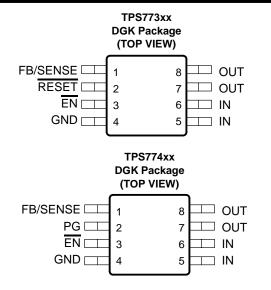
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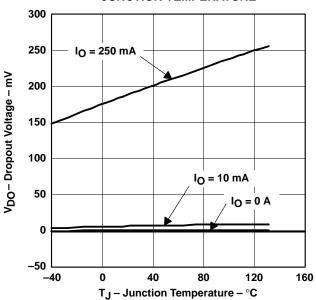
- Open Drain Power-On Reset With 220-ms Delay (TPS773xx)
- Open Drain Power-Good (PG) Status Output (TPS774xx)
- 250-mA Low-Dropout Voltage Regulator
- Available in 1.5-V, 1.6-V (TPS77316 Only),
 1.8-V, 2.7-V, 2.8-V, 3.3-V, 5.0-V Fixed Output and Adjustable Versions
- Dropout Voltage Typically 200 mV at 250 mA (TPS77333, TPS77433)
- Ultralow 92-μA Quiescent Current (Typ)
- 8-Pin MSOP (DGK) Package
- Low Noise (55 μV_{rms}) Without an External Filter (Bypass) Capacitor (TPS77318, TPS77418)
- 2% Tolerance Over Specified Conditions For Fixed-Output Versions
- Fast Transient Response
- Thermal Shutdown Protection
- See the TPS779xx Family of Devices for Active High Enable

description

The TPS773xx and TPS774xx are low-dropout regulators with integrated power-on reset and power good (PG) function respectively. These devices are capable of supplying 250 mA of output current with a dropout of 200 mV (TPS77333, TPS77433). Quiescent current is 92 μA at full load dropping down to 1 µA when device is disabled. These devices are optimized to be stable with a wide range of output capacitors including low ESR ceramic (10 μ F) or low capacitance (1 μ F) tantalum capacitors. These devices have extremely low noise output performance (55 µV_{rms}) without using any added filter capacitors. TPS773xx and TPS774xx are designed to have fast transient response for larger load current changes.



TPS77x33 DROPOUT VOLTAGE vs JUNCTION TEMPERATURE



The TPS773xx or TPS774xx is offered in 1.5-V, 1.6 V (TPS77316 only), 1.8-V, 2.7-V, 2.8-V, 3.3-V, and 5.0-V fixed-voltage versions and in an adjustable version (programmable over the range of 1.5 V to 5.5 V). Output voltage tolerance is 2% over line, load, and temperature ranges. The TPS773xx and TPS774xx families are available in 8-pin MSOP (DGK) packages.



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SLVS281E - FEBRUARY 2000 - REVISED JULY 2001

description (continued)

Because the PMOS device behaves as a low-value resistor, the dropout voltage is very low (typically 200 mV at an output current of 250 mA for 3.3-volt option) and is directly proportional to the output current. Additionally, since the PMOS pass element is a voltage-driven device, the quiescent current is very low and independent of output loading (typically 92 μ A over the full range of output current, 0 mA to 250 mA). These two key specifications yield a significant improvement in operating life for battery-powered systems.

The device is enabled when the \overline{EN} pin is connected to a low-level input voltage. This LDO family also features a sleep mode; applying a TTL high signal to \overline{EN} (enable) shuts down the regulator, reducing the quiescent current to less than 1 μA at $T_{,l} = 25^{\circ} C$.

The TPS773xx features an integrated power-on reset, commonly used as a supply voltage supervisor (SVS), or reset output voltage. The RESET output of the TPS773xx initiates a reset in DSP, microcomputer or microprocessor systems at power up and in the event of an undervoltage condition. An internal comparator in the TPS773xx monitors the output voltage of the regulator to detect an undervoltage condition on the regulated output voltage. When OUT reaches 95% of its regulated voltage, RESET will go to a high-impedance state after a 220-ms delay. RESET will go to low-impedance state when OUT is pulled below 95% (i.e. over load condition) of its regulated voltage.

For the TPS774xx, the power good terminal (PG) is an active high output, which can be used to implement a power-on reset or a low-battery indicator. An internal comparator in the TPS774xx monitors the output voltage of the regulator to detect an undervoltage condition on the regulated output voltage. When OUT falls below 82% of its regulated voltage, PG will go to a low-impedance state. PG will go to a high-impedance state when OUT is above 82% of its regulated voltage.

OUTPUT VOLTAGE (V) PACKAGED DEVICES MSOP (DGK) ТJ TPS774xx TPS773xx **TYP** SYMBOL **SYMBOL** 5.0 TPS77350DGK AGN TPS77450DGK **AGW** AGM AGV 3.3 TPS77333DGK TPS77433DGK 2.8 TPS77328DGK **AGK** TPS77428DGK **AGT** 2.7 TPS77327DGK AGJ TPS77427DGK AGS -40°C to 125°C 1.8 TPS77318DGK AGH TPS77418DGK AGQ **AWF** 1.6 TPS77316DGK 1.5 TPS77315DGK AGG TPS77415DGK AGP Adiustable TPS77301DGK **AGF** TPS77401DGK AGO 1.5 V to 5.5 V

AVAILABLE OPTIONS

NOTE: The TPS77301 and TPS77401 are programmable using an external resistor divider (see application information). The DGK package is available taped and reeled. Add an R suffix to the device type (e.g., TPS77301DGKR).

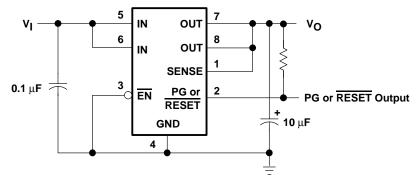


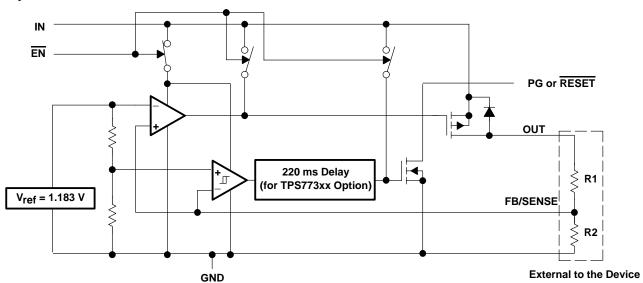
Figure 1. Typical Application Configuration (For Fixed Output Options)



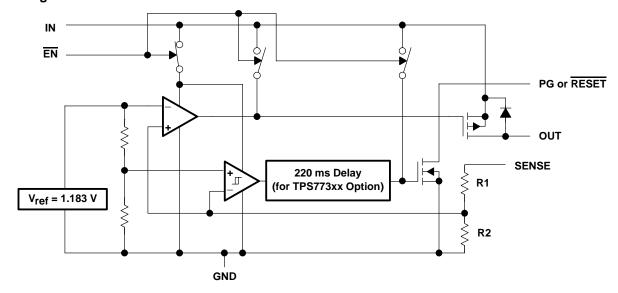
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functional block diagrams

adjustable version



fixed-voltage version

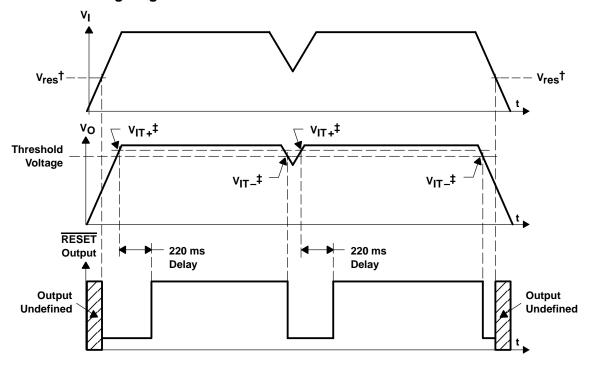


SLVS281E - FEBRUARY 2000 - REVISED JULY 2001

Terminal Functions

TERMI	NAL		DECODIFICAL
NAME	ME NO.		DESCRIPTION
TPS773XX			
FB/SENSE	1	I	Feedback input voltage for adjustable device (sense input for fixed options)
RESET	2	0	Reset output
EN	3	I	Enable input
GND	4		Regulator ground
IN	5, 6	I	Input voltage
OUT	7, 8	0	Regulated output voltage
TPS774XX			
FB/SENSE	1	I	Feedback input voltage for adjustable device (sense input for fixed options)
PG	2	0	Power good
EN	3	I	Enable input
GND	4		Regulator ground
IN	5, 6	I	Input voltage
OUT	7, 8	0	Regulated output voltage

TPS773xx RESET timing diagram



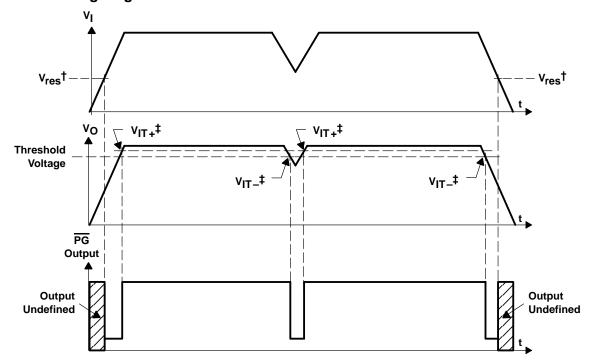
 $^{^{\}dagger} \, V_{res} \, \text{is the minimum input voltage for a valid } \overline{\text{RESET}}. \, \text{The symbol } \, V_{res} \, \text{is not currently listed within EIA or JEDEC standards for semiconductor} \, V_{res} \, \text{is not currently listed within EIA or JEDEC standards} \, V_{res} \, \text{is not currently listed within EIA or JEDEC} \, \text{standards} \, \text{for semiconductor} \, V_{res} \, \text{is not currently listed within EIA} \, \text{or JEDEC} \, \text{standards} \, \text{for semiconductor} \, V_{res} \, \text{is not currently listed} \, \text{within EIA} \, \text{or JEDEC} \, \text{standards} \, \text{for semiconductor} \, \text{for semiconductor} \, \text{or JEDEC} \, \text{standards} \, \text{for semiconductor} \, \text{or JEDEC} \, \text{standards} \, \text{for semiconductor} \, \text{for semiconductor} \, \text{or JEDEC} \, \text{standards} \, \text{for semiconductor} \, \text{for semico$



[‡] V_{IT} – Trip voltage is typically 5% lower than the output voltage (95%V_O) V_{IT} to V_{IT} is the hysteresis voltage.

SLVS281E - FEBRUARY 2000 - REVISED JULY 2001

TPS774xx PG timing diagram



[†] V_{res} is the minimum input voltage for a valid PG. The symbol V_{res} is not currently listed within EIA or JEDEC standards for semiconductor symbology.

[‡] V_{IT} – Trip voltage is typically 18% lower than the output voltage (82%V_O) V_{IT} to V_{IT} is the hysteresis voltage.

SLVS281E - FEBRUARY 2000 - REVISED JULY 2001

absolute maximum ratings over operating free-air temperature (unless otherwise noted)†

Input voltage range [‡] , V _I	
Voltage range at EN	
Maximum RESET voltage (TPS773xx)	16.5 V
Maximum PG voltage (TPS774xx)	16.5 V
Peak output current	Internally limited
Continuous total power dissipation	See Dissipation Rating Table
Output voltage, VO (OUT, FB)	5.5 V
Operating virtual junction temperature range, T _J	–40°C to 125°C
Storage temperature range, T _{stg}	–65°C to 150°C
ESD rating, HBM	

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

DISSIPATION RATING TABLE - FREE-AIR TEMPERATURES

PACKAGE	AIR FLOW (CFM)	θJA (°C/W)	θJC (°C/W)	T _A < 25°C POWER RATING	DERATING FACTOR ABOVE T _A = 25°C	T _A = 70°C POWER RATING	T _A = 85°C POWER RATING
	0	266.2	3.84	376 mW	3.76 mW/°C	207 mW	150 mW
DGK	150	255.2	3.92	392 mW	3.92 mW/°C	216 mW	157 mW
	250	242.8	4.21	412 mW	4.12 mW/°C	227 mW	165 mW

recommended operating conditions

	MIN	MAX	UNIT
Input voltage, V _I §	2.7	10	V
Output voltage range, VO	1.5	5.5	V
Output current, IO (see Note 1)	0	250	mA
Operating virtual junction temperature, T _J (see Note 1)	-40	125	°C

To calculate the minimum input voltage for your maximum output current, use the following equation: $V_{I(min)} = V_{O(max)} + V_{DO(max load)}$. NOTE 1: Continuous current and operating junction temperature are limited by internal protection circuitry, but it is not recommended that the device operate under conditions beyond those specified in this table for extended periods of time.



[‡] All voltage values are with respect to network terminal ground.

SLVS281E - FEBRUARY 2000 - REVISED JULY 2001

electrical characteristics over recommended operating junction temperature range ($T_J = -40^{\circ}\text{C}$ to 125°C), $V_I = V_{O(typ)} + 1$ V, $I_O = 1$ mA, $\overline{\text{EN}} = 0$ V, $C_O = 10~\mu\text{F}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
	Adjustable	$1.5 \text{ V} \le \text{V}_{\text{O}} \le 5.5 \text{ V}, \qquad \text{T}_{\text{J}} = 25^{\circ}\text{C}$		٧o		٧	
	voltage	$1.5 \text{ V} \le \text{V}_{\text{O}} \le 5.5 \text{ V}$	0.98V _O		1.02V _O	V	
	1 E V Output	$T_J = 25^{\circ}C$, $2.7 \text{ V} < V_{IN} < 10 \text{ V}$	1	1.5		٧	
	1.5-V Output	2.7 V < V _{IN} < 10 V	1.470		1.530	V	
	1.6-V Output	$T_J = 25^{\circ}C$, $2.7 \text{ V} < V_{IN} < 10 \text{ V}$	1	1.6		V	
	1.6-v Output	2.7 V < V _{IN} < 10 V	1.568		1.632	V	
	1.8-V Output	$T_J = 25^{\circ}C$, $2.8 \text{ V} < V_{IN} < 10 \text{ V}$	1	1.8			
Output voltage (see Notes 2 and 4)	1.6-V Output	2.8 V < V _{IN} < 10 V	1.764		1.836		
Output voltage (see Notes 2 and 4)	2.7-V Output	$T_J = 25^{\circ}C$, $3.7 \text{ V} < V_{IN} < 10 \text{ V}$	1	2.7			
	2.7-V Output	3.7 V < V _{IN} < 10 V	2.646		2.754	٧	
	2.8-V Output	$T_J = 25^{\circ}C$, $3.8 \text{ V} < V_{IN} < 10 \text{ V}$	1	2.8		V	
	2.0-V Output	3.8 V < V _{IN} < 10 V	2.744		2.856		
	3.3-V Output	$T_J = 25^{\circ}C$, $4.3 \text{ V} < V_{IN} < 10 \text{ V}$	1	3.3			
	3.3-V Output	4.3 V < V _{IN} < 10 V	3.234		3.366		
	5.0-V Output	$T_J = 25^{\circ}C$, $6.0 \text{ V} < V_{IN} < 10 \text{ V}$	1	5.0		V	
		6.0 V < V _{IN} < 10 V	4.900		5.100	٧	
Quiescent current (GND current) (see	Notes 2 and 4)	T _J = 25°C		92		μΑ	
Quiescent current (GND current) (see	r Notes 2 and 4)				125	μΛ	
Output voltage line regulation (ΔV _O /V	(a) (see Note 3)	$V_{O} + 1 V < V_{I} \le 10 V$, $T_{J} = 25^{\circ}C$		0.005		%/V	
Output voltage line regulation (Avo), v	07(300 14010 3)	V _O + 1 V < V _I ≤ 10 V			0.05	%/V	
Load regulation		T _J = 25°C		1		mV	
Output noise voltage		BW = 300 Hz to 100 kHz, T _J = 25°C, TPS77318, TPS77418		55		μVrms	
Output current limit		VO = 0 V		0.9	1.3	Α	
Peak output current		2 ms pulse width, 50% duty cycle		400		mA	
Thermal shutdown junction temperatu	ıre			144		°C	
Ota a discount		$\overline{\text{EN}} = V_{\text{I}}, \qquad T_{\text{J}} = 25^{\circ}\text{C}$			1	μΑ	
Standby current		EN = V _I			3	μΑ	
FB input current	Adjustable voltage	FB = 1.5 V			1	μΑ	
High level enable input voltage			2			V	
Low level enable input voltage					0.7	V	
Enable input current			-1		1	μΑ	
Power supply ripple rejection (TPS77	318, TPS77418)	f = 1 kHz, T _J = 25°C		55		dB	

NOTES: 2. Minimum input operating voltage is 2.7 V or $V_{O(typ)}$ + 1 V, whichever is greater. Maximum input voltage = 10 V, minimum output current 1 mA.

3. If $V_O < 1.8 \text{ V then } V_{I(max)} = 10 \text{ V}, V_{I(min)} = 2.7 \text{ V}$:

Line regulation (mV) =
$$(\%/V) \times \frac{V_O(V_{I(max)} - 2.7 V)}{100} \times 1000$$

If $V_O > 2.5 \text{ V}$ then $V_{I(max)} = 10 \text{ V}$, $V_{I(min)} = V_O + 1 \text{ V}$:

Line regulation (mV) =
$$(\%/V) \times \frac{V_O(V_{I(max)} - (V_O + 1))}{100} \times 1000$$

4. $I_0 = 1 \text{ mA to } 250 \text{ mA}$



SLVS281E - FEBRUARY 2000 - REVISED JULY 2001

electrical characteristics over recommended operating junction temperature range ($T_J = -40^{\circ}\text{C}$ to 125°C), $V_I = V_{O(typ)} + 1$ V, $I_O = 1$ mA, $\overline{\text{EN}} = 0$ V, $C_O = 10~\mu\text{F}$ (unless otherwise noted) (continued)

	PARAMETER		TEST CO	ONDITIONS	MIN	TYP	MAX	UNIT
	Minimum input voltage for valid F	PG	$I(PG) = 300 \mu A$	V _(PG) ≤ 0.8 V		1.1		V
	Trip threshold voltage		V _O decreasing		79		85	%VO
PG (TPS774xx)	Hysteresis voltage		Measured at VO			0.5		%VO
(11 077400)	Output low voltage		V _I = 2.7 V,	I _(PG) = 1 mA		0.15	0.4	V
	Leakage current		V(PG) = 5 V				1	μΑ
	Minimum input voltage for valid F	RESET	I _(RESET) = 300 μ	A		1.1		V
	Trip threshold voltage		V _O decreasing		92		98	%VO
Reset	Hysteresis voltage	Measured at VO			0.5		%Vo	
(TPS773xx)	Output low voltage		V _I = 2.7 V,	I(RESET) = 1 mA		0.15	0.4	V
	Leakage current		V(RESET) = 5 V				1	μΑ
	RESET time-out delay					220		ms
		0.0.1/ 0	I _O = 250 mA, T	∫J = 25°C		270		
.,	December (see Note 5)	2.8-V Output	I _O = 250 mA				475	
VDO	Dropout voltage (see Note 5)	0.0 \ / 0 \ / 1 \	I _O = 250 mA, T	∫ _J = 25°C		200		
		3.3-V Output	I _O = 250 mA				330	mV
V	Drangut voltage (age Nets 5)	5.0-V Output	$I_0 = 250 \text{ mA}, T$	∫J = 25°C		125		
V _{DO}	O Dropout voltage (see Note 5)		I _O = 250 mA				190	

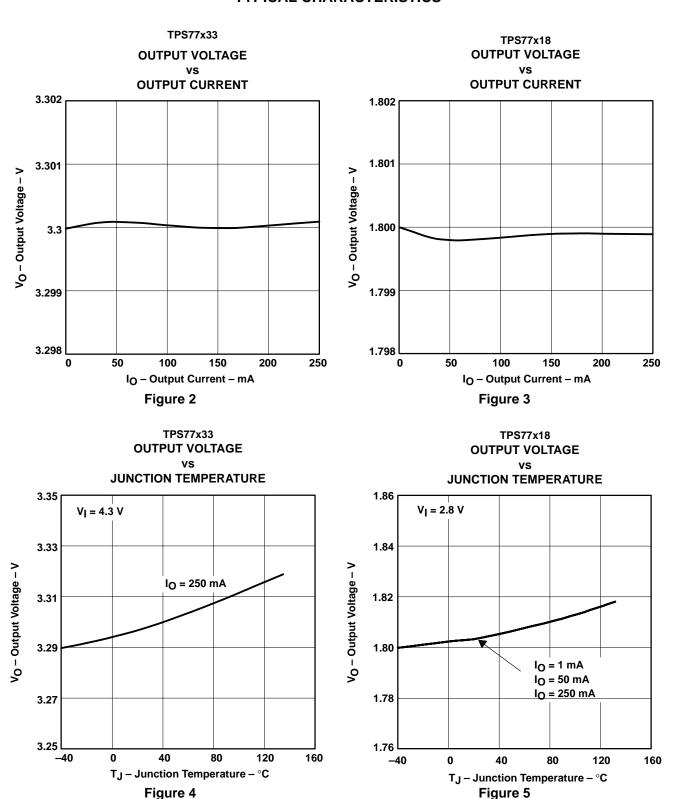
NOTE 5: IN voltage equals V_O(typ) – 100 mV; 1.5 V, 1.6 V, 1.8-V, and 2.7-V dropout voltage limited by input voltage range limitations (i.e., 3.3 V input voltage needs to drop to 3.2 V for purpose of this test).

TYPICAL CHARACTERISTICS

Table of Graphs

•			FIGURE
		vs Output current	2, 3
VO	Output voltage	vs Junction temperature	4, 5
	Ground current	vs Junction temperature	6
	Power supply rejection ratio	vs Frequency	7
	Output spectral noise density	vs Frequency	8
Z _O	Output impedance	vs Frequency	9
.,	5	vs Input voltage	10
V_{DO}	Dropout voltage	vs Junction temperature	11
	Line transient response		12, 14
	Load transient response		13, 15
	Output voltage and enable pulse	vs Time	16
	Equivalent series resistance (ESR)	vs Output current	18 – 21

TYPICAL CHARACTERISTICS



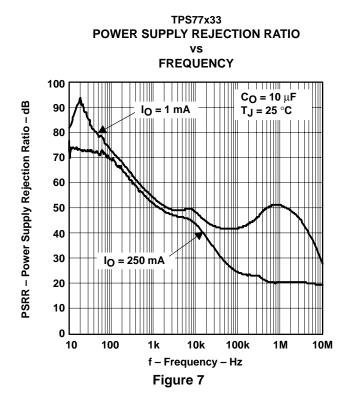


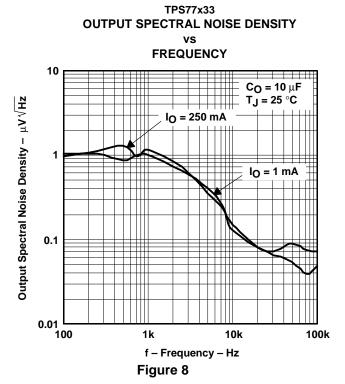
TYPICAL CHARACTERISTICS

TPS77xxx **GROUND CURRENT JUNCTION TEMPERATURE** 115 110 105 Ground Current - µA 100 I_O = 1 mA 95 90 85 I_O = 250 mA 80 10 110 160 -40 60

Figure 6

T_J - Junction Temperature - °C







TYPICAL CHARACTERISTICS

TPS77x33 OUTPUT IMPEDANCE vs FREQUENCY

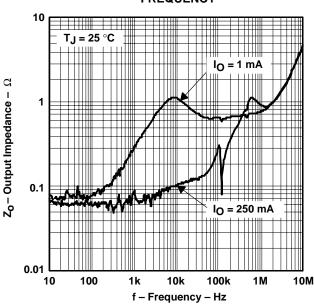
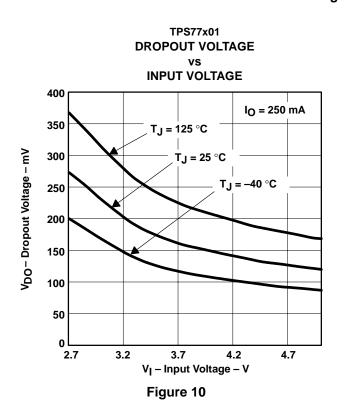
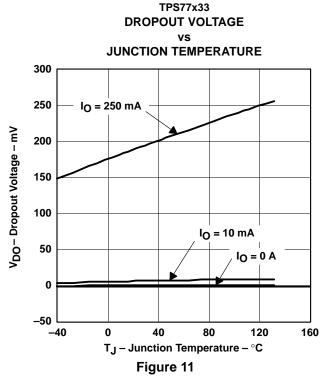
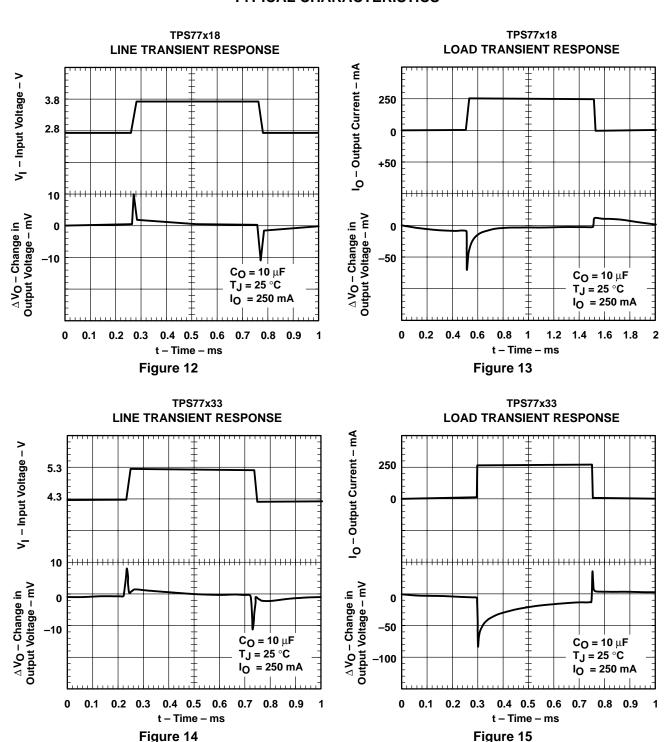


Figure 9





TYPICAL CHARACTERISTICS





TYPICAL CHARACTERISTICS

TPS77x33 OUTPUT VOLTAGE AND ENABLE PULSE

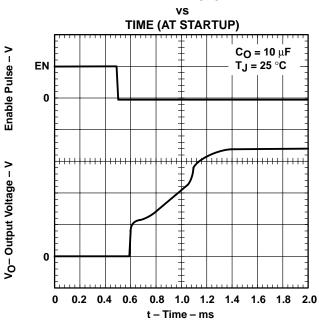


Figure 16

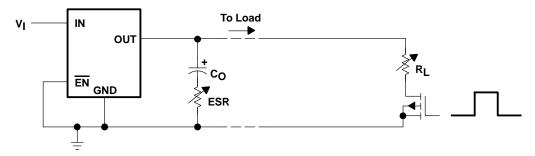
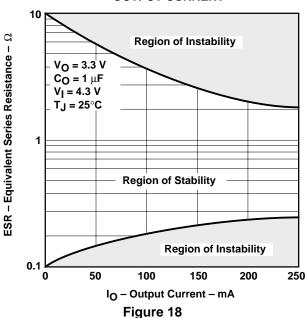


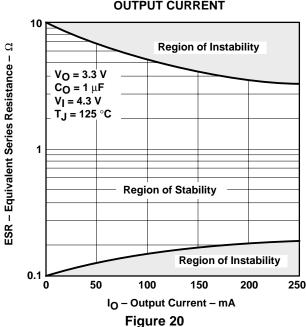
Figure 17. Test Circuit for Typical Regions of Stability (Figures 18 through 21) (Fixed Output Options)

TYPICAL CHARACTERISTICS

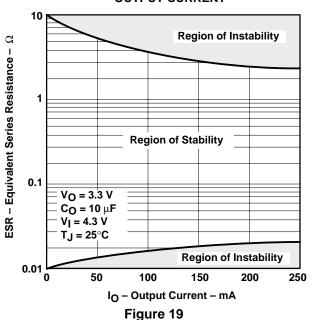
TYPICAL REGION OF STABILITY EQUIVALENT SERIES RESISTANCE[†] vs **OUTPUT CURRENT**



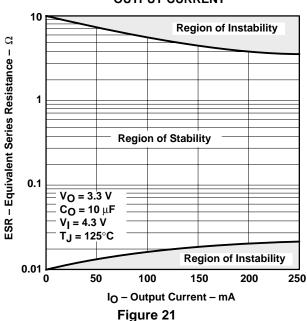
TYPICAL REGION OF STABILITY **EQUIVALENT SERIES RESISTANCE**[†] **OUTPUT CURRENT**



TYPICAL REGION OF STABILITY **EQUIVALENT SERIES RESISTANCE**[†] vs **OUTPUT CURRENT**



TYPICAL REGION OF STABILITY **EQUIVALENT SERIES RESISTANCE**[†] vs **OUTPUT CURRENT**



[†] Equivalent series resistance (ESR) refers to the total series resistance, including the ESR of the capacitor, any series resistance added externally, and PWB trace resistance to Co.



SLVS281E - FEBRUARY 2000 - REVISED JULY 2001

APPLICATION INFORMATION

pin functions

enable (EN)

The $\overline{\mathsf{EN}}$ terminal is an input which enables or shuts down the device. If $\overline{\mathsf{EN}}$ is a logic high, the device will be in shutdown mode. When $\overline{\mathsf{EN}}$ goes to logic low, then the device will be enabled.

power good (PG) (TPS774xx)

The PG terminal is an open drain, active high output that indicates the status of V_{out} (output of the LDO). When V_{out} reaches 82% of the regulated voltage, PG will go to a high-impedance state. It will go to a low-impedance state when V_{out} falls below 82% (i.e. over load condition) of the regulated voltage. The open drain output of the PG terminal requires a pullup resistor

sense (SENSE)

The SENSE terminal of the fixed-output options must be connected to the regulator output, and the connection should be as short as possible. Internally, SENSE connects to a high-impedance wide-bandwidth amplifier through a resistor-divider network and noise pickup feeds through to the regulator output. It is essential to route the SENSE connection in such a way to minimize/avoid noise pickup. Adding RC networks between the SENSE terminal and V_{out} to filter noise is not recommended because it may cause the regulator to oscillate.

feedback (FB)

FB is an input terminal used for the adjustable-output options and must be connected to an external feedback resistor divider. The FB connection should be as short as possible. It is essential to route it in such a way to minimize/avoid noise pickup. Adding RC networks between FB terminal and V_{out} to filter noise is not recommended because it may cause the regulator to oscillate.

reset (RESET) (TPS773xx)

The $\overline{\text{RESET}}$ terminal is an open drain, active low output that indicates the status of V_{out} . When V_{out} reaches 95% of the regulated voltage, $\overline{\text{RESET}}$ will go to a high-impedance state after a 220-ms delay. $\overline{\text{RESET}}$ will go to a low-impedance state when V_{out} is below 95% of the regulated voltage. The open-drain output of the $\overline{\text{RESET}}$ terminal requires a pullup resistor.

external capacitor requirements

An input capacitor is not usually required; however, a bypass capacitor (0.047 μ F or larger) improves load transient response and noise rejection if the TPS773xx or TPS774xx is located more than a few inches from the power supply. A higher-capacitance capacitor may be necessary if large (hundreds of milliamps) load transients with fast rise times are anticipated.

Most low noise LDOs require an external capacitor to further reduce noise. This will impact the cost and board space. The TPS773xx and TPS774xx have very low noise specification requirements without using any external components.

Like all low dropout regulators, the TPS773xx or TPS774xx requires an output capacitor connected between OUT (output of the LDO) and GND (signal ground) to stabilize the internal control loop. The minimum recommended capacitance value is 1 μ F provided the ESR meets the requirement in Figures 19 and 21. In addition, a low-ESR capacitor can be used if the capacitance is at least 10 μ F and the ESR meets the requirements in Figures 18 and 20. Solid tantalum electrolytic, aluminum electrolytic, and multilayer ceramic capacitors are all suitable, provided they meet the requirements described previously.



APPLICATION INFORMATION

external capacitor requirements (continued)

Ceramic capacitors have different types of dielectric material with each exhibiting different temperature and voltage variation. The most common types are X5R, X7R, Y5U, Z5U, and NPO. The NPO type ceramic type capacitors are generally the most stable over temperature. However, the X5R and X7R are also relatively stable over temperature (with the X7R being the more stable of the two) and are therefore acceptable to use. The Y5U and Z5U types provide high capacitance in a small geometry, but exhibit large variations over temperature; therefore, the Y5U and Z5U are not generally recommended for use on this LDO. Independent of which type of capacitor is used, one must make certain that at the worst case condition the capacitance/ESR meets the requirement specified in Figures 18 – 21.

Figure 22 shows the output capacitor and its parasitic impedances in a typical LDO output stage.

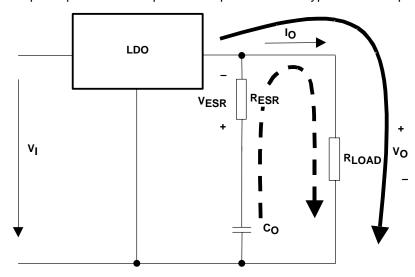


Figure 22. LDO Output Stage With Parasitic Resistances ESR

In steady state (dc state condition), the load current is supplied by the LDO (solid arrow) and the voltage across the capacitor is the same as the output voltage ($V_{Cout} = V_{out}$). This means no current is flowing into the C_{out} branch. If I_{out} suddenly increases (transient condition), the following occurs:

- The LDO is not able to supply the sudden current need due to its response time (t₁ in Figure 23). Therefore, capacitor C_{out} provides the current for the new load condition (dashed arrow). C_{out} now acts like a battery with an internal resistance, ESR. Depending on the current demand at the output, a voltage drop will occur at R_{ESR}. This voltage is shown as V_{ESR} in Figure 22.
- When C_{out} is conducting current to the load, initial voltage at the load will be V_{out} = V_{Cout} V_{ESR}. Due to the discharge of C_{out}, the output voltage V_{out} will drop continuously until the response time t₁ of the LDO is reached and the LDO will resume supplying the load. From this point, the output voltage starts rising again until it reaches the regulated voltage. This period is shown as t₂ in Figure 23.

The figure also shows the impact of different ESRs on the output voltage. The left brackets show different levels of ESRs where number 1 displays the lowest and number 3 displays the highest ESR.

From above, the following conclusions can be drawn:

- The higher the ESR, the larger the droop at the beginning of load transient.
- The smaller the output capacitor, the faster the discharge time and the bigger the voltage droop during the LDO response period.



APPLICATION INFORMATION

conclusion

To minimize the transient output droop, capacitors must have a low ESR and be large enough to support the minimum output voltage requirement.

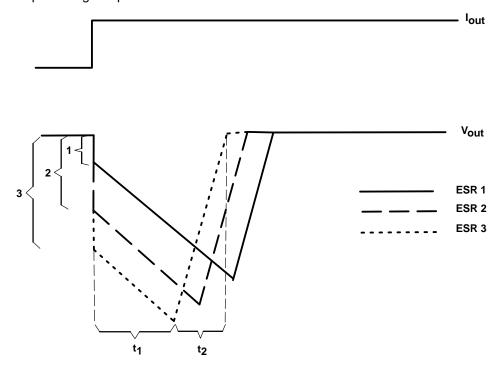


Figure 23. Correlation of Different ESRs and Their Influence to the Regulation of V_{out} at a Load Step From Low-to-High Output Current

APPLICATION INFORMATION

programming the TPS77x01 adjustable LDO regulator

The output voltage of the TPS77x01 adjustable regulator is programmed using an external resistor divider as shown in Figure 24. The output voltage is calculated using:

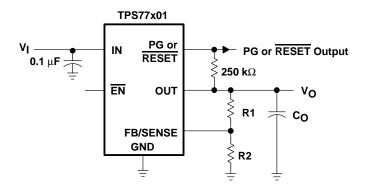
$$V_{O} = V_{ref} \times \left(1 + \frac{R1}{R2}\right) \tag{1}$$

Where:

 $V_{ref} = 1.1834 \text{ V}$ typ (the internal reference voltage)

Resistors R1 and R2 should be chosen for approximately $50-\mu A$ divider current. Lower value resistors can be used but offer no inherent advantage and waste more power. Higher values should be avoided, as leakage currents at FB increase the output voltage error. The recommended design procedure is to choose R2 = $30.1 \text{ k}\Omega$ to set the divider current at $50 \mu A$ and then calculate R1 using:

$$R1 = \left(\frac{V_{O}}{V_{ref}} - 1\right) \times R2 \tag{2}$$



OUTPUT VOLTAGE PROGRAMMING GUIDE

OUTPUT VOLTAGE	R1	R2	UNIT
2.5 V	33.5	30.1	kΩ
3.3 V	53.8	30.1	kΩ
3.6 V	61.5	30.1	kΩ

NOTE: To reduce noise and prevent oscillation, R1 and R2 need to be as close as possible to the FB/SENSE terminal.

Figure 24. TPS77x01 Adjustable LDO Regulator Programming



SLVS281E - FEBRUARY 2000 - REVISED JULY 2001

APPLICATION INFORMATION

regulator protection

The TPS773xx or TPS774xx PMOS-pass transistor has a built-in back diode that conducts reverse currents when the input voltage drops below the output voltage (e.g., during power down). Current is conducted from the output to the input and is not internally limited. When extended reverse voltage is anticipated, external limiting may be appropriate.

The TPS773xx or TPS774xx also features internal current limiting and thermal protection. During normal operation, the TPS773xx or TPS774xx limits output current to approximately 0.9 A. When current limiting engages, the output voltage scales back linearly until the overcurrent condition ends. While current limiting is designed to prevent gross device failure, care should be taken not to exceed the power dissipation ratings of the package. If the temperature of the device exceeds 150°C(typ), thermal-protection circuitry shuts it down. Once the device has cooled below 130°C(typ), regulator operation resumes.

power dissipation and junction temperature

Specified regulator operation is assured to a junction temperature of 125° C; the maximum junction temperature should be restricted to 125° C under normal operating conditions. This restriction limits the power dissipation the regulator can handle in any given application. To ensure the junction temperature is within acceptable limits, calculate the maximum allowable dissipation, $P_{D(max)}$, and the actual dissipation, P_{D} , which must be less than or equal to $P_{D(max)}$.

The maximum-power-dissipation limit is determined using the following equation:

$$P_{D(max)} = \frac{T_{J}max - T_{A}}{R_{H,IA}}$$

Where:

T_.Imax is the maximum allowable junction temperature.

 $R_{\theta JA}$ is the thermal resistance junction-to-ambient for the package, i.e., 266.2°C/W for the 8-terminal MSOP with no airflow.

T_A is the ambient temperature.

The regulator dissipation is calculated using:

$$P_{D} = (V_{I} - V_{O}) \times I_{O}$$

Power dissipation resulting from quiescent current is negligible. Excessive power dissipation will trigger the thermal protection circuit.



16-Aug-2012

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/ Ball Finish	MSL Peak Temp ⁽³⁾	Samples (Requires Login)
TPS77301DGK	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77301DGKG4	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77301DGKR	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77301DGKRG4	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77315DGK	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77315DGKG4	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77315DGKR	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77315DGKRG4	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77316DGK	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77316DGKG4	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77316DGKR	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77316DGKRG4	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77318DGK	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77318DGKG4	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77318DGKR	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77318DGKRG4	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77327DGK	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	



16-Aug-2012

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/ Ball Finish	MSL Peak Temp ⁽³⁾	Samples (Requires Login)
TPS77327DGKG4	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77328DGK	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77328DGKG4	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77333DGK	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77333DGKG4	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77333DGKR	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77333DGKRG4	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77350DGK	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77350DGKG4	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77350DGKR	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77350DGKRG4	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77401DGK	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77401DGKG4	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77401DGKR	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77401DGKRG4	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77415DGK	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77415DGKG4	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77418DGK	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	





16-Aug-2012

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/ Ball Finish	MSL Peak Temp ⁽³⁾	Samples (Requires Login)
TPS77418DGKG4	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77428DGK	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77428DGKG4	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77433DGK	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77433DGKG4	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77433DGKR	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77433DGKRG4	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77450DGK	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TPS77450DGKG4	ACTIVE	VSSOP	DGK	8	80	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.



PACKAGE OPTION ADDENDUM

16-Aug-2012

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OTHER QUALIFIED VERSIONS OF TPS77401:

● Enhanced Product: TPS77401-EP

NOTE: Qualified Version Definitions:

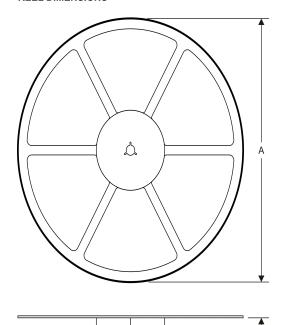
• Enhanced Product - Supports Defense, Aerospace and Medical Applications

PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION

REEL DIMENSIONS



TAPE DIMENSIONS



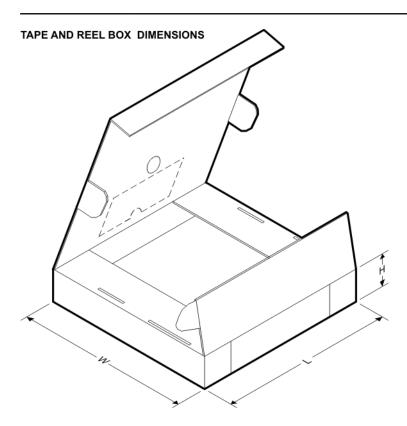
A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

TAPE AND REEL INFORMATION

*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS77301DGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TPS77315DGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TPS77316DGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TPS77318DGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TPS77333DGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TPS77350DGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TPS77401DGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TPS77433DGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1

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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS77301DGKR	VSSOP	DGK	8	2500	358.0	335.0	35.0
TPS77315DGKR	VSSOP	DGK	8	2500	358.0	335.0	35.0
TPS77316DGKR	VSSOP	DGK	8	2500	358.0	335.0	35.0
TPS77318DGKR	VSSOP	DGK	8	2500	358.0	335.0	35.0
TPS77333DGKR	VSSOP	DGK	8	2500	358.0	335.0	35.0
TPS77350DGKR	VSSOP	DGK	8	2500	358.0	335.0	35.0
TPS77401DGKR	VSSOP	DGK	8	2500	358.0	335.0	35.0
TPS77433DGKR	VSSOP	DGK	8	2500	358.0	335.0	35.0

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