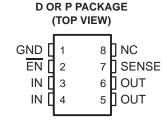
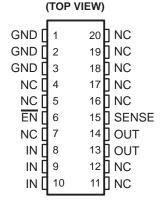
- 2.5-V Fixed-Output Regulator
- Very Low-Dropout (LDO) Voltage . . . 57 mV
 Typical at I_O = 100 mA
- Very Low Quiescent Current, Independent of Load . . . 292 μA Typ
- Extremely Low Sleep-State Current,
 0.5 μA Max
- 2% Tolerance Over Specified Conditions
- Output Current Range . . . 0 mA to 500 mA
- Available in Space Saving 8-Pin SOIC and 20-Pin TSSOP Packages
- 0°C to 125°C Operating Junction Temperature Range

description

The TPS71025 low-dropout regulator offers an order of magnitude reduction in both dropout voltage and quiescent current over conventional LDO performance. The improvement results from replacing the typical pnp pass transistor with a PMOS device.



PW PACKAGE



NC - No internal connection

Because the PMOS device behaves as a low-value resistor, the dropout voltage is very low (maximum of 95 mV at an output current of 100 mA) and is directly proportional to the output current (see Figure 1). Additionally, since the PMOS pass element is a voltage-driven device, the quiescent current is very low and remains independent of output loading (typically 292 μ A over the full range of output current, 0 mA to 500 mA). These two key specifications yield a significant improvement in operating life for battery-powered systems. The TPS71025 also features a sleep mode; applying a TTL high signal to $\overline{\text{EN}}$ (enable) shuts down the regulator, reducing the quiescent current to 0.5 μ A maximum at $T_J = 25^{\circ}\text{C}$.

AVAILABLE OPTIONS

	т.	OUTP	JTPUT VOLTAGE PACKAGED DEVICE		CKAGED DEVICES		CHIP FORM		
	IJ	MIN	TYP	MAX	SMALL OUTLINE (D)	PLASTIC DIP (P)	TSSOP (PW)	(Y)	
Г	0°C to 125°C	2.45	2.5	2.55	TPS71025D	TPS71025P	TPS71025PWLE	TPS71025Y	

The D package is available taped and reeled. Add R suffix to device type (e.g., TPS71025DR). The PW package is only available left-end taped and reeled and is indicated by the LE suffix on the device type.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



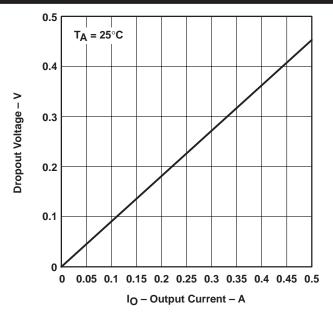
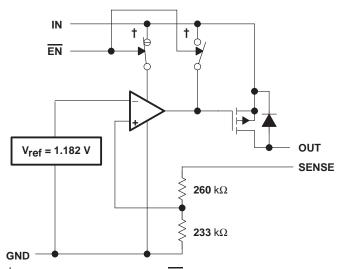


Figure 1. Dropout Voltage Versus Output Current

functional block diagram



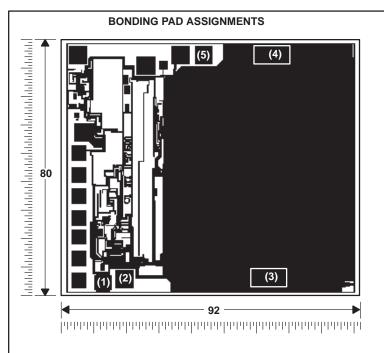
† Switch positions are shown with $\overline{\sf EN}$ low (active).

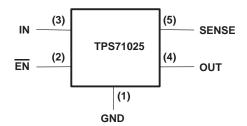
Terminal Functions

	TERMINAL		
NAME	N	0.	DESCRIPTION
NAME	D or P	PW	
EN	2 6		Enable input. Logic low enables output
GND	1	1–3	Ground
IN	3, 4	8–10	Input supply voltage
OUT	5, 6	13, 14	Output voltage
SENSE 7 15 (15	Output voltage sense input

TPS71025Y chip information

These chips, when properly assembled, display characteristics similar to those of the TPS71025. Thermal compression or ultrasonic bonding may be used on the doped aluminum bonding pads. The chips may be mounted with conductive epoxy or a gold-silicon preform.





CHIP THICKNESS: 15 MILS TYPICAL

BONDING PADS: 4 × 4 MILS MINIMUM

T_Jmax = 150°C

TOLERANCES ARE $\pm 10\%$.

ALL DIMENSIONS ARE IN MILS.

NOTE A: For most applications, OUT and SENSE should be tied together as close as possible to the device; for other implementations, refer to SENSE-pin connection discussion in the Application Information section of this data sheet.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Continuous total power dissipation See Dissipation Rating Tables 1 and 2 Operating virtual junction temperature range, T_J –0°C to 150°C

NOTE 1: All voltage values are with respect to GND

DISSIPATION RATING TABLE 1 - FREE-AIR TEMPERATURE‡

PACKAGE	$T_A \le 25^{\circ}C$ POWER RATING	DERATING FACTOR ABOVE T _A = 25°C	T _A = 70°C POWER RATING	T _A = 125°C POWER RATING
D	725 mW	5.8 mW/°C	464 mW	145 mW
Р	1175 mW	9.4 mW/°C	752 mW	235 mW
PW	700 mW	5.6 mW/°C	448 mW	140 mW

DISSIPATION RATING TABLE 2 - CASE TEMPERATURE[‡]

PACKAGE	$T_C \le 25^{\circ}C$ POWER RATING	DERATING FACTOR ABOVE T _C = 25°C	T _C = 70°C POWER RATING	T _C = 125°C POWER RATING
D	2188 mW	17.5 mW/°C	1400 mW	438 mW
Р	2738 mW	21.9 mW/°C	1752 mW	548 mW
PW	4025 mW	32.2 mW/°C	2576 mW	805 mW

[‡] Dissipation rating tables and figures are provided for maintenance of junction temperature at or below absolute maximum temperature of 150°C. For guidelines on maintaining junction temperature within recommended operating range, see the Thermal Information section.

recommended operating conditions

	MIN	MAX	UNIT
Input voltage, V _I	2.97	10	V
High-level input voltage at EN, VIH	2		V
Low-level input voltage at EN, V _{IL}	0	0.5	V
Output current range, IO	0	500	mA
Operating virtual junction temperature range, TJ	0	125	°C



[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

electrical characteristics over recommended operating junction temperature range, V_{I(IN)} = 3.5 V, I_O = 10 mA, $\overline{\text{EN}}$ = 0 V, C_O = 4.7 $\mu\text{F/CSR}^{\dagger}$ = 1 Ω , SENSE shorted to OUT (unless otherwise noted)

PARAMETER	TEST CONDITIONS‡		TJ	MIN	TYP	MAX	UNIT	
Output voltage	3.5 V ≤ V _I ≤ 10 V		25°C		2.5		V	
Output voltage	3.5 V \(\) V \(\) \(\)		0°C to 125°C	2.45		2.55	V	
	I _O = 10 mA,	V _I = 2.45 V	25°C		5.7	7.5		
	10 = 10 111/4,	V - 2.40 V	0°C to 125°C			10	m∨	
Dropout voltage	$I_{O} = 100 \text{ mA},$	V _I = 2.45 V	25°C		57	95		
Diopout voltage	10 = 100 m/r,	V ₁ = 2.10 V	0°C to 125°C			105		
	I _O = 500 mA,	V _I = 2.45 V	25°C		330	450		
	10 = 000 11// 1,	V = 2.40 V	0°C to 125°C			500		
Pass-element series resistance			25°C		0.66	0.9	Ω	
T doo olollion concertodictaries			0°C to 125°C			1		
Input regulation	$V_I = 3.5 \text{ V to } 10 \text{ V},$		25°C		7	23	m∨	
par.ogaiane	$50 \mu\text{A} \le I_{\text{O}} \le 500 \text{mA}$		0°C to 125°C		12.7	29		
	$I_0 = 5 \text{ mA to } 500 \text{ mA}$	25°C		18	38	mV		
Output regulation	$3.5 \text{ V} \le \text{V}_{\text{I}} \le 10 \text{ V}$		0°C to 125°C			75		
Carpat regulation	$I_0 = 50 \mu\text{A} \text{ to } 500 \text{mA}$	λ,	25°C		24	60	mV	
	3.5 V ≤ V _I ≤ 10 V		0°C to 125°C			120		
	f = 120 Hz.	ΙΟ = 50 μΑ	25°C	43	53			
Ripple rejection	1 = 120 112,		0°C to 125°C	40			dB	
	f = 120 Hz,	I _O = 500 mA	25°C	39	51			
	1 = 120112,	10 = 300 mz	0°C to 125°C	36				
Output noise-spectral density	f = 120 Hz		25°C		2		μV/√ Hz	
	40.11- 46.4400.111-	$C_0 = 4.7 \mu F$	25°C		274			
Output noise voltage	10 Hz \leq f \leq 100 kHz, CSR = 1 Ω	$C_0 = 10 \mu F$	25°C		228		μVrms	
		C _O = 100 μF	25°C		159			
0	<u>EN</u> ≤ 0.5 V,	•	25°C		292	390	^	
Quiescent current (active mode)	$0 \text{ mA} \le I_{\text{O}} \le 500 \text{ mA}$		0°C to 125°C			540	μΑ	
Complete our month (store dless and de)	$\overline{EN} = V_{I},$	0.7.1/ < 10.1/	25°C		18	475	A	
Supply current (standby mode)	$\square N = V$,	2.7 V ≤ V _I ≤ 10 V	0°C to 125°C			1900	nA	
Output current limit	Va = 0	V _I = 10 V	25°C		1.07	2	Α	
Output current limit	$V_O = 0$,	V = 10 V	0°C to 125°C			2	A	
Pass-element leakage current in standby	$\overline{EN} = V_1,$	2.7 V ≤ V _I ≤ 10 V	25°C		0.223	0.5		
mode	$\square V = V$,	$2.7 \text{ V} \leq \text{V} \leq 10 \text{ V}$	0°C to 125°C			1	μΑ	
Output voltage temperature coefficient			0°C to 125°C		61	75	ppm/°C	
Thermal shutdown junction temperature					165		°C	
Louis high input valtage (standburgeds) FNI	2.5 V ≤ V _I ≤ 6 V		25°C	2				
Logic high input voltage (standby mode), EN	6 V ≤ V _I ≤ 10 V		0°C to 125°C	2.7			V	
Logic low input voltage (active mode), EN	2.7 V ≤ V _I ≤ 10 V		25°C			0.5	V	
7	$ 2.7 \text{ V} \leq \text{V} \leq 10 \text{ V}$		0°C to 125°C			0.5	V	
Hysteresis voltage, EN			0°C to 125°C		50		mV	
Input current, EN	0 V ≤ V _I ≤ 10 V		25°C	-0.5		0.5	^	
Input current, EN	0 v > v > 10 v		0°C to 125°C	-0.5		0.5	μΑ	
Input voltage, minimum for active pass			25°C		2	2.5	V	
element			0°C to 125°C			2.5	v	

[†] CSR (compensation series resistance) refers to the total series resistance, including the equivalent series resistance (ESR) of the capacitor, any

series resistance added externally, and PWB trace resistance to C₀.

† Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.



electrical characteristics at T_J = 25°C, V_{I(IN)} = 3.5 V, I_O = 10 mA, $\overline{\text{EN}}$ = 0 V, C_O = 4.7 $\mu\text{F/CSR}^{\dagger}$ = 1 Ω , SENSE shorted to OUT (unless otherwise noted)

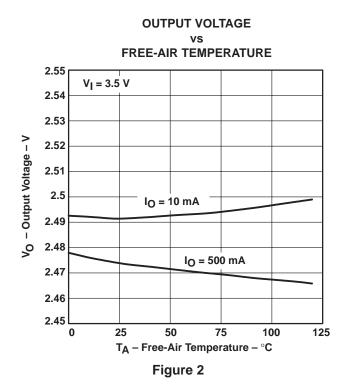
DADAMETED	TEGT 00115	NTIONOT	TP	S71025	Y	UNIT	
PARAMETER	TEST CONI	JIIIONS+	MIN	TYP	MAX	UNII	
Output voltage	$3.5 \text{ V} \leq \text{V}_{I} \leq 10 \text{ V}$			2.5		V	
	$I_O = 10 \text{ mA},$	V _I = 2.45 V	5.7 57		mV		
Dropout voltage	$I_O = 100 \text{ mA},$	V _I = 2.45 V					
	$I_O = 500 \text{ mA},$	V _I = 2.45 V		330			
Pass-element series resistance				0.66		Ω	
Input regulation	V _I = 3.5 V to 10 V			7		mV	
Output regulation	$I_{O} = 5 \text{ mA to } 500 \text{ mA}$			18		mV	
Output regulation	$I_{O} = 50 \mu\text{A} \text{ to } 500 \text{mA}$	١		24		mV	
Disale unication	f = 120 Hz,	ΙΟ = 50 μΑ		53		dB	
Ripple rejection	f = 120 Hz,	I _O = 500 mA	51] ub	
Output noise-spectral density	f = 120 Hz			2		μV/√ Hz	
		$C_0 = 4.7 \mu\text{F}$		274			
Output noise voltage	10 Hz \leq f \leq 100 kHz, CSR = 1 Ω	C _O = 10 μF		228		μVrms	
	001(= 1 52	C _O = 100 μF		159			
Quiescent current (active mode)	$\overline{\text{EN}} = 0 \text{ V},$ $0 \text{ mA} \le I_{\text{O}} \le 500 \text{ mA}$			292		μА	
Supply current (standby mode)	EN = V _I ,	2.7 V ≤ V _I ≤ 10 V		18		nA	
Output current limit	$V_{O} = 0$,	V _I = 10 V		1.07		Α	
Pass-element leakage current in standby mode	EN = V _I ,	2.7 V ≤ V _I ≤ 10 V		0.223		μА	
Output voltage temperature coefficient				61		ppm/°C	
Thermal shutdown junction temperature				165		°C	
Lacia biab innutualta sa (standhu anada) EN	$2.5~V \leq V_{I} \leq 6~V$		2			V	
Logic high input voltage (standby mode), \overline{EN}	6 V ≤ V _I ≤ 10 V			2.7		٧	
Logic low input voltage (active mode), EN	$2.7 \text{ V} \leq \text{V}_{\text{I}} \leq 10 \text{ V}$				0.5	V	
Hysteresis voltage, EN				50		mV	
Input current, EN	0 V ≤ V _I ≤ 10 V			0		μΑ	
Input voltage, minimum for active pass element				2		V	

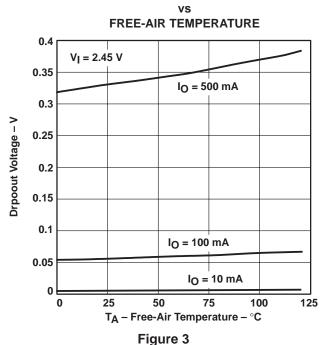
[†]CSR (compensation series resistance) refers to the total series resistance, including the equivalent series resistance (ESR) of the capacitor , any series resistance added externally, and PWB trace resistance to C₀.

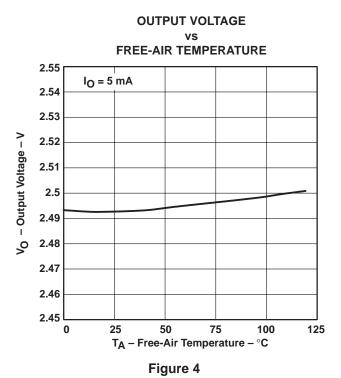
[‡] Pulse-testing techniques are used to maintain virtual junction temperature as close as possible to ambient temperature; thermal effects must be taken into account separately.

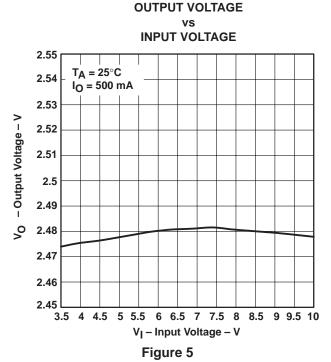
DROPOUT VOLTAGE

TYPICAL CHARACTERISTICS

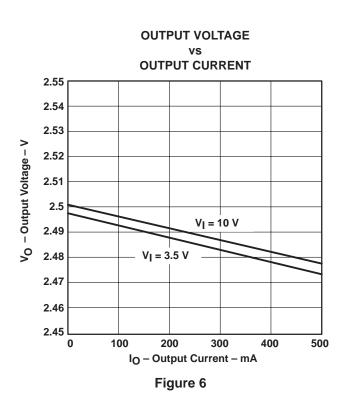


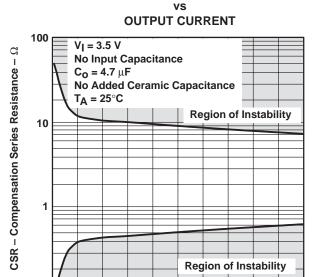






TYPICAL CHARACTERISTICS



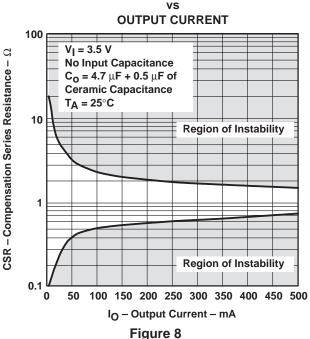


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TYPICAL REGIONS OF STABILITY

COMPENSATION SERIES RESISTANCE

TYPICAL REGIONS OF STABILITY COMPENSATION SERIES RESISTANCE vs

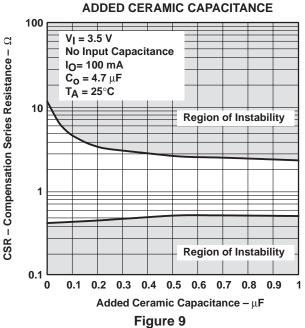


TYPICAL REGIONS OF STABILITY COMPENSATION SERIES RESISTANCE vs

IO - Output Current - mA

Figure 7

100 150 200 250 300 350 400 450 500



TYPICAL CHARACTERISTICS

TYPICAL REGIONS OF STABILITY COMPENSATION SERIES RESISTANCE vs

ADDED CERAMIC CAPACITANCE

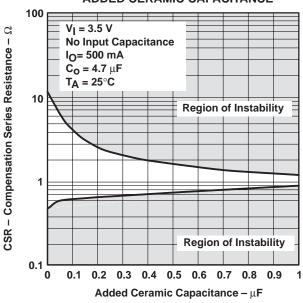
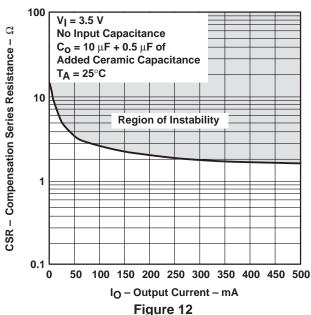


Figure 10

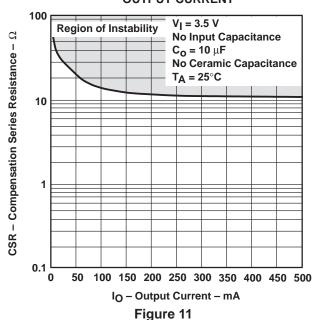
TYPICAL REGIONS OF STABILITY COMPENSATION SERIES RESISTANCE

OUTPUT CURRENT



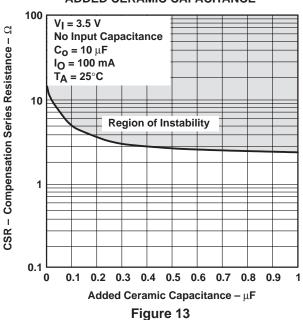
[†] CSR values below 0.1 Ω are not recommended.

TYPICAL REGIONS OF STABILITY[†] COMPENSATION SERIES RESISTANCE vs OUTPUT CURRENT



TYPICAL REGIONS OF STABILITY † COMPENSATION SERIES RESISTANCE

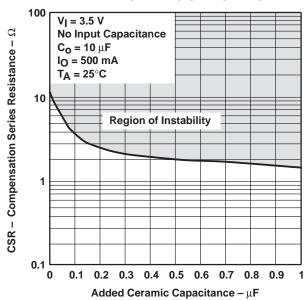
vs ADDED CERAMIC CAPACITANCE



TYPICAL CHARACTERISTICS

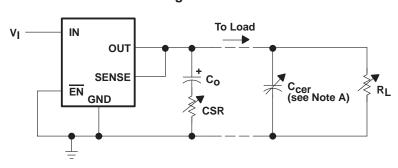
TYPICAL REGIONS OF STABILITY † COMPENSATION SERIES RESISTANCE

ADDED CERAMIC CAPACITANCE



 † CSR values below 0.1 Ω are not recommended.

Figure 14



NOTE A: Ceramic capacitor

Figure 15. Test Circuit for Typical Regions of Stability (Figures 7 through 14)

THERMAL INFORMATION

In response to system-miniaturization trends, integrated circuits are being offered in low-profile and fine-pitch surface-mount packages. Implementation of many of today's high-performance devices in these packages requires special attention to power dissipation. Many system-dependent issues such as thermal coupling, airflow, added heat sinks and convection surfaces, and the presence of other heat-generating components affect the power-dissipation limits of a given component.

Three basic approaches for enhancing thermal performance are illustrated in this discussion:

- Improving the power-dissipation capability of the PWB design
- Improving the thermal coupling of the component to the PWB
- Introducing airflow in the system

Figure 16 is an example of a thermally enhanced PWB layout for the 20-lead TSSOP package. This layout involves adding copper on the PWB to conduct heat away from the device. The $R_{\theta JA}$ for this component/board system is illustrated in Figure 17. The family of curves illustrates the effect of increasing the size of the copper-heat-sink surface area. The PWB is a standard FR4 board (L × W × H = 3.2 inch × 3.2 inch × 0.062 inch); the board traces and heat sink area are 1-oz (per square foot) copper.

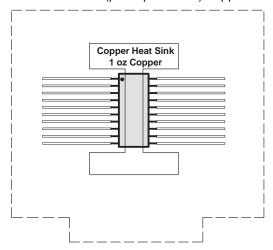
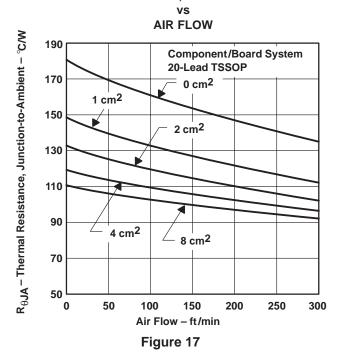


Figure 16. Thermally Enhanced PWB Layout (Not to Scale) for the 20-Pin TSSOP

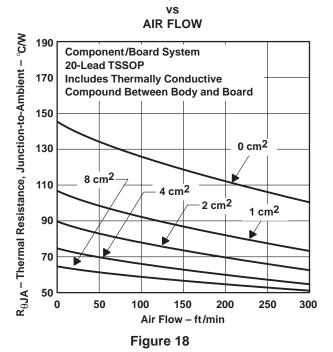
Figure 18 shows the thermal resistance for the same system with the addition of a thermally conductive compound between the body of the TSSOP package and the PWB copper routed directly beneath the device. The thermal conductivity for the compound used in this analysis is $0.815 \text{ W/m} \times {}^{\circ}\text{C}$.

THERMAL INFORMATION

THERMAL RESISTANCE, JUNCTION-TO-AMBIENT



THERMAL RESISTANCE, JUNCTION-TO-AMBIENT



Using these figures to determine the system $R_{\theta JA}$ allows the maximum power-dissipation $P_{D(max)}$ limit to be calculated with the equation:

$$P_{D(max)} = \frac{T_{J(max)} - T_{A}}{R_{\theta JA(system)}}$$

Where

T_{J(max)} is the maximum allowable junction temperature (i.e., 150°C absolute maximum or 125°C maximum recommended operating temperature for specified operation).

This limit should then be applied to the internal power dissipated by the TPS71025 regulator. The equation for calculating total internal power dissipation of the device is:

$$P_{D(total)} = (V_{I} - V_{O}) \times I_{O} + (V_{I} \times I_{Q})$$

Because the quiescent current is very low, the second term is negligible, further simplifying the equation to:

$$P_{D(total)} = (V_{I} - V_{O}) \times I_{O}$$

THERMAL INFORMATION

For a 20-lead TSSOP/FR4 board system with thermally conductive compound between the board and the device body, where $T_A = 55^{\circ}C$, airflow = 100 ft/min, and copper heat sink area = 1 cm², the maximum power-dissipation limit can be calculated. As indicated in Figure 18, the system $R_{\theta JA}$ is 94°C/W; therefore, the maximum power-dissipation limit is:

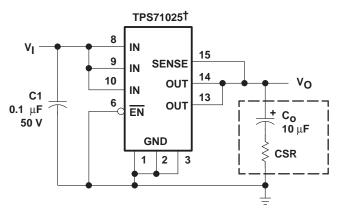
$$P_{D(max)} = \frac{T_{J(max)} - T_{A}}{R_{\theta JA(system)}} = \frac{125^{\circ}C - 55^{\circ}C}{94^{\circ}C/W} = 745 \text{ mW}$$

If the system implements a TPS71025 regulator where $V_1 = 3.3 \text{ V}$ and $I_O = 385 \text{ mA}$, the internal power dissipation is:

$$P_{D(total)} = (V_I - V_O) \times I_O = (3.3 - 2.5) \times 0.385 = 308 \text{ mW}$$

Comparing $P_{D(total)}$ with $P_{D(max)}$ reveals that the power dissipation in this example does not exceed the maximum limit. When it does, one of two corrective actions can be taken. The power-dissipation limit can be raised by increasing the airflow or the heat-sink area. Alternatively, the internal power dissipation of the regulator can be lowered by reducing the input voltage or the load current. In either case, the above calculations should be repeated with the new system parameters.

APPLICATION INFORMATION



† Capacitor selection is nontrivial. See external capacitor requirements section.

Figure 19. Typical Application Circuit

The TPS71025 low-dropout (LDO) regulator overcomes many of the shortcomings of earlier-generation LDOs, while adding features such as a power-saving shutdown mode.

device operation

The TPS71025, unlike many other LDOs, features very low quiescent current that remains virtually constant even with varying loads. Conventional LDO regulators use a pnp-pass element, the base current of which is directly proportional to the load current through the regulator ($I_B = I_C/\beta$). Examination of the data sheets reveals that those devices are typically specified under near no-load conditions; actual operating currents are much higher as evidenced by typical quiescent current versus load current curves. The TPS71025 uses a PMOS transistor to pass current; because the gate of the PMOS element is voltage driven, operating currents are low and stable over the full load range. The TPS71025 specifications reflect actual performance under load.

Another pitfall associated with the pnp-pass element is its tendency to saturate when the device goes into dropout. The resulting drop in β forces an increase in I_B to maintain the load. During power up, this translates to large start-up currents. Systems with limited supply current may fail to start up. In battery-powered systems, it means rapid battery discharge when the voltage decays below the minimum required for regulation. The TPS71025 quiescent current remains low even when the regulator drops out, eliminating both problems.

The TPS71025 also features a shutdown mode that places the output in the high-impedance state (essentially equal to the feedback-divider resistance) and reduces quiescent current to under $2\,\mu\text{A}$. If the shutdown feature is not used, $\overline{\text{EN}}$ should be tied to ground. Response to an enable transition is quick; regulated output voltage is reestablished in typically 120 μs .

minimum load requirements

The TPS71025 family is stable even at zero load; no minimum load is required for operation.

SENSE-pin connection

The SENSE pin must be connected to the regulator output for proper functioning of the regulator. Normally, this connection should be as short as possible; however, the connection can be made near a critical circuit (remote sense) to improve performance at that point. Internally, SENSE connects to a high-impedance wide-bandwidth amplifier through a resistor-divider network, and noise pickup feeds through to the regulator output. Routing the SENSE connection to minimize/avoid noise pickup is essential. Adding an RC network between SENSE and OUT to filter noise is not recommended because it can cause the regulator to oscillate.



APPLICATION INFORMATION

external capacitor requirements

An input capacitor is not required; however, a ceramic bypass capacitor (0.047 pF to 0.1 μ F) improves load transient response and noise rejection if the TPS71025 is located more than a few inches from the power supply. A higher-capacitance electrolytic capacitor may be necessary if large (hundreds of milliamps) load transients with fast rise times are anticipated.

As with most LDO regulators, the TPS71025 requires an output capacitor for stability. A low-ESR 10- μ F solid-tantalum capacitor connected from the regulator output to ground is sufficient to ensure stability over the full load range (see Figure 11). Adding high-frequency ceramic or film capacitors (such as power-supply bypass capacitors for digital or analog ICs) can cause the regulator to become unstable unless the ESR of the tantalum capacitor is less than 1.2 Ω over temperature. Capacitors with published ESR specifications such as the AVX TPSD106K035R0300 and the Sprague 593D106X0035D2W work well because the maximum ESR at 25°C is 300 m Ω (typically, the ESR in solid-tantalum capacitors increases by a factor of 2 or less when the temperature drops from 25°C to -40°C). Where component height and/or mounting area is a problem, physically smaller, 10- μ F devices can be screened for ESR. Figure 7 through Figure 14 show the stable regions of operation using different values of output capacitance with various values of ceramic load capacitance.

In applications with little or no high-frequency bypass capacitance (< 0.2 μ F), the output capacitance can be reduced to 4.7 μ F, provided ESR is maintained between 0.7 and 2.5 Ω . Because minimum capacitor ESR is seldom if ever specified, it may be necessary to add a 0.5- Ω to 1- Ω resistor in series with the capacitor and limit ESR to 1.5 Ω maximum. As shown in the ESR graphs (Figure 7 through Figure 14), minimum ESR is not a problem when using 10- μ F or larger output capacitors.

Below is a partial listing of surface-mount capacitors usable with the TPS71025. This information (along with the ESR graphs, Figure 7 through Figure 14) is included to assist in selection of suitable capacitance for the application. When necessary to achieve low height requirements along with high output current and/or high ceramic load capacitance, several higher ESR capacitors can be used in parallel to meet the guidelines above.



APPLICATION INFORMATION

external capacitor requirements (continued)

All load and temperature conditions with up to 1 μF of added ceramic load capacitance:

PART NO.	MFR.	VALUE	MAX ESR†	SIZE $(H \times L \times W)^{\dagger}$
T421C226M010AS	Kemet	$22~\mu F,10~V$	0.5	$2.8\times 6\times 3.2$
593D156X0025D2W	Sprague	15 μF, 25 V	0.3	$2.8\times7.3\times4.3$
593D106X0035D2W	Sprague	10 μ F, 35 V	0.3	$2.8\times7.3\times4.3$
TPSD106M035R0300	AVX	10 μF, 35 V	0.3	$2.8\times7.3\times4.3$

Load < 200 mA, ceramic load capacitance < 0.2 μ F, full temperature range:

PART NO.	MFR.	VALUE	MAX ESR†	SIZE $(H \times L \times W)^{\dagger}$
592D156X0020R2T	Sprague	15 μ F, 20 V	1.1	$1.2\times7.2\times6$
595D156X0025C2T	Sprague	15 μF, 25 V	1	$2.5\times7.1\times3.2$
595D106X0025C2T	Sprague	10 μ F, 25 V	1.2	$2.5\times7.1\times3.2$
293D226X0016D2W	Sprague	$22\mu\text{F}$, 16V	1.1	$2.8\times7.3\times4.3$

Load < 100 mA, ceramic load capacitance < 0.2 μ F, full temperature range:

PART NO.	MFR.	VALUE	MAX ESR†	SIZE $(H \times L \times W)^{\dagger}$
195D106X06R3V2T	Sprague	10 $\mu F, 6.3 \; V$	1.5	$1.3\times3.5\times2.7$
195D106X0016X2T	Sprague	10 μ F, 16 V	1.5	$1.3\times7\times2.7$
595D156X0016B2T	Sprague	15 μ F, 16 V	1.8	$1.6\times3.8\times2.6$
695D226X0015F2T	Sprague	$22~\mu\text{F},~15~\text{V}$	1.4	$1.8\times6.5\times3.4$
695D156X0020F2T	Sprague	15 μ F, 20 V	1.5	$1.8\times6.5\times3.4$
695D106X0035G2T	Sprague	10 μF, 35 V	1.3	$2.5\times7.6\times2.5$

 $^{^\}dagger$ Size is in mm. ESR is maximum resistance at 100 kHz and T_A = 25°C. Listings are sorted by height.

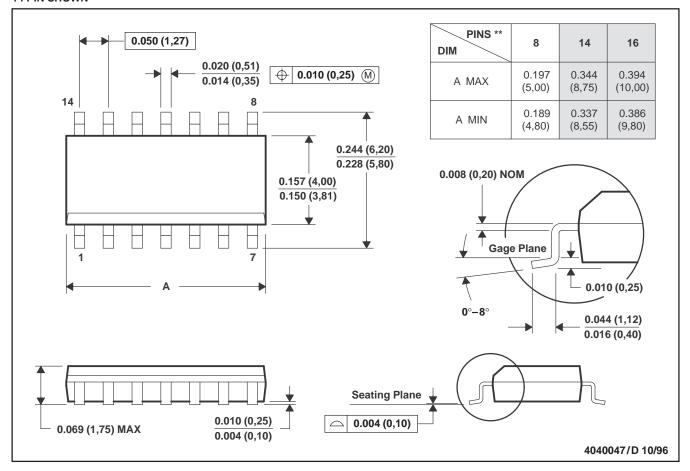


MECHANICAL DATA

D (R-PDSO-G**)

14 PIN SHOWN

PLASTIC SMALL-OUTLINE PACKAGE



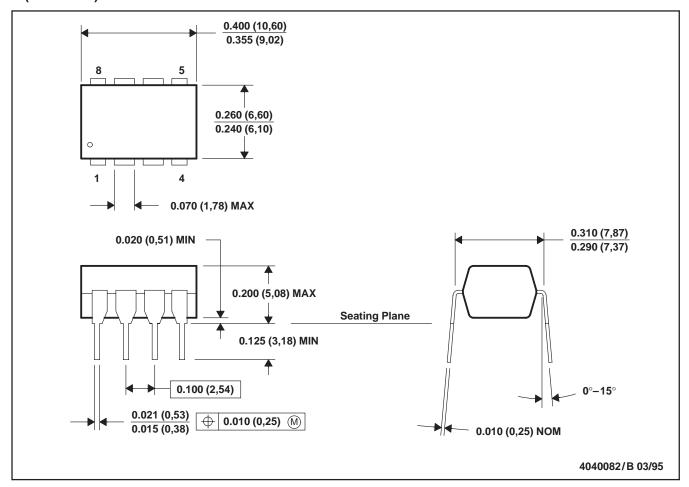
NOTES: A. All linear dimensions are in inches (millimeters).

- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion, not to exceed 0.006 (0,15).
- D. Falls within JEDEC MS-012

MECHANICAL DATA

P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE PACKAGE



NOTES: A. All linear dimensions are in inches (millimeters).

B. This drawing is subject to change without notice.

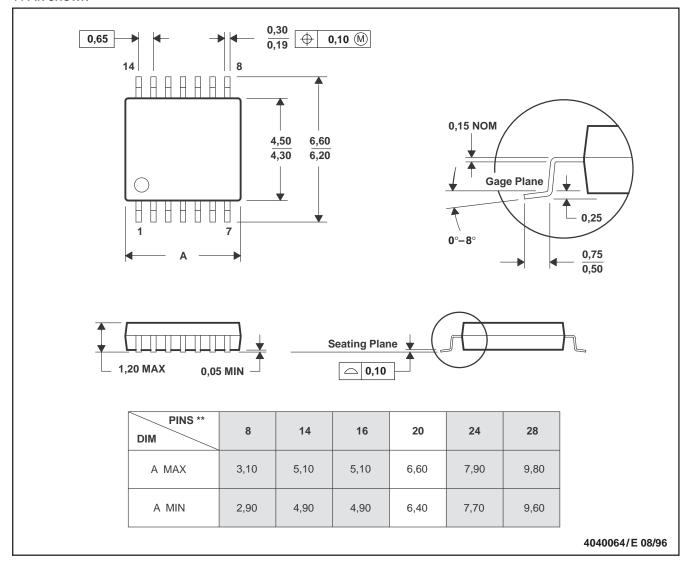
C. Falls within JEDEC MS-001

MECHANICAL DATA

PW (R-PDSO-G**)

14 PIN SHOWN

PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.

C. Body dimensions do not include mold flash or protrusion not to exceed 0,15.

D. Falls within JEDEC MO-153





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PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	e Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp ⁽³⁾
TPS71025D	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS71025DG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS71025DR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS71025DRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS71025P	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TPS71025PE4	ACTIVE	PDIP	Р	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type
TPS71025PWLE	OBSOLETE	TSSOP	PW	20		TBD	Call TI	Call TI
TPS71025PWR	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS71025PWRG4	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

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(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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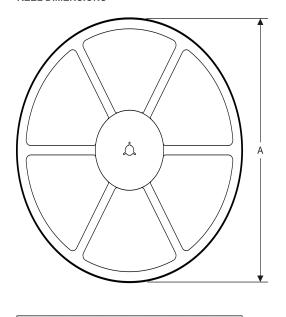
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PACKAGE MATERIALS INFORMATION

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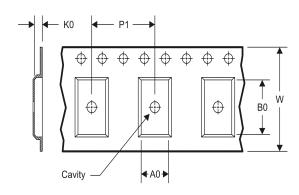
TAPE AND REEL INFORMATION

REEL DIMENSIONS



B0 Dimension designed to accommodate the component length K0 Dimension designed to accommodate the component thickness W Overall width of the carrier tape P1 Pitch between successive cavity centers

TAPE DIMENSIONS



Dimension designed to accommodate the component width

TAPE AND REEL INFORMATION

*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS71025DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TPS71025PWR	TSSOP	PW	20	2000	330.0	16.4	6.95	7.1	1.6	8.0	16.0	Q1

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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS71025DR	SOIC	D	8	2500	367.0	367.0	35.0
TPS71025PWR	TSSOP	PW	20	2000	367.0	367.0	38.0

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