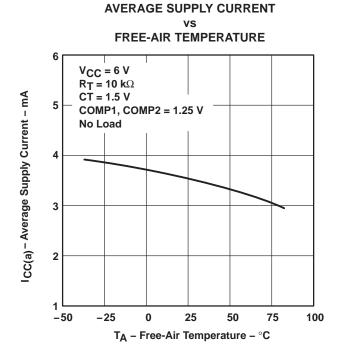


Figure 27



STANDBY SUPPLY CURRENT **SUPPLY VOLTAGE** VCC = 6 VRT = Open CT = 1.5 V ICC(stby) - Standby Supply Current - mA **COMP1, COMP2 = 1.25 V** No Load T_A = 25°C 3 2 5 0 15 20 25 10

Figure 28

V_{CC} - Supply Voltage - V

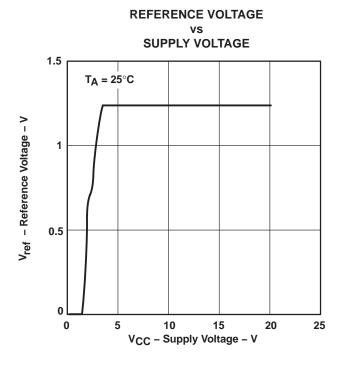
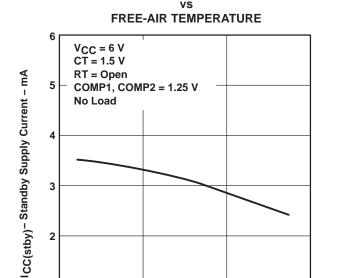


Figure 30



STANDBY SUPPLY CURRENT

Figure 29

T_A - Free-Air Temperature - °C

50

100

0

- 50

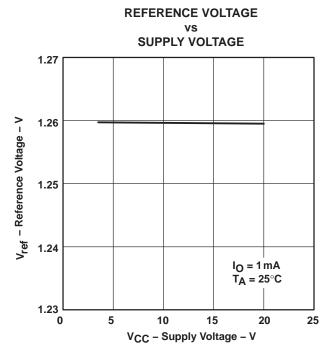


Figure 31

REFERENCE VOLTAGE vs FREE-AIR TEMPERATURE

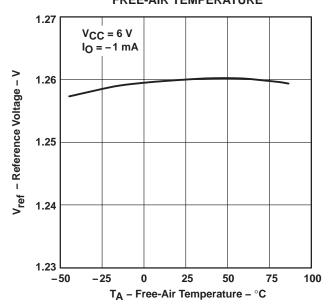


Figure 32







i.com 16-Mar-2007

PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	e Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp (3)
TL1454ACD	ACTIVE	SOIC	D	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TL1454ACDB	ACTIVE	SSOP	DB	16	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TL1454ACDBG4	ACTIVE	SSOP	DB	16	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TL1454ACDBR	ACTIVE	SSOP	DB	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TL1454ACDBRG4	ACTIVE	SSOP	DB	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TL1454ACDG4	ACTIVE	SOIC	D	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TL1454ACDR	ACTIVE	SOIC	D	16	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TL1454ACDRG4	ACTIVE	SOIC	D	16	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TL1454ACN	ACTIVE	PDIP	N	16	25	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TL1454ACNE4	ACTIVE	PDIP	N	16	25	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TL1454ACNSR	ACTIVE	SO	NS	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TL1454ACNSRG4	ACTIVE	SO	NS	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TL1454ACPW	ACTIVE	TSSOP	PW	16	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TL1454ACPWG4	ACTIVE	TSSOP	PW	16	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TL1454ACPWR	ACTIVE	TSSOP	PW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TL1454ACPWRG4	ACTIVE	TSSOP	PW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽²⁾ Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.



PACKAGE OPTION ADDENDUM

16-Mar-2007

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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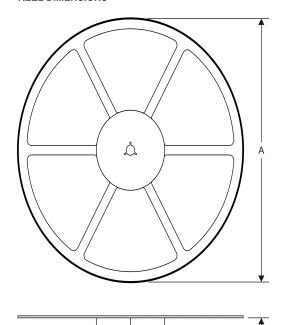
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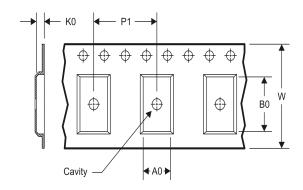
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TAPE AND REEL INFORMATION

REEL DIMENSIONS



TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

TAPE AND REEL INFORMATION

*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TL1454ACDBR	SSOP	DB	16	2000	330.0	16.4	8.2	6.6	2.5	12.0	16.0	Q1
TL1454ACDR	SOIC	D	16	2500	330.0	16.4	6.5	10.3	2.1	8.0	16.0	Q1
TL1454ACDR	SOIC	D	16	2500	330.0	16.4	6.5	10.3	2.1	8.0	16.0	Q1
TL1454ACNSR	SO	NS	16	2000	330.0	16.4	8.2	10.5	2.5	12.0	16.0	Q1
TL1454ACPWR	TSSOP	PW	16	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1

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*All dimensions are nominal

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Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TL1454ACDBR	SSOP	DB	16	2000	367.0	367.0	38.0
TL1454ACDR	SOIC	D	16	2500	367.0	367.0	38.0
TL1454ACDR	SOIC	D	16	2500	333.2	345.9	28.6
TL1454ACNSR	SO	NS	16	2000	367.0	367.0	38.0
TL1454ACPWR	TSSOP	PW	16	2000	367.0	367.0	35.0

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